

# EE101: Bode plots

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- \* Bel turned out to be too large in practice  $\rightarrow$  deciBel (i.e., one tenth of a Bel).

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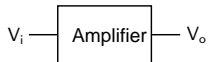
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- \* The voltage gain of an amplifier is

$$A_V \text{ in dB} = 20 \log (V_o/V_i),$$

with  $V_i$  serving as the reference voltage.

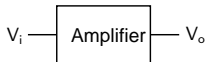
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Given  $V_i = 2.5 \text{ mV}$  and  $A_v = 36.3 \text{ dB}$ ,  
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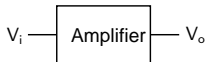
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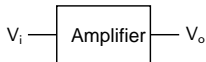
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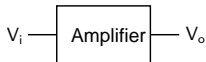
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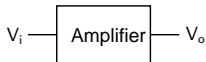
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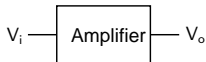
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$$A_V = 36.3 \text{ dB}$$

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$$V_o = A_V \times V_i = 65 \times 2.5 \text{ mV} = 162.5 \text{ mV}.$$

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## dB in audio measurements

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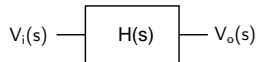
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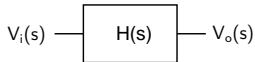
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windows break	163 dB

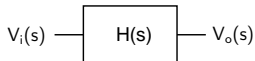




- \* The transfer function of a circuit such as an amplifier or a filter is given by,

$$H(s) = V_o(s)/V_i(s), \quad s = j\omega.$$

$$\text{e.g., } H(s) = \frac{K}{1 + s\tau} = \frac{K}{1 + j\omega\tau}$$

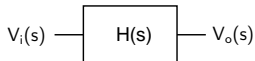


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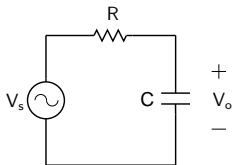
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- \* Bode gave simple rules which allow construction of the above “Bode plots” in an approximate (asymptotic) manner.

## A simple transfer function

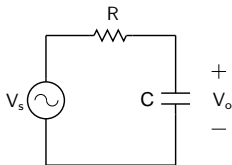


$$V_o = \frac{(1/sC)}{R + (1/sC)} V_s,$$

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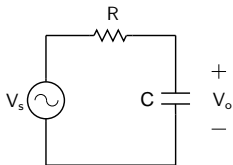
$$\omega_0 = \frac{1}{RC}.$$

- \* The circuit behaves like a low-pass filter.

For  $\omega \ll \omega_0$ ,  $|H(j\omega)| \rightarrow 1$ .

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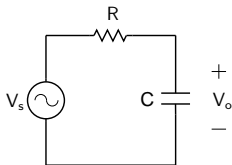
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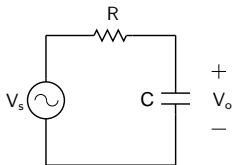
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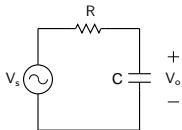
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- \* The magnitude ( $|H|$ ) varies by orders of magnitude as well.  
The phase ( $\angle H$ ) varies from 0 (for  $\omega \ll \omega_0$ ) to  $-\pi/2$  (for  $\omega \gg \omega_0$ ).

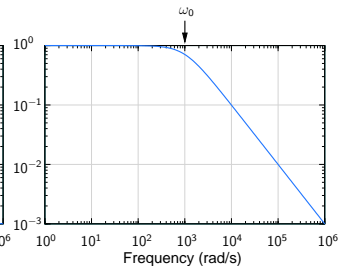
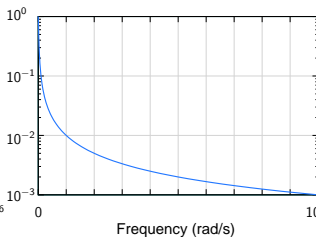
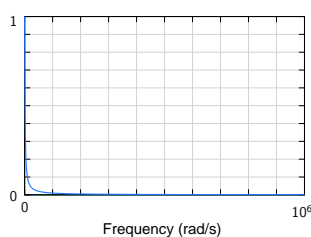
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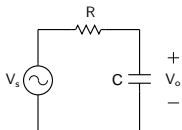
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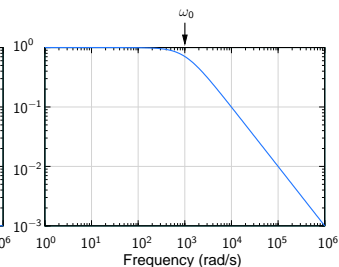
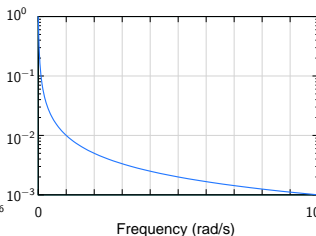
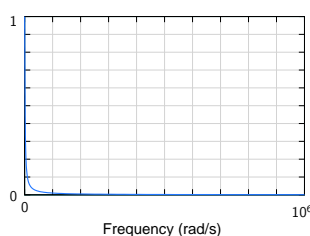
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$$\rightarrow H(s) = \frac{1}{1+sRC} = \frac{1}{1+(j\omega/\omega_0)},$$

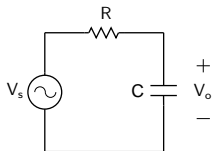
$$\omega_0 = \frac{1}{RC}.$$



Since  $\omega$  and  $|H(j\omega)|$  vary by several orders of magnitude, a linear  $\omega$ - or  $|H|$ -axis is not appropriate  $\rightarrow \log |H|$  is plotted against  $\log \omega$ .



