

EE101: Bode plots



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- Bell considered the telephone an intrusion and refused to put one in his office.
- * Bel turned out to be too large in practice \rightarrow deciBel (i.e., one tenth of a Bel).

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- * The voltage gain of an amplifier is
 $A_V \text{ in dB} = 20 \log (V_o/V_i)$,
with V_i serving as the reference voltage.

Example



Given $V_i = 2.5 \text{ mV}$ and $A_V = 36.3 \text{ dB}$,
compute V_o in dBm and in mV.

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- * When sound intensity is specified in dB, the reference pressure is $P_{\text{ref}} = 20 \mu\text{Pa}$ (our hearing threshold).

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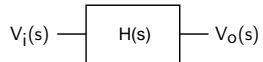
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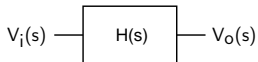
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windows break	163 dB

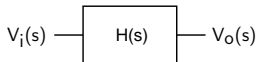




- * The transfer function of a circuit such as an amplifier or a filter is given by,

$$H(s) = V_o(s)/V_i(s), \quad s = j\omega.$$

$$\text{e.g., } H(s) = \frac{K}{1 + s\tau} = \frac{K}{1 + j\omega\tau}$$

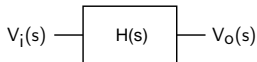


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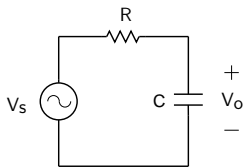
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- * Bode gave simple rules which allow construction of the above “Bode plots” in an approximate (asymptotic) manner.

A simple transfer function

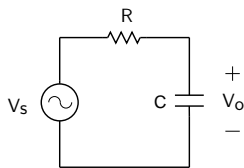


$$V_o = \frac{(1/sC)}{R + (1/sC)} V_s,$$

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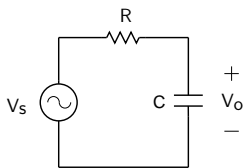
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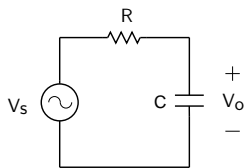
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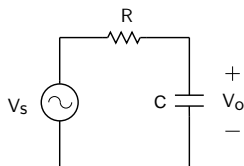
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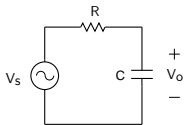
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- * We are generally interested in a large variation in ω (several orders), and its effect on $|H|$ and $\angle H$.
- * The magnitude ($|H|$) varies by orders of magnitude as well.
The phase ($\angle H$) varies from 0 (for $\omega \ll \omega_0$) to $-\pi/2$ (for $\omega \gg \omega_0$).

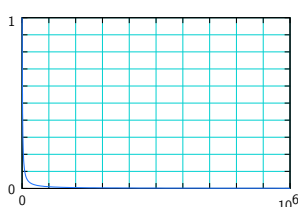
A simple transfer function: magnitude



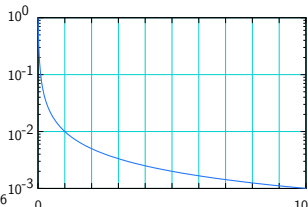
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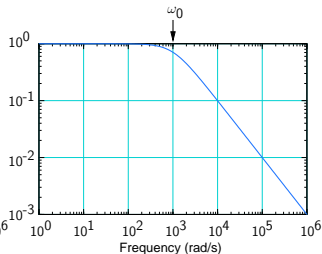
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Frequency (rad/s)

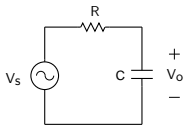


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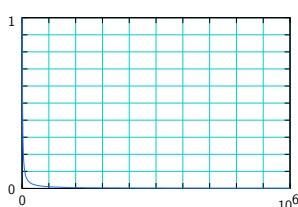
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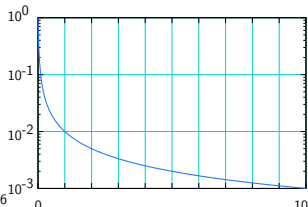
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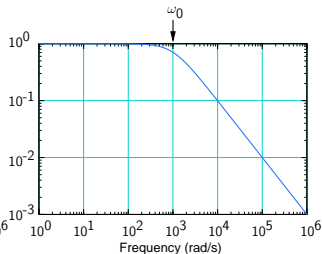
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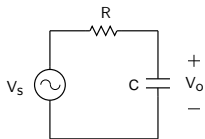
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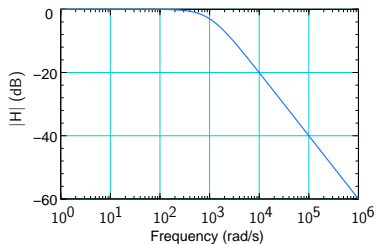
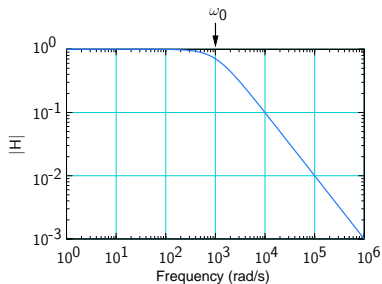
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Since ω and $|H(j\omega)|$ vary by several orders of magnitude, a linear ω - or $|H|$ -axis is not appropriate $\rightarrow \log |H|$ is plotted against $\log \omega$.

