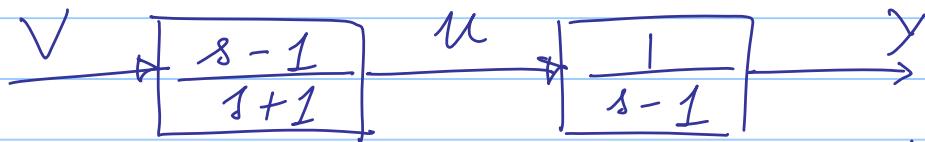


## 1) State Variables in Control

Review : Transfer function, Laplace Transform, Time response from Transfer function representations

Q We have learnt about transfer fn and their properties. Then,  
 \* Why do we need state variables?

Problem 1. Consider the following block diagram.

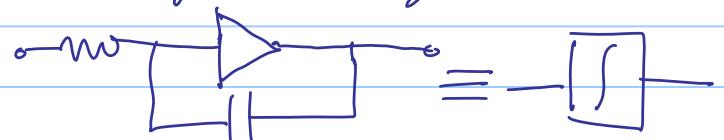


$$G_{\text{cl}}(s) = \frac{Y(s)}{V(s)} = \frac{s-1}{s+1} \cdot \frac{1}{s-1} = \frac{1}{s+1}$$

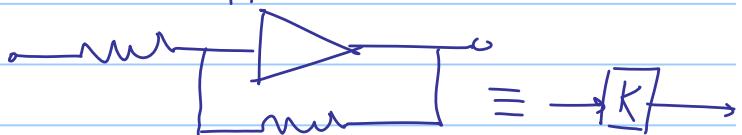
The overall system looks stable.  
 To verify let us try and simulate this system on an analog computer

Recall that an analog computer usually have the following components

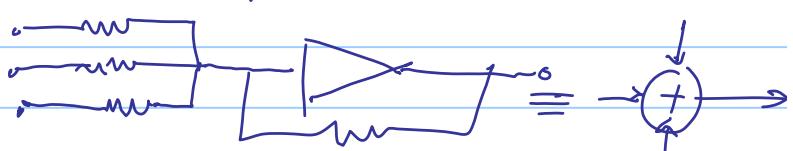
1) Integrator

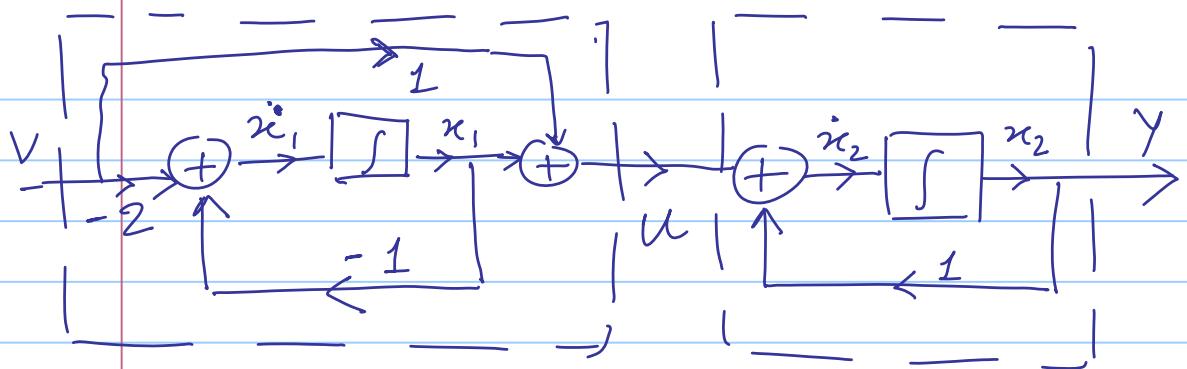


2) Amplifier



3) Adder





If you actually implement this figure on a analog computer in the lab, you will find that the output  $y$  (visible on a C.R.O e.g.) saturates or the system burns out.

→ The T.F.  $G_c(s)$  DID NOT say anything about this!

To find out why this happens look at the differential equation involved in the diagram above.

$$(1) \left\{ \begin{array}{l} \dot{x}_1 = -x_1 - 2v \\ \dot{x}_2 = x_1 + x_2 + v \end{array} \right. \quad \left| \begin{array}{l} x_1(0) = x_{10} \\ x_2(0) = x_{20} \end{array} \right.$$

For solving these equations we need initial conditions for  $x_1(0)$  &  $x_2(0)$ . In the analog computer simulation these are voltages at the output of the integrators when the simulation started.