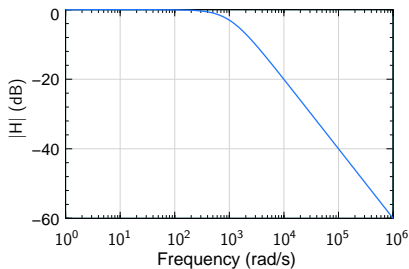
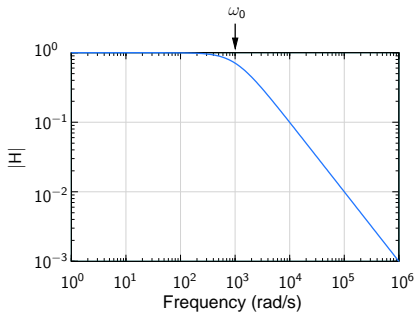
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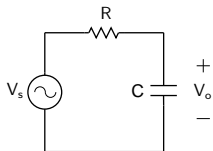
$$V_o = \frac{(1/sC)}{R + (1/sC)} V_s,$$

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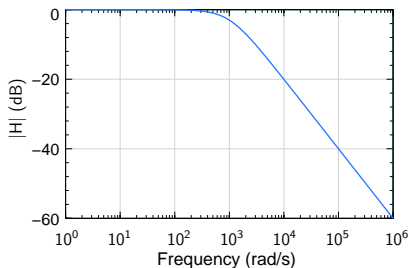
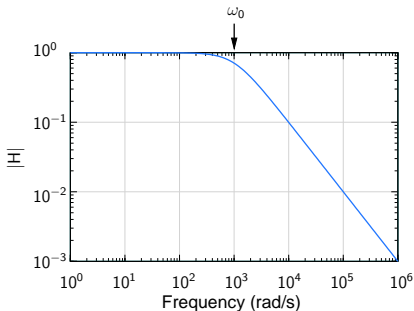
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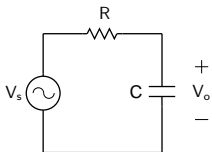
$$\omega_0 = \frac{1}{RC}.$$



Note that the *shape* of the plot does not change.

$|H|$  (dB) =  $20 \log |H|$  is simply a scaled version of  $\log |H|$ .

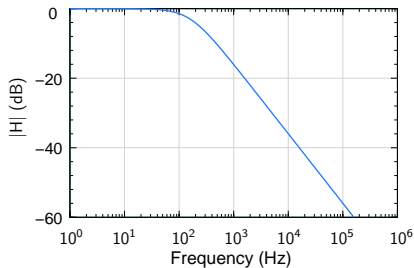
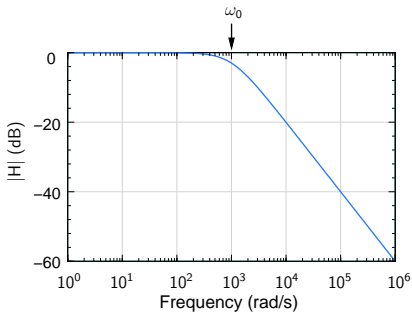
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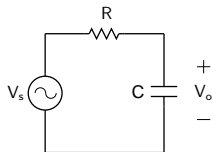
$$V_o = \frac{(1/sC)}{R + (1/sC)} V_s,$$

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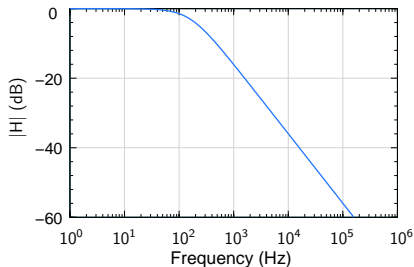
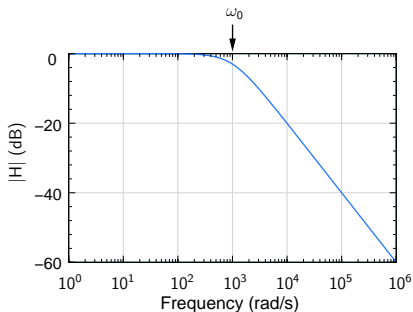
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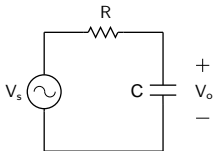


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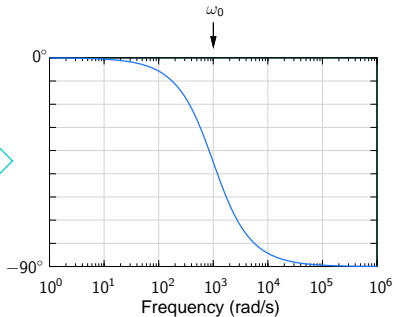
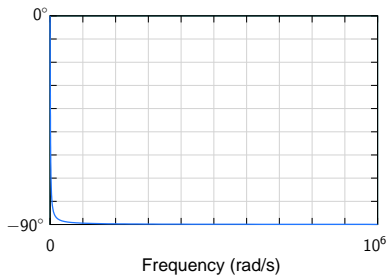


Since  $\omega = 2\pi f$ , the *shape* of the plot does not change.

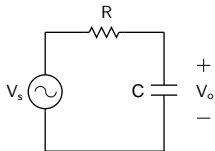
## A simple transfer function: phase



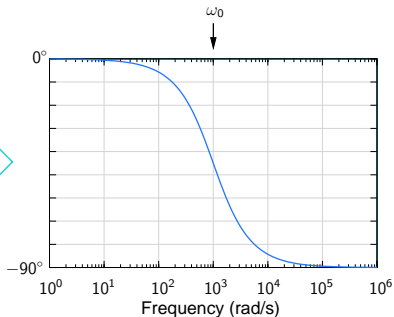
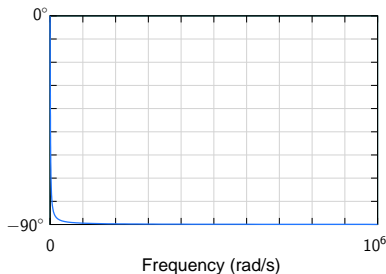
$$V_o = \frac{(1/sC)}{R + (1/sC)} V_s,$$
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## A simple transfer function: phase

