# Performance of ML Receiver for *M*-ary Signaling

Saravanan Vijayakumaran sarva@ee.iitb.ac.in

Department of Electrical Engineering Indian Institute of Technology Bombay

October 8, 2013

# Performance of ML Decision Rule for *M*-ary signaling

#### ML Decision Rule for *M*-ary Signaling

• *M* equally likely hypotheses

$$\begin{array}{rcl} H_1 & : & y(t) = s_1(t) + n(t) \\ H_2 & : & y(t) = s_2(t) + n(t) \\ \vdots & & \vdots \\ H_M & : & y(t) = s_M(t) + n(t) \end{array}$$

• The ML decision rule for real AWGN channel is

$$\delta_{ML}(y) = \underset{1 \le i \le M}{\operatorname{argmin}} \|y - s_i\|^2 = \underset{1 \le i \le M}{\operatorname{argmax}} \langle y, s_i \rangle - \frac{\|s_i\|^2}{2}$$

The ML decision rule for complex AWGN channel is

$$\delta_{ML}(y) = \underset{1 \le i \le M}{\operatorname{argmin}} \|y - s_i\|^2 = \underset{1 \le i \le M}{\operatorname{argmax}} \operatorname{Re}\left(\langle y, s_i \rangle\right) - \frac{\|s_i\|^2}{2}$$

• In general, there is no neat expression for Pe as in the binary case

#### **QPSK**

• QPSK signals where *p*(*t*) is a real baseband pulse of duration *T* 

$$\begin{split} s_{1}^{p}(t) &= \sqrt{2}p(t)\cos\left(2\pi f_{c}t + \frac{\pi}{4}\right) \\ s_{2}^{p}(t) &= \sqrt{2}p(t)\cos\left(2\pi f_{c}t + \frac{3\pi}{4}\right) \\ s_{3}^{p}(t) &= \sqrt{2}p(t)\cos\left(2\pi f_{c}t + \frac{5\pi}{4}\right) \\ s_{4}^{p}(t) &= \sqrt{2}p(t)\cos\left(2\pi f_{c}t + \frac{7\pi}{4}\right) \end{split}$$

Complex envelopes of QPSK Signals

$$s_1(t) = p(t)e^{j\frac{\pi}{4}}, s_2(t) = p(t)e^{j\frac{3\pi}{4}}, s_3(t) = p(t)e^{j\frac{5\pi}{4}}, s_4(t) = p(t)e^{j\frac{7\pi}{4}}$$

Orthonormal basis for the complex envelopes consists of only

$$\phi(t) = \frac{\rho(t)}{\sqrt{E_{\rho}}}$$

## ML Receiver for QPSK

- $E_b = E_p/2$
- The vector representation of the QPSK signals is

$$\begin{aligned} s_1 &= \sqrt{E_b} + j\sqrt{E_b} \\ s_2 &= -\sqrt{E_b} + j\sqrt{E_b} \\ s_3 &= -\sqrt{E_b} - j\sqrt{E_b} \\ s_4 &= \sqrt{E_b} - j\sqrt{E_b} \end{aligned}$$

The hypothesis testing problem in terms of vectors is

$$H_i: Y = s_i + N, i = 1, ..., 4$$

where  $N \sim \mathcal{CN}(0, 2\sigma^2)$ 

• The ML decision rule is given by

$$\delta_{ML}(y) = \underset{1 \leq i \leq 4}{\operatorname{argmin}} \|y - s_i\|^2 = \underset{1 \leq i \leq 4}{\operatorname{argmax}} \operatorname{Re}\left(\langle y, s_i \rangle\right) - \frac{\|s_i\|^2}{2}$$

The ML decision rule decides s<sub>i</sub> was transmitted if y belongs to the *i*th quadrant

#### ML Decision Rule for QPSK



$$P_{e|1} = \Pr\left[Y_c < 0 \text{ or } Y_s < 0 \middle| (\sqrt{E_b}, \sqrt{E_b}) \text{ was sent} \right]$$