Among various other methods, these equations can be solved using Laplace transforms:

Taking L.T. of ①,

$$sX_1(s) - X_1(0^-) = -X_1(s) - 2V(s)$$

$$sX_2(s) - X_2(0^-) = X_1(s) + X_2(s) + V(s)$$

$$\therefore \frac{Y(s)}{X_2(s)} = \frac{X_2(0)}{s-1} + \frac{X_1(0)}{(s-1)(s+1)} + \frac{V(s)}{s+1}$$
$$\Rightarrow y(t) = e^t x_{20} + \frac{1}{2}(e^t - e^{-t})x_{10} + e^{-t} * v(t)$$

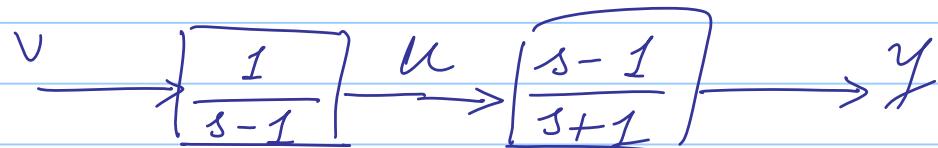
Hence the "actual" transfer function matches the "original"  $G_C(s)$ , only when  $x_{10} = x_{20} = 0$ .

For most other initial conditions,  $y(t)$  saturates in a short while and THAT IS WHAT IS VISIBLE ON THE ANALOG COMPUTER.

Exercise : For what other initial conditions in this problem the output does not saturate?

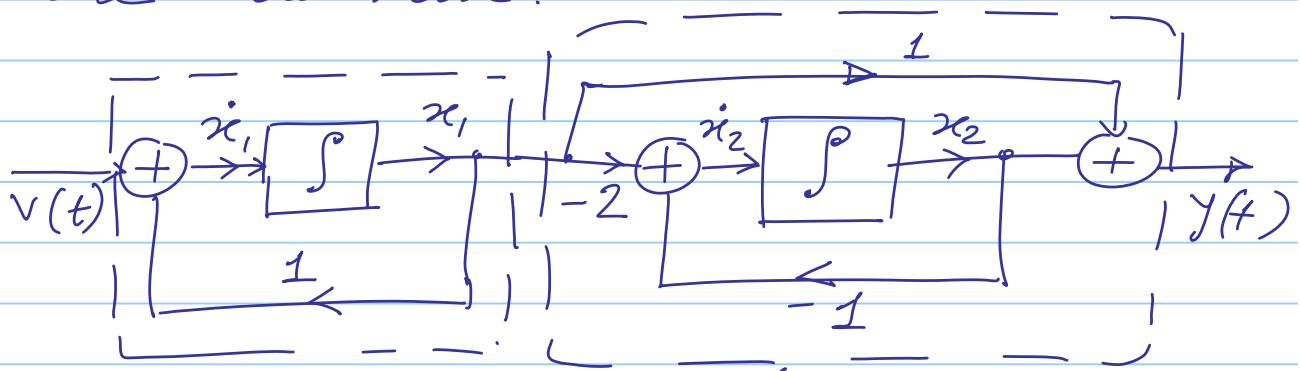
Exercise : Is this a good method for stabilizing  $G(s) = \frac{1}{s-1}$ ?

Problem 2 : Now consider the blocks in reverse order :



$$G_C'(s) = \frac{Y(s)}{V(s)} = \frac{1}{s-1} \cdot \frac{s-1}{s+1} = \frac{1}{s+1}$$

Hence, from a transfer function approach,  $G_C(s)$  and  $G_C'(s)$  are identical.



The corresponding differential eqns:

$$\left\{ \begin{array}{l} \dot{x}_1 = x_1 + V(t) \\ \dot{x}_2 = -2x_1 - x_2 \end{array} \right. \quad \left| \begin{array}{l} x_1(0) = x_{10} \\ x_2(0) = x_{20} \end{array} \right.$$

Solving these equations:

$$y(t) = (x_{10} + x_{20}) e^{-t} + e^{-t} * v(t)$$

In this case  $y(t)$  don't saturate even for non-zero  $x_{10}, x_{20}$ .