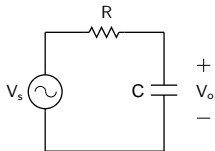


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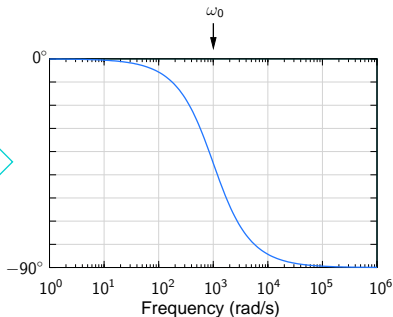
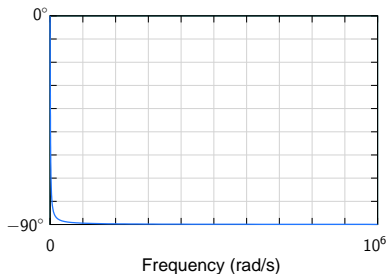


- \* Since  $\angle H = -\tan^{-1}(\omega/\omega_0)$  varies in a limited range ( $0^\circ$  to  $-90^\circ$  in this example), a linear axis is appropriate for  $\angle H$ .

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- \* As in the magnitude plot, we use a log axis for  $\omega$ , since we are interested in a wide range of  $\omega$ .

# Construction of Bode plots

Consider  $H(s) = \frac{K (1 + s/z_1)(1 + s/z_2) \cdots (1 + s/z_M)}{(1 + s/p_1)(1 + s/p_2) \cdots (1 + s/p_N)}$ .



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$-z_1, -z_2, \dots$  are called the “zeros” of  $H(s)$ .

$-p_1, -p_2, \dots$  are called the “poles” of  $H(s)$ .

(In addition, there could be terms like  $s, s^2, \dots$  in the numerator.)

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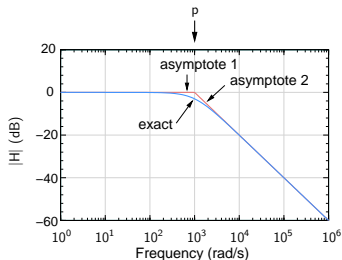
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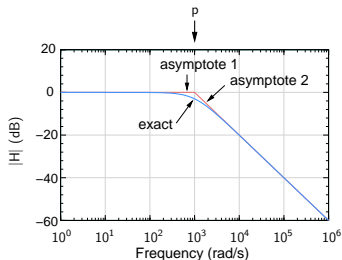
- (a) computing approximate contribution of each pole/zero as a function  $\omega$ .
- (b) combining the various contributions to obtain  $|H|$  and  $\angle H$  versus  $\omega$ .

## Contribution of a pole: magnitude



$$\text{Consider } H(s) = \frac{1}{1 + s/p} \rightarrow H(j\omega) = \frac{1}{1 + j(\omega/p)}, |H(j\omega)| = \frac{1}{\sqrt{1 + (\omega/p)^2}}.$$

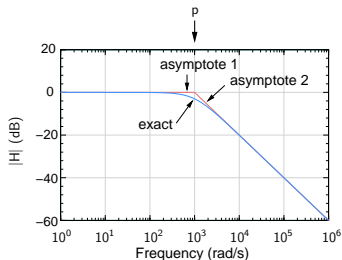
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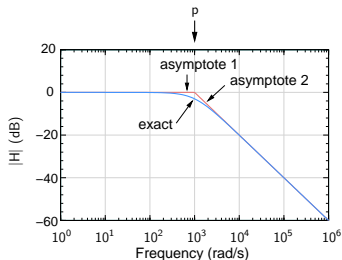


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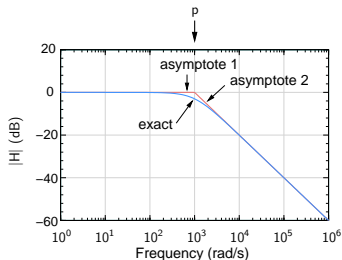
Asymptote 2:  $\omega \gg p$ :  $|H| \rightarrow \frac{1}{\omega/p} = \frac{p}{\omega} \rightarrow |H| = 20 \log p - 20 \log \omega \text{ (dB)}$

Consider two values of  $\omega$ :  $\omega_1$  and  $10\omega_1$ .

$$|H|_1 = 20 \log p - 20 \log \omega_1 \text{ (dB)}$$

$$|H|_2 = 20 \log p - 20 \log (10\omega_1) \text{ (dB)}$$

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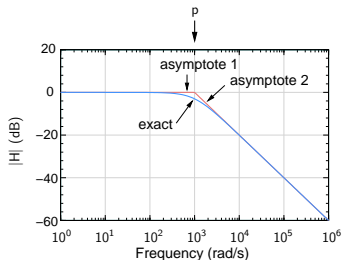
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Consider two values of  $\omega$ :  $\omega_1$  and  $10\omega_1$ .

$$|H|_1 = 20 \log p - 20 \log \omega_1 \text{ (dB)}$$

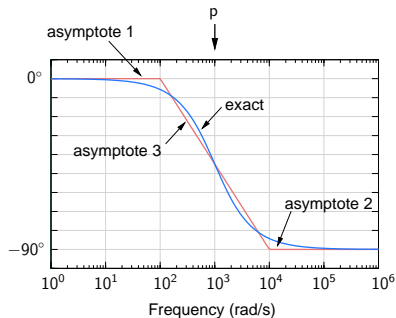
$$|H|_2 = 20 \log p - 20 \log (10 \omega_1) \text{ (dB)}$$

$$|H|_1 - |H|_2 = -20 \log \frac{\omega_1}{10\omega_1} = 20 \text{ dB.}$$

→  $|H|$  versus  $\omega$  has a slope of  $-20$  dB/decade.