#### ML Decision Rule for QPSK

• Probability of error when s<sub>1</sub> is transmitted is

$$P_{e|1} = \Pr\left[Y_c < 0 \text{ or } Y_s < 0 \middle| (\sqrt{E_b}, \sqrt{E_b}) \text{ was sent}\right]$$
$$= 2Q\left(\sqrt{\frac{2E_b}{N_0}}\right) - Q^2\left(\sqrt{\frac{2E_b}{N_0}}\right)$$

• By symmetry,

$$P_{e|1} = P_{e|2} = P_{e|3} = P_{e|4}$$

• The average probability of error is

$$P_{e} = \frac{1}{4} \sum_{i=1}^{4} P_{e|i} = P_{e|1} = 2Q\left(\sqrt{\frac{2E_{b}}{N_{0}}}\right) - Q^{2}\left(\sqrt{\frac{2E_{b}}{N_{0}}}\right)$$

## ML Decision Rule for 16-QAM

16-QAM



Exact analysis is tedious. Approximate analysis is sufficient.

## Revisiting the Q function

## Revisiting the *Q* function

 $X \sim N(0, 1)$ 

$$Q(x) = P[X > x] = \int_x^\infty \frac{1}{\sqrt{2\pi}} \exp\left(\frac{-t^2}{2}\right) dt$$



#### Bounds on Q(x) for Large Arguments



 $\left(1 - \frac{1}{x^2}\right) \frac{e^{-\frac{x^2}{2}}}{x\sqrt{2\pi}} \le Q(x) \le \frac{e^{-\frac{x^2}{2}}}{x\sqrt{2\pi}}$ (1)

#### Q Functions with Smallest Arguments Dominate



- P<sub>e</sub> in AWGN channels can be bounded by a sum of Q functions
- The Q function with the smallest argument is used to approximate P<sub>e</sub>

## Union Bound

### Union Bound for *M*-ary Signaling in AWGN

- Let  $Z_i$  be  $\langle y, s_i \rangle \frac{\|s_i\|^2}{2}$  or Re  $(\langle y, s_i \rangle) \frac{\|s_i\|^2}{2}$
- The conditional error probability given *H<sub>i</sub>* is true is

$$P_{e|i} = \Pr\left[\bigcup_{j \neq i} \left\{Z_i < Z_j\right\} \middle| H_i\right]$$

• Since  $P(A \cup B) \leq P(A) + P(B)$ , we have

$$P_{e|i} \leq \sum_{j \neq i} \Pr\left[Z_i < Z_j \middle| H_i
ight] = \sum_{j \neq i} Q\left(rac{\|s_j - s_i\|}{2\sigma}
ight)$$

The error probability is given by

$$P_{e} = \frac{1}{M} \sum_{i=1}^{M} P_{e|i} \leq \frac{1}{M} \sum_{i=1}^{M} \sum_{j \neq i} Q\left(\frac{\|\boldsymbol{s}_{j} - \boldsymbol{s}_{i}\|}{2\sigma}\right)$$

#### Union Bound for QPSK



$$P_{e|1} = \Pr\left[\bigcup_{j \neq 1} \{Z_1 < Z_j\} \middle| H_1\right] \le \sum_{j \neq 1} \Pr\left[Z_1 < Z_j \middle| H_1\right]$$
$$P_{e|1} \le Q\left(\frac{||s_2 - s_1||}{2\sigma}\right) + Q\left(\frac{||s_3 - s_1||}{2\sigma}\right) + Q\left(\frac{||s_4 - s_1||}{2\sigma}\right)$$
$$= 2Q\left(\sqrt{\frac{2E_b}{N_0}}\right) + Q\left(\sqrt{\frac{4E_b}{N_0}}\right)$$

## Union Bound for QPSK

• Union bound on error probability of ML rule

$$P_{e} \leq 2Q\left(\sqrt{\frac{2E_{b}}{N_{0}}}
ight) + Q\left(\sqrt{\frac{4E_{b}}{N_{0}}}
ight)$$

• Exact error probability of ML rule

$$P_{e} = 2Q\left(\sqrt{\frac{2E_{b}}{N_{0}}}\right) - Q^{2}\left(\sqrt{\frac{2E_{b}}{N_{0}}}\right)$$