But consider,  $x_1(t)$

$$x_1(t) = e^{t x_{10}} + e^{t v(t)}$$

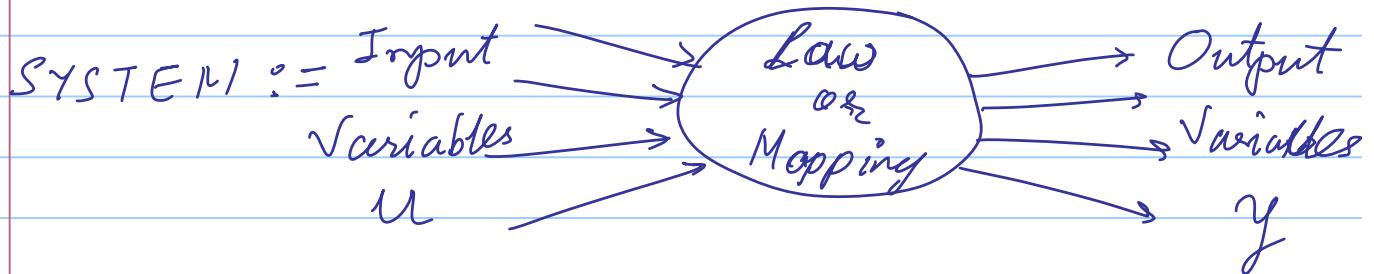
Hence  $u(t) = x_1(t)$  saturates ..

so, The analog computer simulation won't work here either.

SUMMARY: INTERNAL BEHAVIOR OF SYSTEMS ARE COMPLICATED.

T.F. cannot describe internal behavior. We require STATE VARIABLES.

Some Internal CONCEPTS

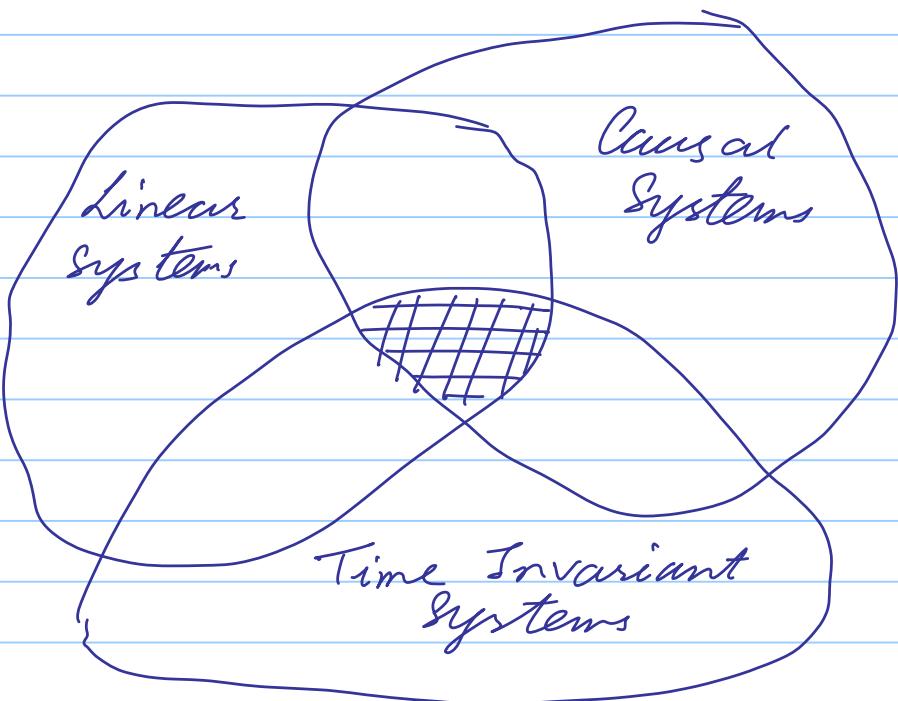


The law or mapping can take many forms e.g.:

- 1) A long list of Input and output.
- 1) Differential Equations  $\leftarrow$  ode
- 2) Transfer functions  $\leftarrow$  pdl
- 3) Impulse response
- 4) State Equations

Dynamical systems:  $u$  and  $y$  are functions of time ( $u(t)$  &  $y(t)$ ) or other independent variables.

## An Informal Classification Diagram



# EE 640 covers ONLY the shaded area

# In addition we assume that our systems can be described by Ordinary Differential Equations

Hence we deal exclusively with equations like :

$$y^{(n)}(t) + a_1 y^{(n-1)} + \dots + a_n y(t)$$

$$= b_0 u^{(m)}(t) + b_1 u^{(m-1)}(t) + \dots + b_m u(t)$$

↳ [Linear Time-Invariant Systems]

Exercise : Why do you think there equations are so important so as to form the basis of an entire course at IITB ?.

Exercise : Why are systems described by such equations LTI ?