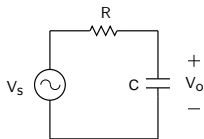
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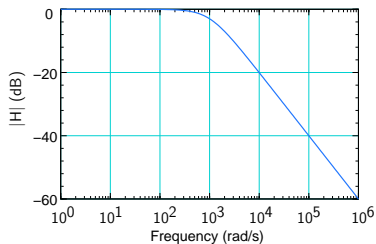
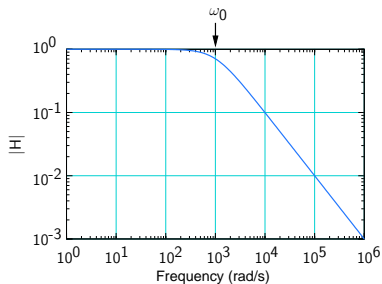
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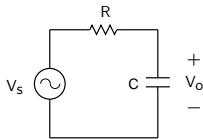
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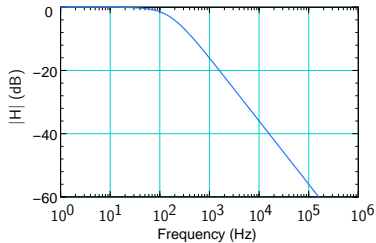
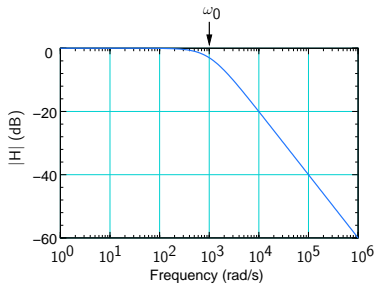
Note that the *shape* of the plot does not change.

$|H|$ (dB) = $20 \log |H|$ is simply a scaled version of $\log |H|$.

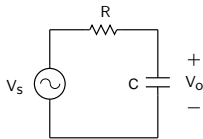
A simple transfer function: magnitude



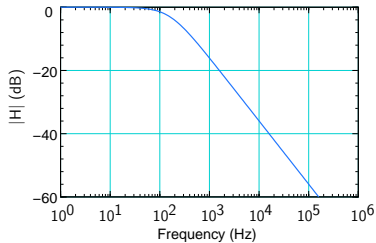
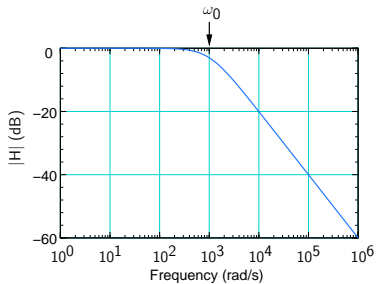
$$V_o = \frac{(1/sC)}{R + (1/sC)} V_s,$$
$$\rightarrow H(s) = \frac{1}{1+sRC} = \frac{1}{1+(j\omega/\omega_0)},$$
$$\omega_0 = \frac{1}{RC}.$$



A simple transfer function: magnitude

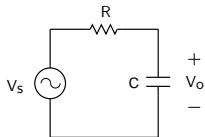


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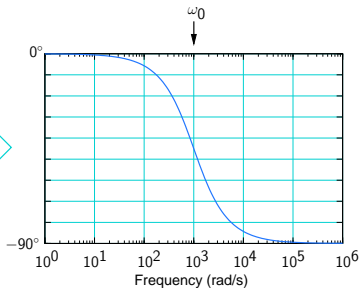
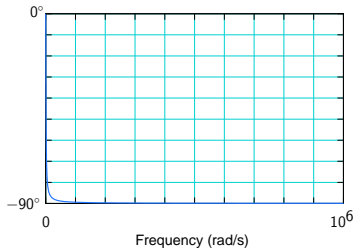


Since $\omega = 2\pi f$, the *shape* of the plot does not change.