#### Union Bound and Exact Error Probability for QPSK



## Intelligent Union Bound

#### **QPSK Error Events**

 $E_1 = [Z_2 > Z_1] \cup [Z_3 > Z_1] \cup [Z_4 > Z_1] = [Z_2 > Z_1] \cup [Z_4 > Z_1]$ 



#### Intelligent Union Bound for QPSK

• Intelligent union bound on P<sub>e|1</sub>

$$P_{e|1} = \Pr\left[ (Z_2 > Z_1) \cup (Z_4 > Z_1) \middle| H_1 \right]$$
  
$$\leq \Pr\left[ Z_2 > Z_1 \middle| H_1 \right] + \Pr\left[ Z_4 > Z_1 \middle| H_1 \right]$$
  
$$= Q\left( \frac{\|S_2 - S_1\|}{2\sigma} \right) + Q\left( \frac{\|S_4 - S_1\|}{2\sigma} \right)$$
  
$$= 2Q\left( \sqrt{\frac{2E_b}{N_0}} \right)$$

• By symmetry  $P_{e|1} = P_{e|2} = P_{e|3} = P_{e|4}$  and

$$P_e \leq 2Q\left(\sqrt{rac{2E_b}{N_0}}
ight)$$

## Intelligent Union Bound and Exact Error Probability for QPSK



#### General Strategy for Intelligent Union Bound

 Let N<sub>ML</sub>(i) be the smallest set of neighbors of s<sub>i</sub> which define the decision region Γ<sub>i</sub>

$$\Gamma_i = \left\{ y \middle| \delta_{ML}(y) = i \right\} = \left\{ y \middle| Z_i \ge Z_j \text{ for all } j \in N_{ML}(i) \right\}$$

• Probability of error when *s<sub>i</sub>* is transmitted is

$$\begin{aligned} \mathcal{P}_{e|i} &= \Pr\left[y \notin \Gamma_i | H_i\right] = \Pr\left[Z_i < Z_j \text{ for some } j \in N_{ML}(i) \middle| H_i\right] \\ &\leq \sum_{j \in N_{ML}(i)} Q\left(\frac{\|s_j - s_i\|}{2\sigma}\right) \end{aligned}$$

Average probability of error is bounded by

$$P_e \leq rac{1}{M} \sum_{i=1}^{M} \sum_{j \in N_{ML}(i)} Q\left(rac{\|m{s}_j - m{s}_i\|}{2\sigma}
ight)$$

#### Intelligent Union Bound for 16-QAM



## Nearest Neighbors Approximation

#### Nearest Neighbors Approximation

Let d<sub>min</sub> be the minimum distance between constellation points

$$d_{min} = \min_{i \neq j} \|s_i - s_j\|$$

• Let  $N_{d_{min}}(i)$  denote the number of nearest neighbors of  $s_i$ 

$$P_{e|i} pprox N_{d_{min}}(i) Q\left(rac{d_{min}}{2\sigma}
ight)$$

Averaging over i we get

$$P_e pprox ar{N}_{d_{min}} Q\left(rac{d_{min}}{2\sigma}
ight)$$

where  $\bar{N}_{d_{min}}$  denotes the average number of nearest neighbors

#### Nearest Neighbors Approximation for QPSK



## Summary of results for QPSK

• Exact error probability of ML rule

$$P_e = 2Q\left(\sqrt{rac{2E_b}{N_0}}
ight) - Q^2\left(\sqrt{rac{2E_b}{N_0}}
ight)$$

• Union bound on error probability of ML rule

$$P_{e} \leq 2Q\left(\sqrt{\frac{2E_{b}}{N_{0}}}
ight) + Q\left(\sqrt{\frac{4E_{b}}{N_{0}}}
ight)$$

Intelligent union bound on error probability of ML rule

$$P_e \leq 2Q\left(\sqrt{rac{2E_b}{N_0}}
ight)$$

Nearest neighbors approximation of error probability of ML rule

$$P_e pprox 2Q\left(\sqrt{rac{2E_b}{N_0}}
ight)$$

#### Nearest Neighbors Approximation for 16-QAM



Thanks for your attention