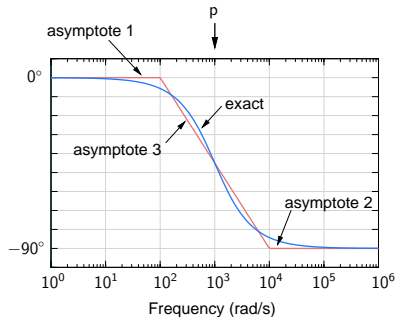
Note that, at  $\omega = p$ , the actual value of  $|H|$  is  $1/\sqrt{2}$  (i.e.,  $-3$  dB).

## Contribution of a pole: phase



$$\text{Consider } H(s) = \frac{1}{1 + s/p} = \frac{1}{1 + j(\omega/p)} \rightarrow \angle H = -\tan^{-1} \left( \frac{\omega}{p} \right)$$

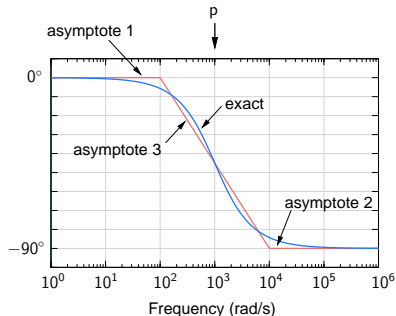
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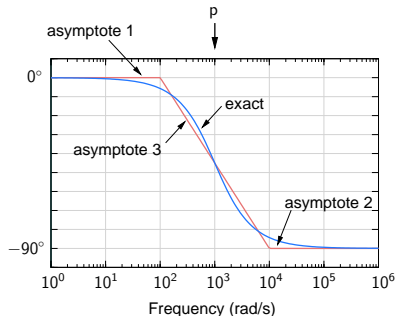


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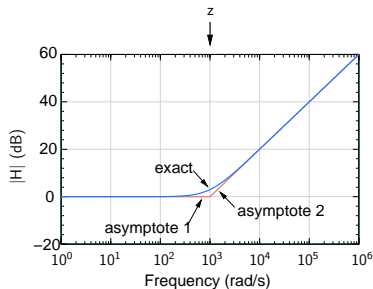
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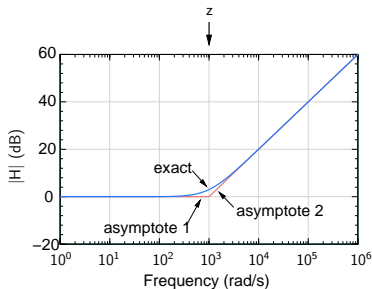
Asymptote 3: For  $p/10 < \omega < 10p$ ,  $\angle H$  is assumed to vary linearly with  $\log \omega$   
 $\rightarrow$  at  $\omega = p$ ,  $\angle H = -\pi/4$  (which is also the actual value of  $\angle H$ ).

## Contribution of a zero: magnitude



Consider  $H(s) = 1 + s/z \rightarrow H(j\omega) = 1 + j(\omega/z)$ ,  $|H(j\omega)| = \sqrt{1 + (\omega/z)^2}$ .

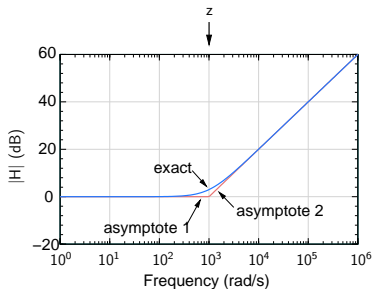
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## Contribution of a zero: magnitude



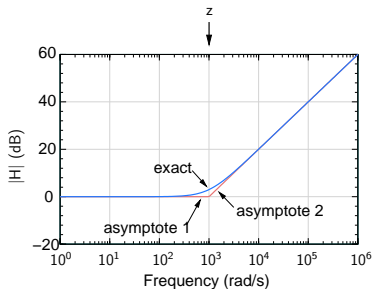
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Asymptote 2:  $\omega \gg z$ :  $|H| \rightarrow \frac{\omega}{z} \rightarrow |H| = 20 \log \omega - 20 \log z$  (dB)



## Contribution of a zero: magnitude



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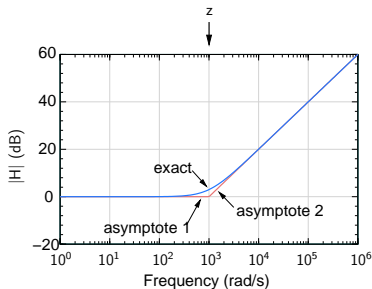
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Consider two values of  $\omega$ :  $\omega_1$  and  $10\omega_1$ .

$$|H|_1 = 20 \log \omega_1 - 20 \log z \text{ (dB)}$$

$$|H|_2 = 20 \log (10\omega_1) - 20 \log z \text{ (dB)}$$

## Contribution of a zero: magnitude



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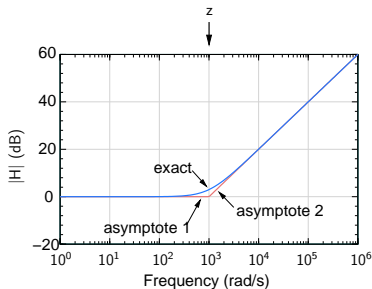
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## Contribution of a zero: magnitude



Consider  $H(s) = 1 + s/z \rightarrow H(j\omega) = 1 + j(\omega/z)$ ,  $|H(j\omega)| = \sqrt{1 + (\omega/z)^2}$ .

Asymptote 1:  $\omega \ll z$ :  $|H| \rightarrow 1$ ,  $20 \log |H| = 0$  dB.

