# Symbol Detection in CDMA-OFDM Coexistence

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#### Problem Statement

- One user transmits the QPSK symbol b using CDMA (spreading code). Another user transmits a complex vector c using OFDM. Both use the same channel (carrier frequency).
- The transmitted OFDM signal

$$c(t) = \sum_{m=0}^{M-1} c_m e^{j2\pi mt/T} \qquad 0 \le t \le T$$
 (1)

where  $c_m$  are the complex data symbols being transmitted.

► The signal transmitted by the CDMA user

$$s(t) = \frac{b}{\sqrt{N}} \sum_{n=0}^{N-1} s_n p(t - nT_c) \qquad 0 \le t \le T$$
 (2)

where b is the transmitted symbol,  $s_n$  the spreading code,  $T_c$  the chip duration and p(t) the pulse shaping function.

OFDM demodulator output, assuming that the noise is AWGN

$$y = br + c + \sigma n$$

where **r** is the DFT of  $s_n$  and  $\mathbf{c} = \{c\}_{m=1}^{M-1}$ . The noise **n**, is complex circularly Gaussian with covariance matrix 2l.

We need to detect b and c.



## **Proposed Solution**

- ▶ Let the vector w be the linear filter used to demodulate the symbol b.
- ▶ The decision metric

$$D = \mathbf{w}^{H}\mathbf{y}$$

$$= b\mathbf{w}^{H}\mathbf{r} + \mathbf{w}^{H}\mathbf{c} + \sigma\mathbf{w}^{H}\mathbf{n}$$
 (3)

where  $\mathbf{w}^H$  is the conjugate transpose of the vector  $\mathbf{w}$ .

▶ An estimate of the symbol *b* is then obtained as

$$\hat{b} = sgn(Re\{D\}) + \iota sgn(Im\{D\})$$

- ► The challenge is to evolve an adaptive algorithm for updating the filter w.
- We would like to exploit the Gaussian nature of the additive noise to do this.

# The Adaptive Algorithm: MCPOE

The mean and variance of the real part of D, conditioned on b are

$$\mu_D = Re \left\{ \mathbf{w}^H (b\mathbf{r} + \mathbf{c}) \right\}$$

$$\sigma_D^2 = \sigma^2 \|\mathbf{w}\|^2$$
(4)

▶ If  $b^+$  and  $b^-$  are obtained from b by making its real part +1 and -1 respectively, the conditional probability of error is

$$P_{e|b} = \frac{1}{2}Q\left(\frac{\mu_{D+}}{\sigma_D}\right) + \frac{1}{2}Q\left(\frac{\mu_{D-}}{\sigma_D}\right)$$
 (5)

where  $\mu_{D+}$  and  $\mu_{D-}$  are obtained from  $\mu_D$  by substituting b by  $b^+$  and  $b^-$  respectively.

Using the gradient descent approach,

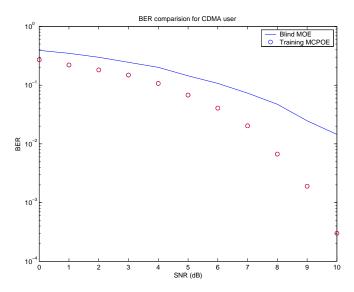
$$\mathbf{w}^{(i+1)} = \mathbf{w}^{(i)} - \lambda \nabla P_{e|b}^{(i)}, \tag{6}$$

where

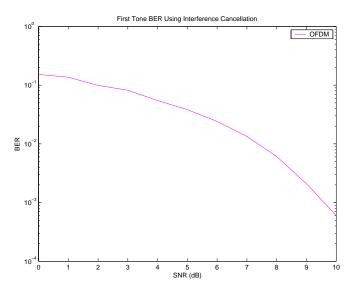
$$\nabla P_{e|b} = \frac{1}{\sqrt{2\pi}} \exp\left(-\frac{\mu_{D-}^2}{\sigma_D^2}\right) \frac{\|\mathbf{w}\|^2 \left(b^- \mathbf{r} + \mathbf{c}\right) - \mu_{D-} \mathbf{w}}{2\sigma \|\mathbf{w}\|^3}$$

$$- \frac{1}{\sqrt{2\pi}} \exp\left(-\frac{\mu_{D+}^2}{\sigma_D^2}\right) \frac{\|\mathbf{w}\|^2 \left(b^+ \mathbf{r} + \mathbf{c}\right) - \mu_{D+} \mathbf{w}}{2\sigma \|\mathbf{w}\|^3}$$
(7)

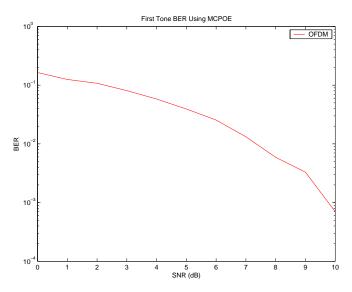
### Results



### Results Contd.



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### **Observations**

- Performance sensitive to choice of CDMA spreading code
- MCPOE with training performs better than blind MOE
- Using MCPOE to detect the OFDM user directly gives same performance as that obtained by detecting the CDMA user first and using interference cancellation
- ▶ BER performance for all tones more all less the same

#### References



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