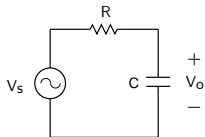
A simple transfer function: phase



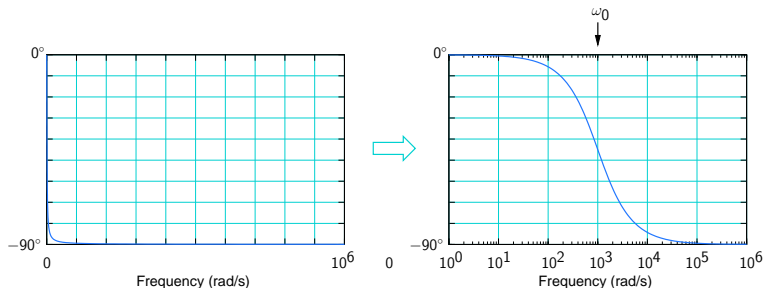
$$V_o = \frac{(1/sC)}{R + (1/sC)} V_s,$$
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A simple transfer function: phase

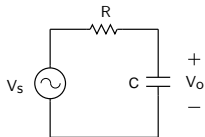


$$V_o = \frac{(1/sC)}{R + (1/sC)} V_s,$$
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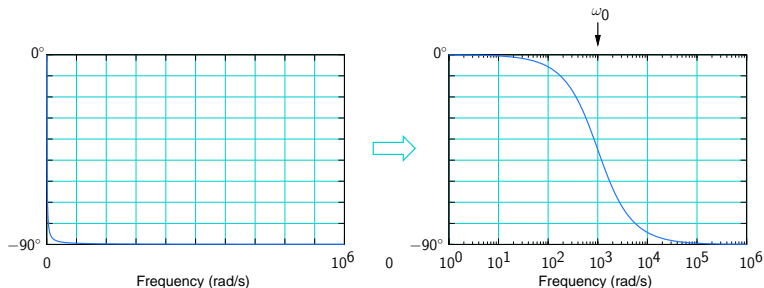


- * Since $\angle H = -\tan^{-1}(\omega/\omega_0)$ varies in a limited range (0° to -90°), a linear axis is appropriate for $\angle H$.

A simple transfer function: phase



$$V_o = \frac{(1/sC)}{R + (1/sC)} V_s,$$
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- * Since $\angle H = -\tan^{-1}(\omega/\omega_0)$ varies in a limited range (0° to -90°), a linear axis is appropriate for $\angle H$.
- * As in the magnitude plot, we use a log axis for ω , since we are interested in a wide range of ω .

$$\text{Consider } H(s) = \frac{K(1 + s/z_1)(1 + s/z_2) \cdots (1 + s/z_M)}{(1 + s/p_1)(1 + s/p_2) \cdots (1 + s/p_N)} .$$

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$-z_1, -z_2, \dots$ are called the “zeros” of $H(s)$.

$-p_1, -p_2, \dots$ are called the “poles” of $H(s)$.

(In addition, there could be terms like s, s^2, \dots in the numerator.)

We will assume, for simplicity, that the zeros (and poles) are real and distinct.

Construction of Bode plots involves

$$\text{Consider } H(s) = \frac{K(1 + s/z_1)(1 + s/z_2) \cdots (1 + s/z_M)}{(1 + s/p_1)(1 + s/p_2) \cdots (1 + s/p_N)}.$$

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Construction of Bode plots involves

- (a) computing approximate contribution of each pole/zero as a function ω .

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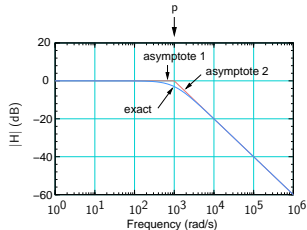
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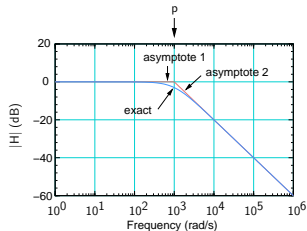
- computing approximate contribution of each pole/zero as a function ω .
- combining the various contributions to obtain $|H|$ and $\angle H$ versus ω .

Contribution of a pole: magnitude



$$\text{Consider } H(s) = \frac{1}{1 + s/p} \rightarrow H(j\omega) = \frac{1}{1 + j(\omega/p)}, |H(j\omega)| = \frac{1}{\sqrt{1 + (\omega/p)^2}}.$$

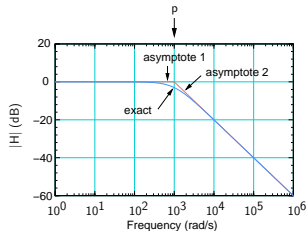
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Contribution of a pole: magnitude

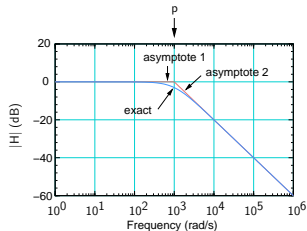


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Asymptote 2: $\omega \gg p$: $|H| \rightarrow \frac{1}{\omega/p} = \frac{p}{\omega} \rightarrow |H| = 20 \log p - 20 \log \omega$ (dB)

Contribution of a pole: magnitude



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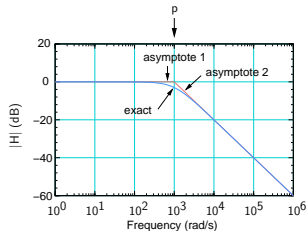
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Consider two values of ω : ω_1 and $10\omega_1$.