Asymptote 2:  $\omega \gg z$ :  $|H| \rightarrow \frac{\omega}{z} \rightarrow |H| = 20 \log \omega - 20 \log z$  (dB)

Consider two values of  $\omega$ :  $\omega_1$  and  $10\omega_1$ .

$$|H|_1 = 20 \log \omega_1 - 20 \log z \text{ (dB)}$$

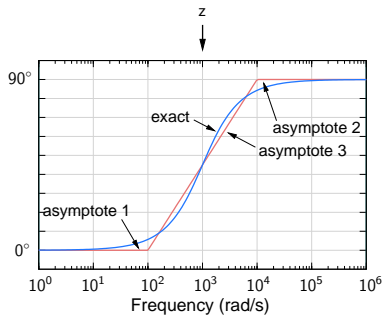
$$|H|_2 = 20 \log (10\omega_1) - 20 \log z \text{ (dB)}$$

$$|H|_1 - |H|_2 = 20 \log \frac{\omega_1}{10\omega_1} = -20 \text{ dB.}$$

$\rightarrow |H|$  versus  $\omega$  has a slope of +20 dB/decade.

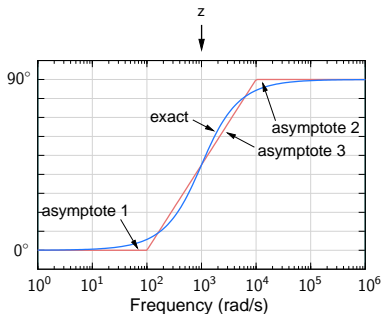
Note that, at  $\omega = z$ , the actual value of  $|H|$  is  $\sqrt{2}$  (i.e., 3 dB).

## Contribution of a zero: phase



Consider  $H(s) = 1 + s/z = 1 + j(\omega/z) \rightarrow \angle H = \tan^{-1} \left( \frac{\omega}{z} \right)$

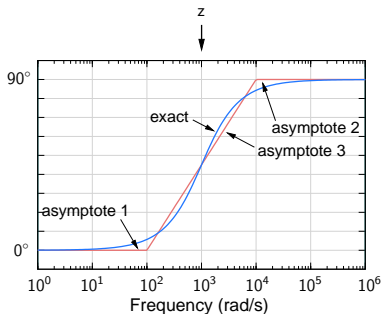
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## Contribution of a zero: phase

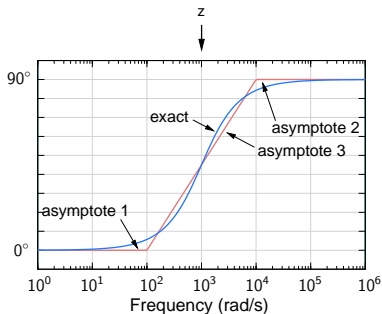


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Asymptote 3: For  $z/10 < \omega < 10z$ ,  $\angle H$  is assumed to vary linearly with  $\log \omega$   
 $\rightarrow$  at  $\omega = z$ ,  $\angle H = \pi/4$  (which is also the actual value of  $\angle H$ ).

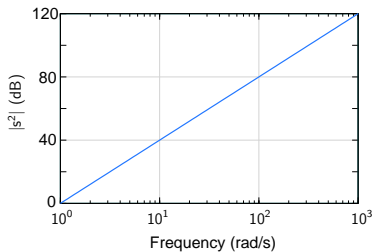
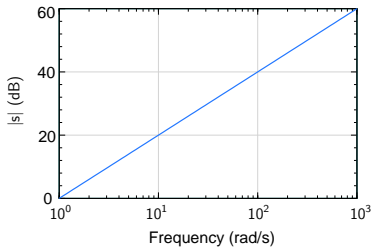
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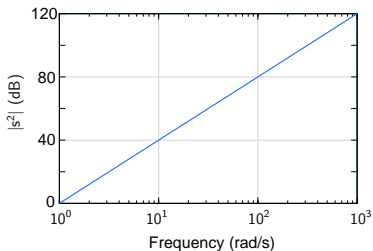
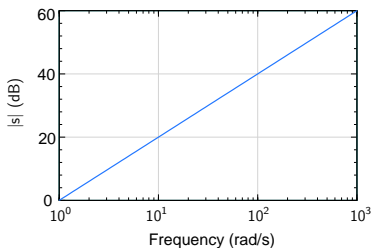
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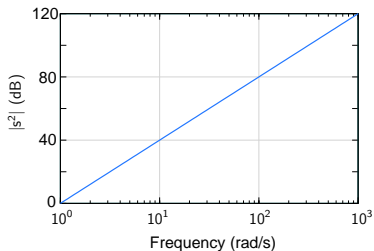
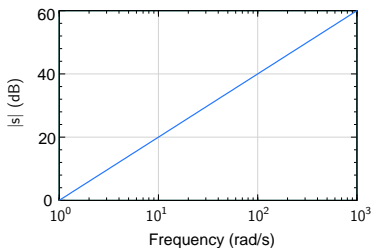
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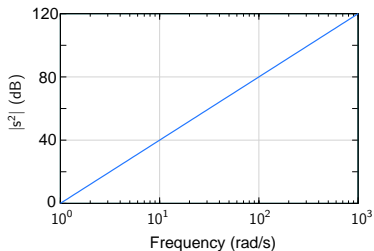
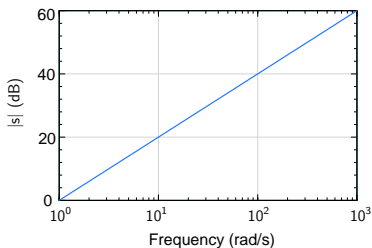


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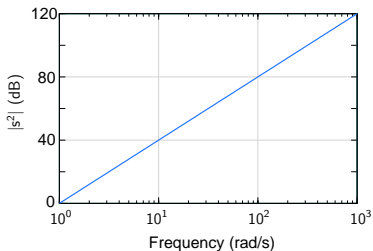
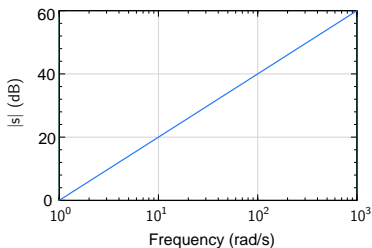
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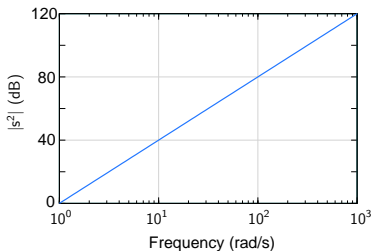
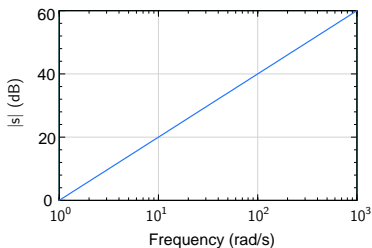
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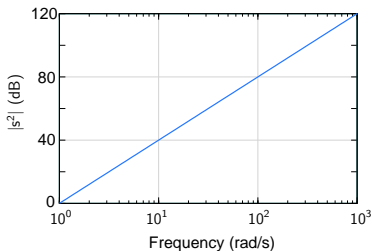
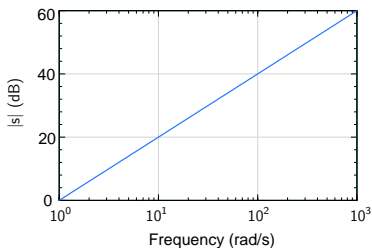
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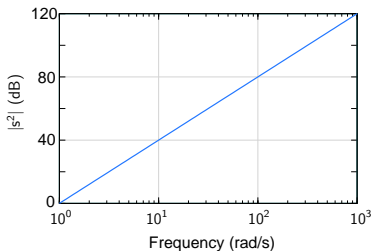
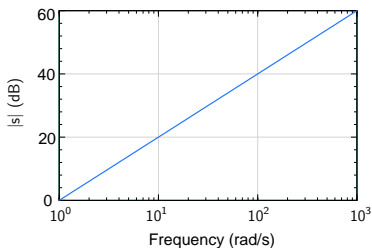
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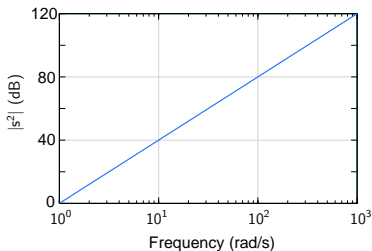
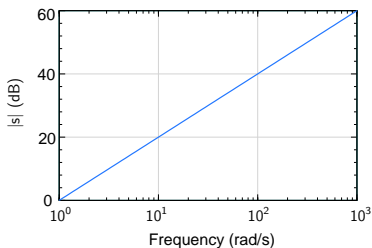
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Phase:

$H_1(j\omega)$  and  $H_2(j\omega)$  are complex numbers.

At a given  $\omega$ , let  $H_1 = K_1 \angle \alpha = K_1 e^{j\alpha}$ , and  $H_2 = K_2 \angle \beta = K_2 e^{j\beta}$ .

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In the Bode phase plot, the contributions due to  $H_1$  and  $H_2$  also get added.

The same reasoning applies to more than two terms as well.

## Combining different terms: example

Consider  $H(s) = \frac{10 s}{(1 + s/10^2)(1 + s/10^5)}$ .



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Consider  $H(s) = \frac{10 s}{(1 + s/10^2)(1 + s/10^5)}$ .

Let  $H(s) = H_1(s) H_2(s) H_3(s) H_4(s)$ , where

$$H_1(s) = 10,$$

$$H_2(s) = s,$$

$$H_3(s) = \frac{1}{1 + s/p_1}, p_1 = 10^2 \text{ rad/s},$$

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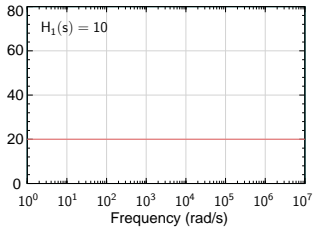
$$H_2(s) = s,$$

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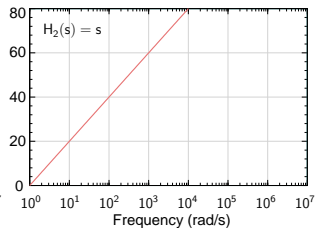
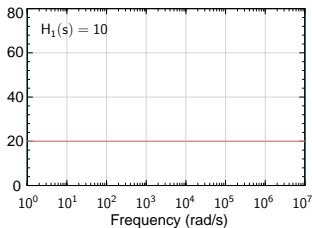
$$H_4(s) = \frac{1}{1 + s/p_2}, p_2 = 10^5 \text{ rad/s}.$$

We can now plot the magnitude and phase of  $H_1$ ,  $H_2$ ,  $H_3$ ,  $H_4$  *individually* versus  $\omega$  and then simply add them to obtain  $|H|$  and  $\angle H$ .