$$|H|_1 = 20 \log p - 20 \log \omega_1 \text{ (dB)}$$

$$|H|_2 = 20 \log p - 20 \log (10\omega_1) \text{ (dB)}$$

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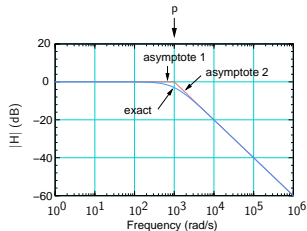
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$$|H|_1 - |H|_2 = -20 \log \frac{\omega_1}{10\omega_1} = 20 \text{ dB.}$$

Contribution of a pole: magnitude



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Consider two values of ω : ω_1 and $10\omega_1$.

$$|H|_1 = 20 \log p - 20 \log \omega_1 \text{ (dB)}$$

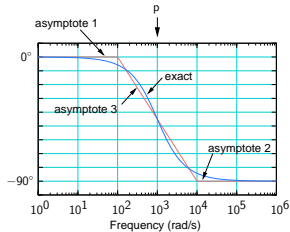
$$|H|_2 = 20 \log p - 20 \log (10\omega_1) \text{ (dB)}$$

$$|H|_1 - |H|_2 = -20 \log \frac{\omega_1}{10\omega_1} = 20 \text{ dB.}$$

$\rightarrow |H|$ versus ω has a slope of -20 dB/decade.

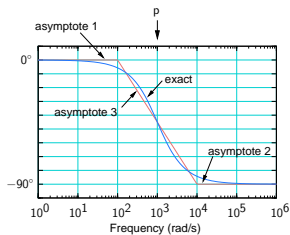
Note that, at $\omega = p$, the actual value of $|H|$ is $1/\sqrt{2}$ (i.e., -3 dB).

Contribution of a pole: phase



$$\text{Consider } H(s) = \frac{1}{1 + s/p} = \frac{1}{1 + j(\omega/p)} \rightarrow \angle H = -\tan^{-1}\left(\frac{\omega}{p}\right)$$

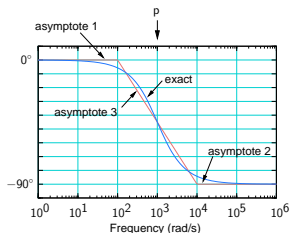
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Contribution of a pole: phase

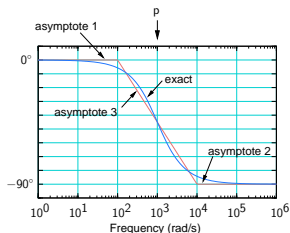


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Contribution of a pole: phase



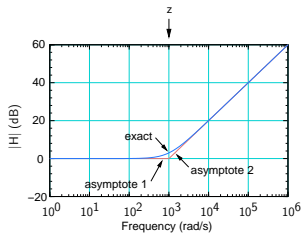
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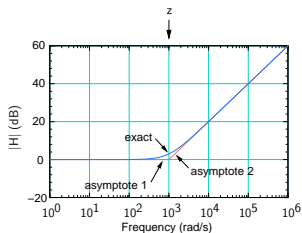
Asymptote 3: For $p/10 < \omega < 10p$, $\angle H$ is assumed to vary linearly with $\log \omega$
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Contribution of a zero: magnitude



Consider $H(s) = 1 + s/z \rightarrow H(j\omega) = 1 + j(\omega/z)$, $|H(j\omega)| = \sqrt{1 + (\omega/z)^2}$.

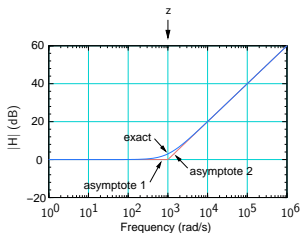
Contribution of a zero: magnitude



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Asymptote 1: $\omega \ll z$: $|H| \rightarrow 1$, $20 \log |H| = 0$ dB.

Contribution of a zero: magnitude

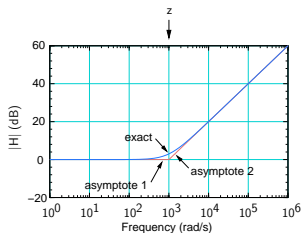


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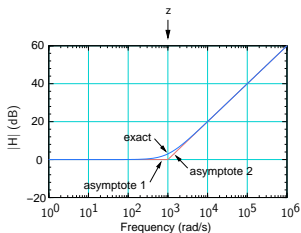
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Consider two values of ω : ω_1 and $10\omega_1$.

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$$|H|_2 = 20 \log (10\omega_1) - 20 \log z \text{ (dB)}$$

Contribution of a zero: magnitude



Consider $H(s) = 1 + s/z \rightarrow H(j\omega) = 1 + j(\omega/z)$, $|H(j\omega)| = \sqrt{1 + (\omega/z)^2}$.