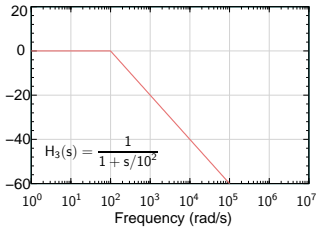
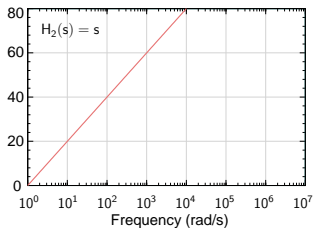
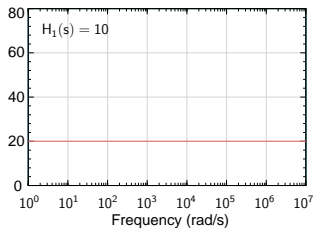
## Magnitude plot ( $|H|$ in dB)



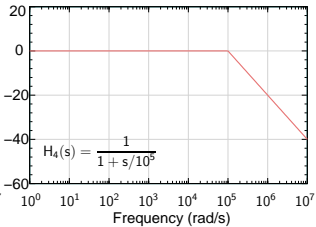
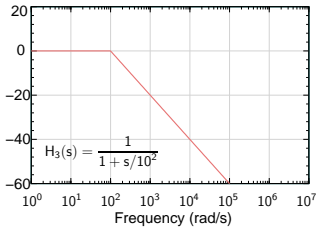
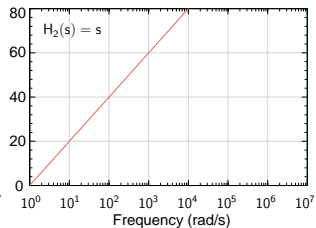
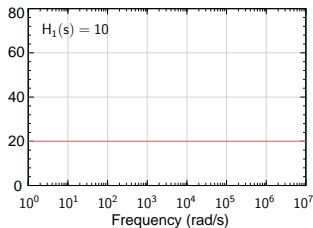
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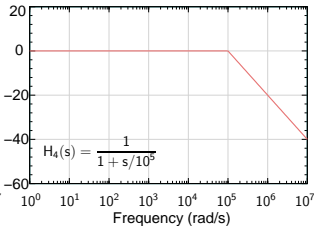
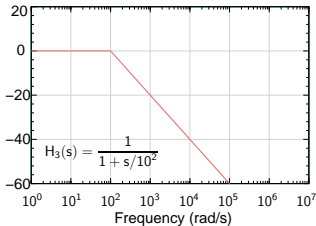
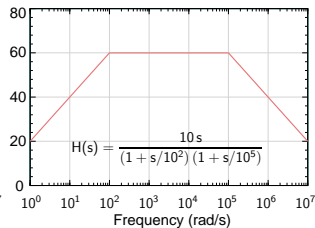
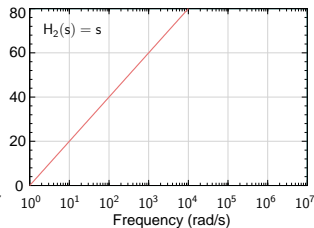
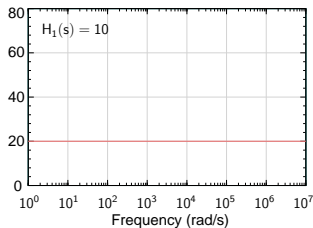
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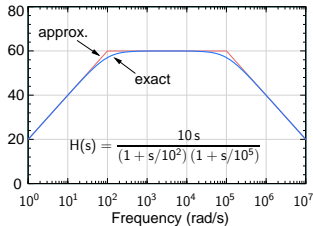
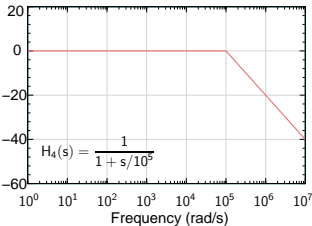
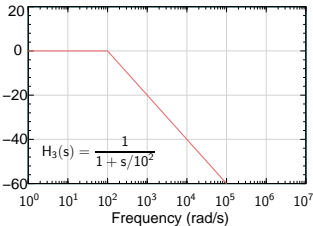
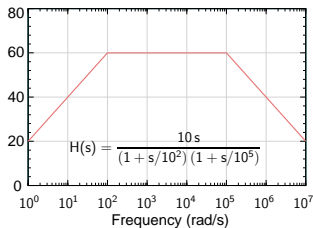
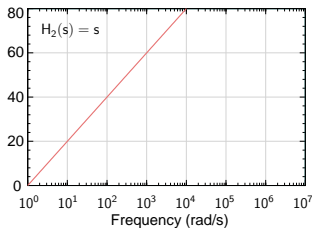
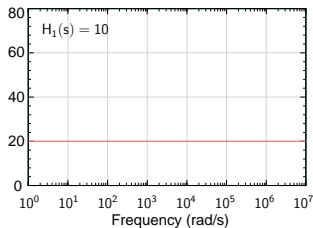
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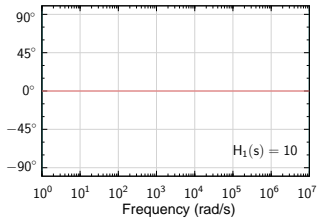


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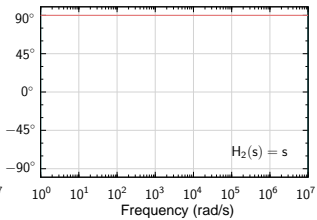
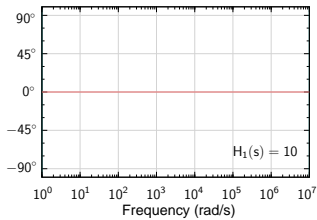




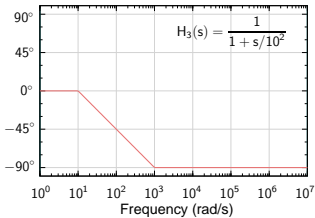
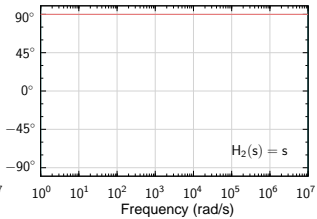
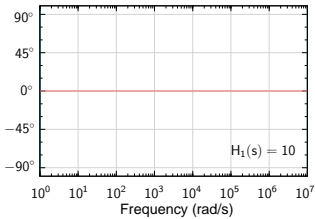
# Phase plot