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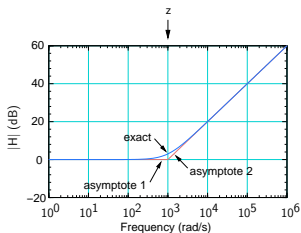
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Asymptote 1: $\omega \ll z$: $|H| \rightarrow 1$, $20 \log |H| = 0$ dB.

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Consider two values of ω : ω_1 and $10\omega_1$.

$$|H|_1 = 20 \log \omega_1 - 20 \log z \text{ (dB)}$$

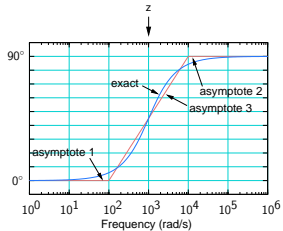
$$|H|_2 = 20 \log (10\omega_1) - 20 \log z \text{ (dB)}$$

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$\rightarrow |H|$ versus ω has a slope of +20 dB/decade.

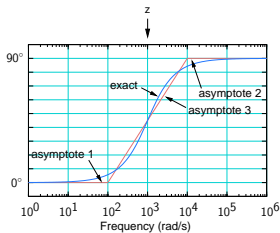
Note that, at $\omega = z$, the actual value of $|H|$ is $\sqrt{2}$ (i.e., 3 dB).

Contribution of a zero: phase



$$\text{Consider } H(s) = 1 + s/z = 1 + j(\omega/z) \rightarrow \angle H = \tan^{-1} \left(\frac{\omega}{z} \right)$$

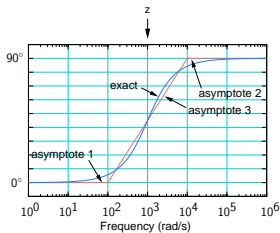
Contribution of a zero: phase



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Asymptote 1: $\omega \ll z$ (say, $\omega < z/10$): $\angle H = 0$.

Contribution of a zero: phase

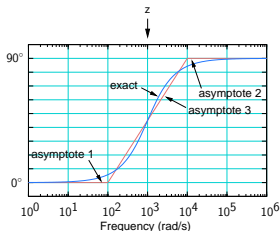


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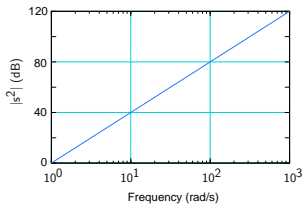
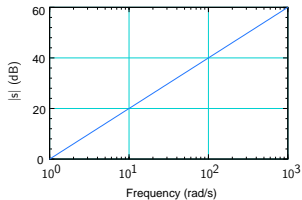
Asymptote 3: For $z/10 < \omega < 10z$, $\angle H$ is assumed to vary linearly with $\log \omega$
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Contribution of K (constant), s , and s^2

For $H(s) = K$, $20 \log |H| = 20 \log K$ (a constant), and $\angle H = 0$.

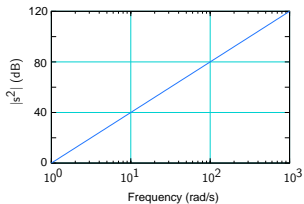
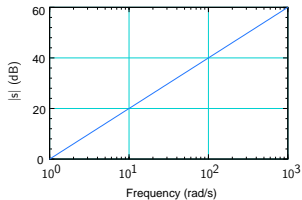
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Contribution of K (constant), s , and s^2

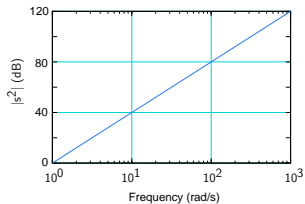
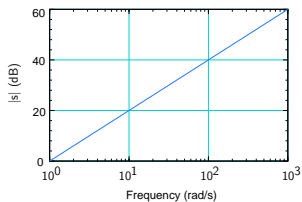
For $H(s) = K$, $20 \log |H| = 20 \log K$ (a constant), and $\angle H = 0$.



For $H(s) = s$, i.e., $H(j\omega) = j\omega$, $|H| = \omega$.

Contribution of K (constant), s , and s^2

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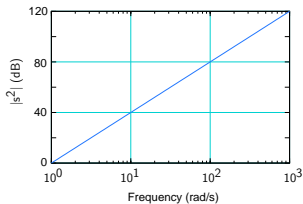
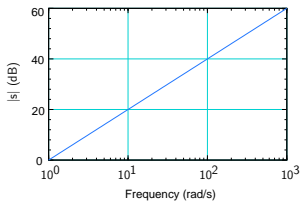


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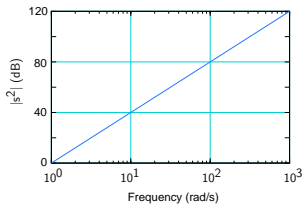
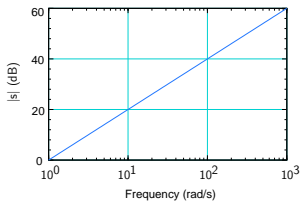
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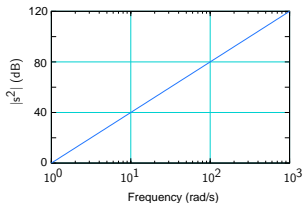
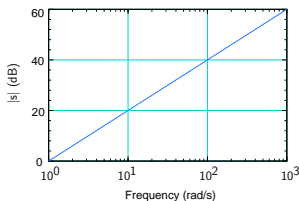
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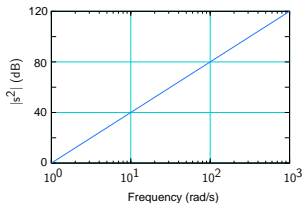
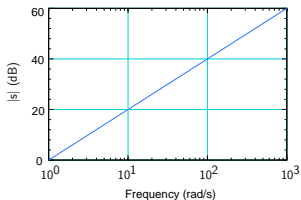
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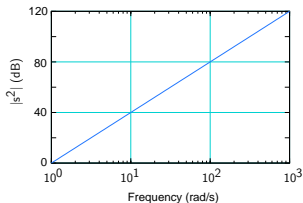
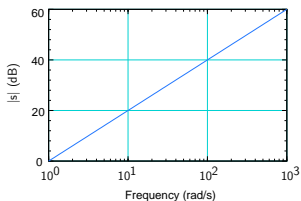
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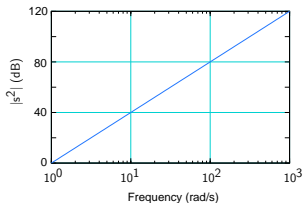
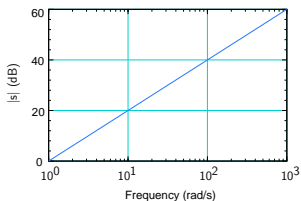
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The same reasoning applies to more than two terms as well.

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Consider $H(s) = \frac{10s}{(1 + s/10^2)(1 + s/10^5)}$.

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Let $H(s) = H_1(s) H_2(s) H_3(s) H_4(s)$, where

$$H_1(s) = 10,$$

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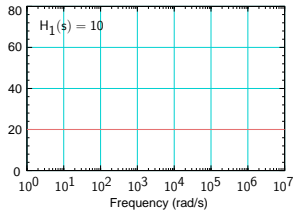
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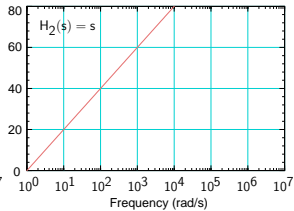
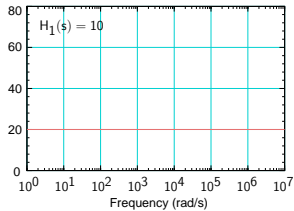
$$H_4(s) = \frac{1}{1 + s/p_2}, p_2 = 10^5 \text{ rad/s}.$$

We can now plot the magnitude and phase of H_1 , H_2 , H_3 , H_4 *individually* versus ω and then simply add them to obtain $|H|$ and $\angle H$.

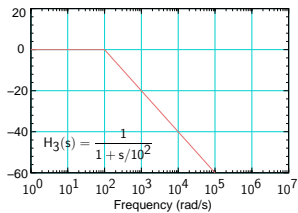
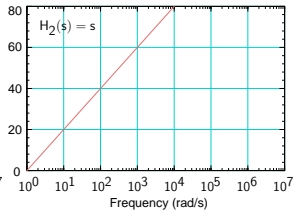
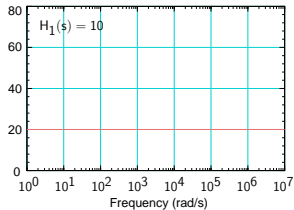
Magnitude plot ($|H|$ in dB)