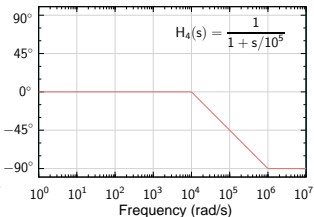
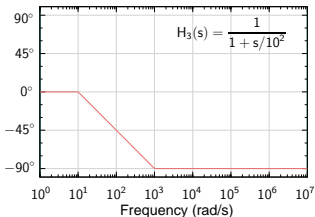
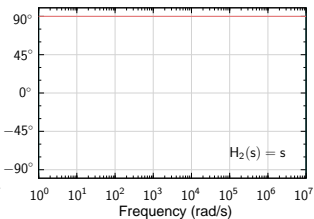
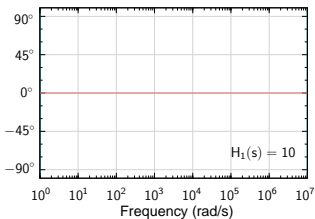
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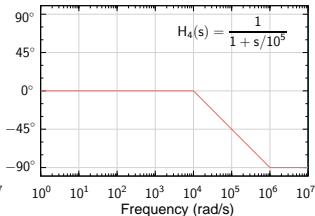
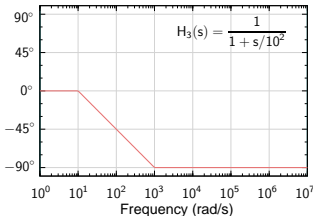
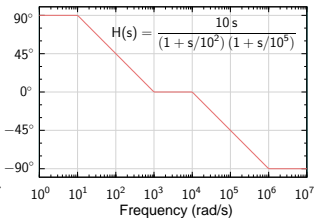
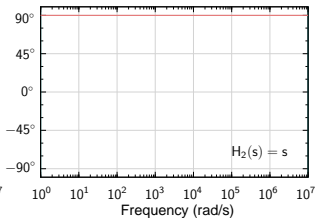
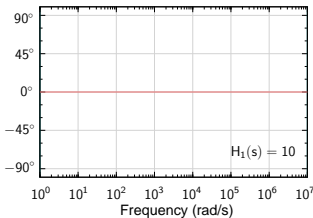
# Phase plot



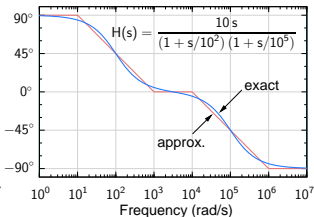
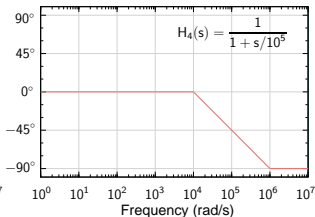
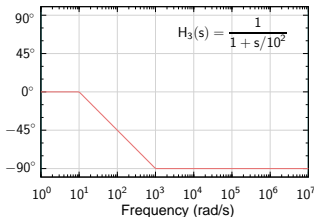
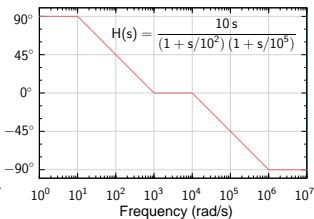
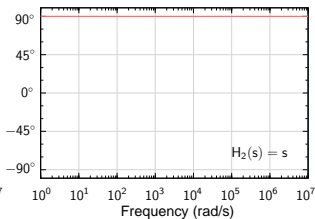
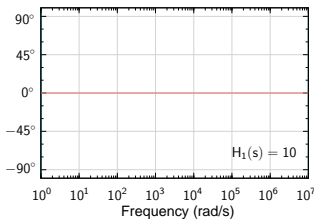
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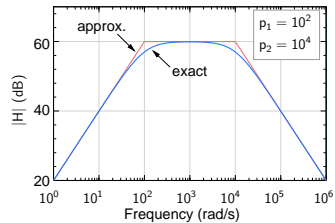
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- \* When the poles and zeros are not sufficiently separated, the Bode approximation should be used only for a rough estimate, followed by a numerical calculation. However, even in such cases, it does give a good idea of the *asymptotic* magnitude and phase plots, which is valuable in amplifier design.

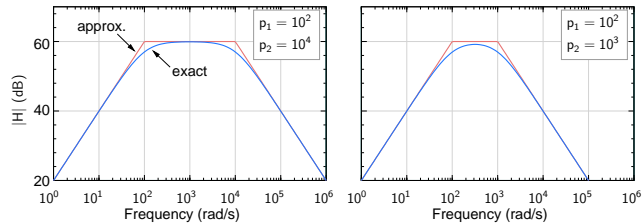
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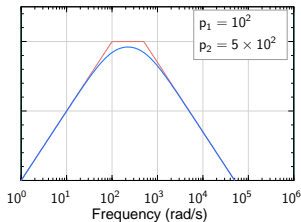
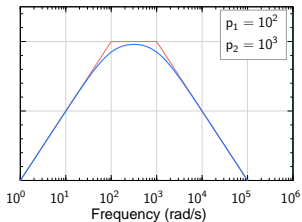
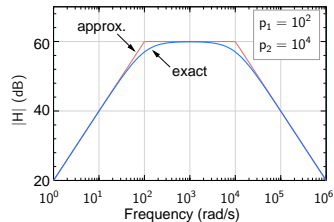
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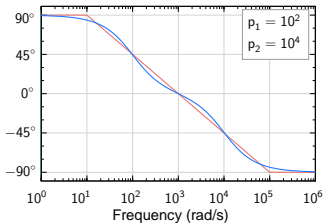
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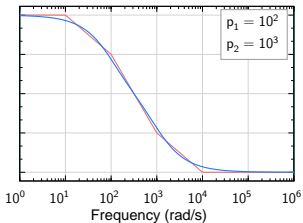
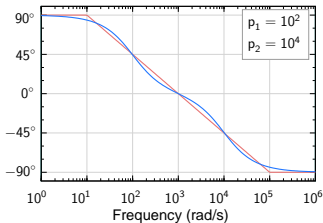
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