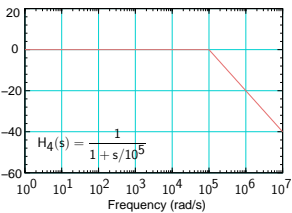
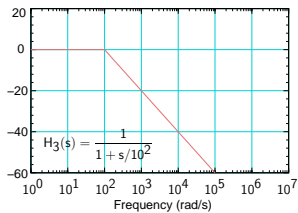
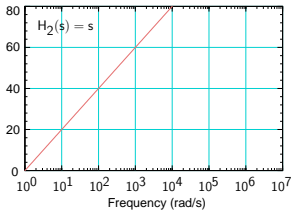
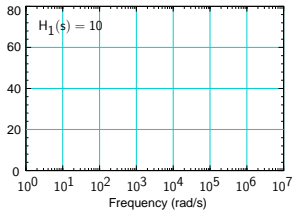
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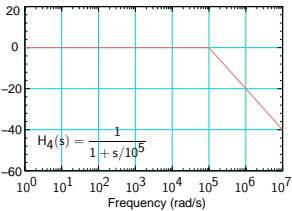
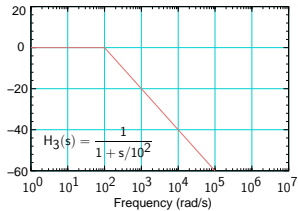
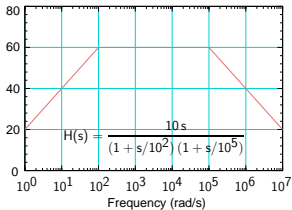
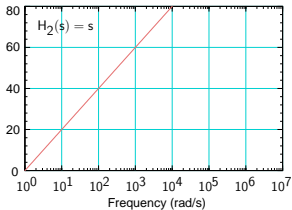
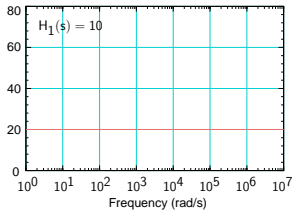
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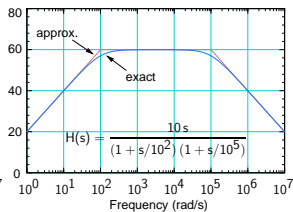
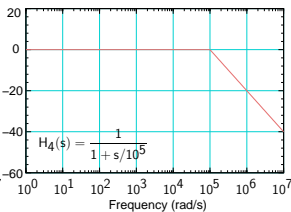
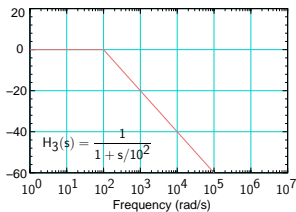
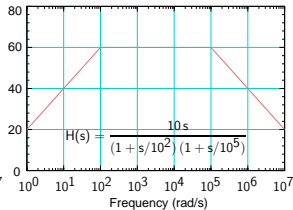
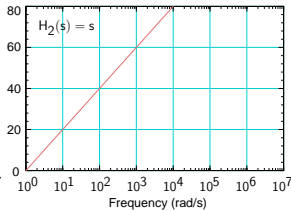
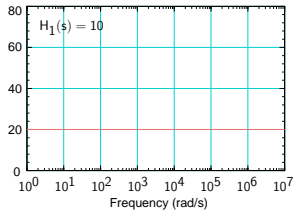
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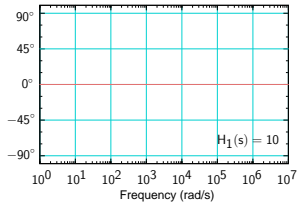
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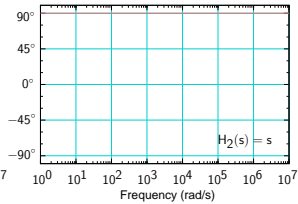
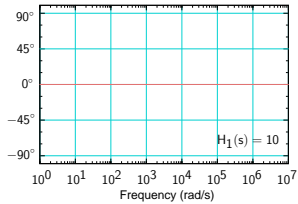
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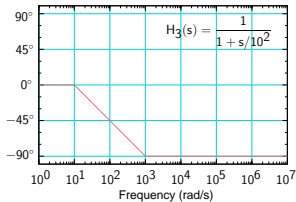
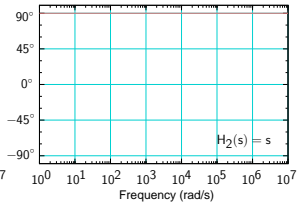
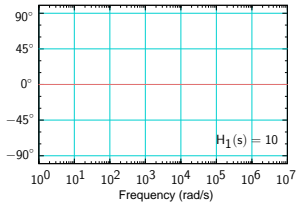
Phase plot



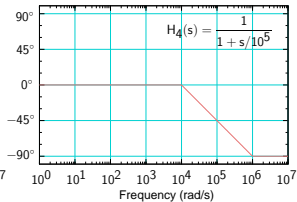
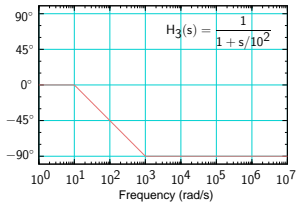
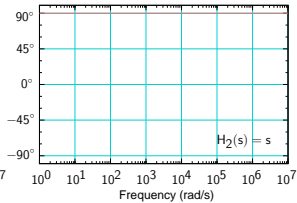
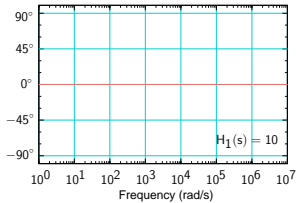
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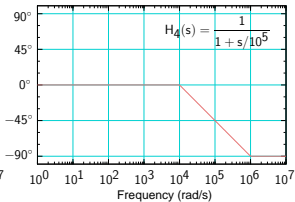
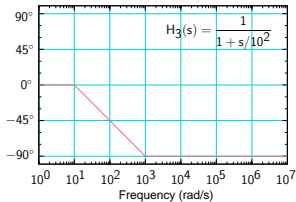
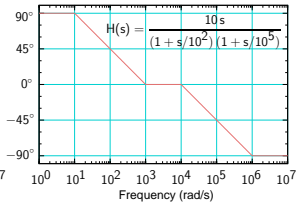
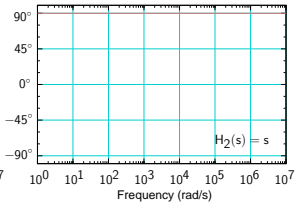
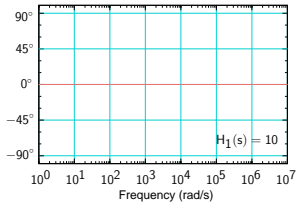
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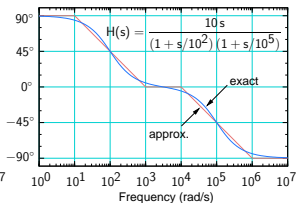
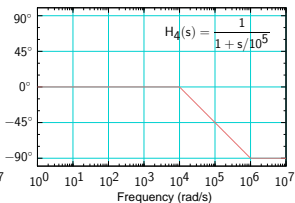
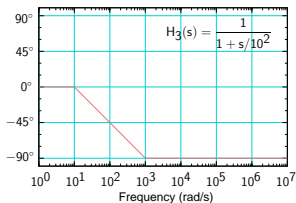
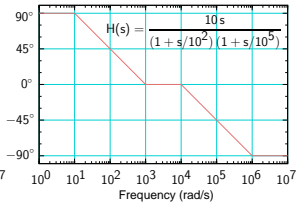
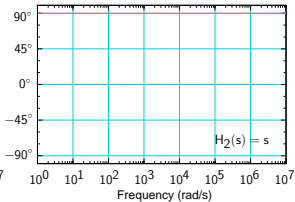
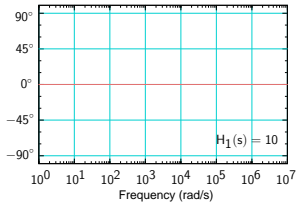
Phase plot



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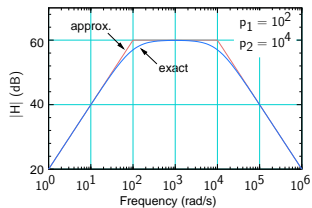
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- * When the poles and zeros are not sufficiently separated, the Bode approximation should be used only for a rough estimate, followed by a numerical calculation. However, even in such cases, it does give a good idea of the *asymptotic* magnitude and phase plots, which is valuable in amplifier design.

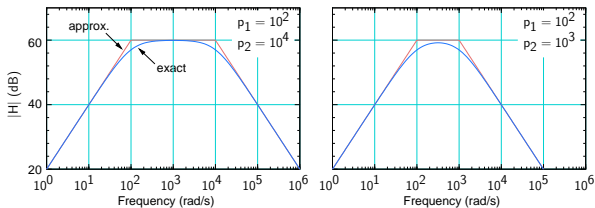
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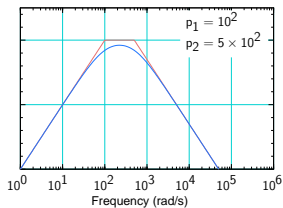
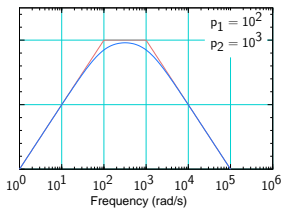
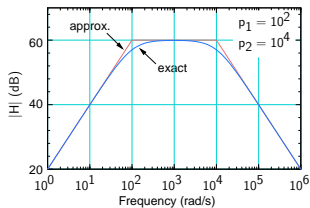
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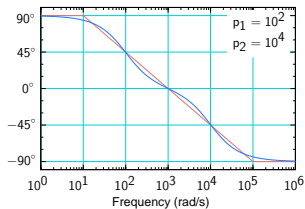
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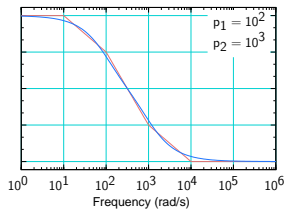
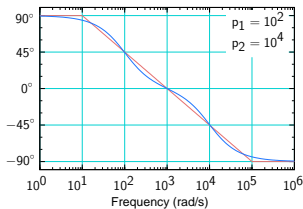
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