Diode Circuits: Part 2

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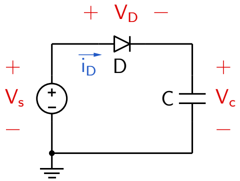
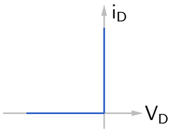
Indian Institute of Technology Bombay

Peak detector (with $V_{on} = 0\,\text{V}$)



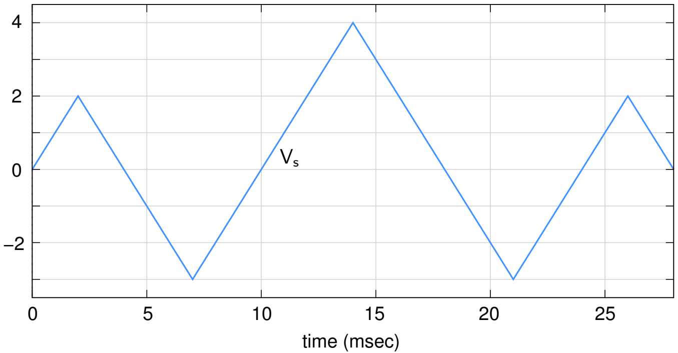


$V_c(0) = 0\,\text{V}$

$V_{on} = 0\,\text{V}$

$R_{on} \rightarrow 0\,\Omega$

$R_{off} \rightarrow \infty\,\Omega$



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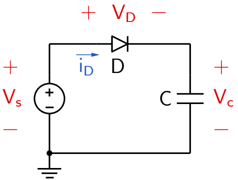

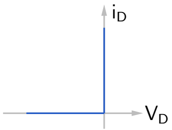

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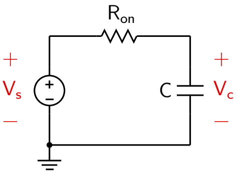
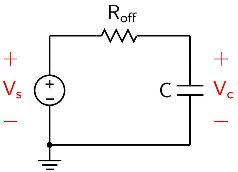


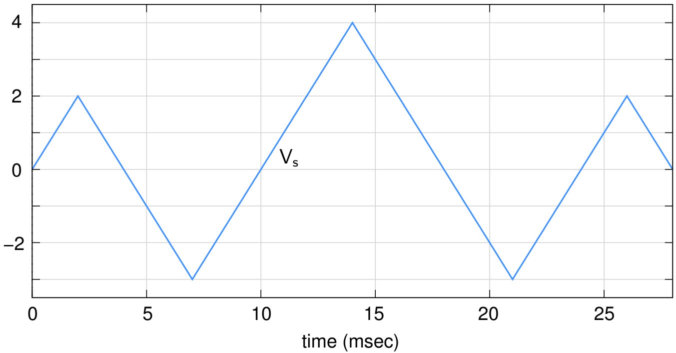
$V_s > V_c$

$\tau = R_{on} C$

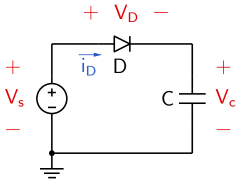


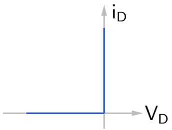
$V_s < V_c$

$\tau = R_{off} C$



Peak detector (with $V_{on} = 0\,V$)



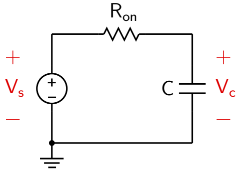


$V_c(0) = 0\,V$

$V_{on} = 0\,V$

$R_{on} \to 0\,\Omega$

$R_{off} \to \infty\,\Omega$



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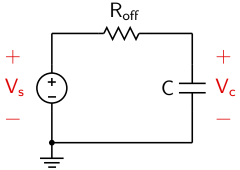
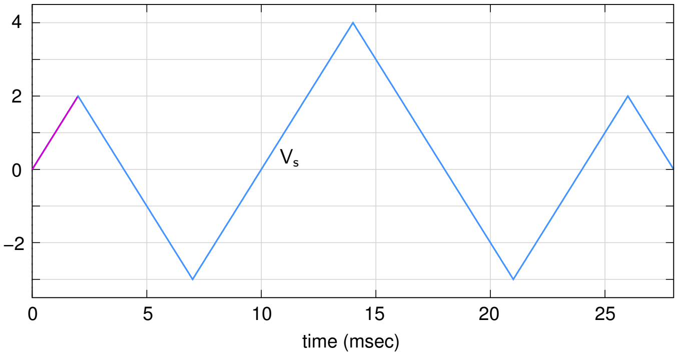


$V_s < V_c$

$\tau = R_{off} C$



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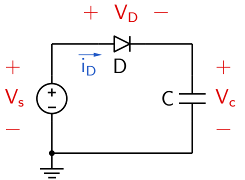

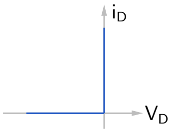

$V_c(0) = 0\,V$
$V_{on} = 0\,V$
$R_{on} \to 0\,\Omega$
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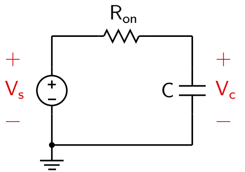

$V_s > V_c$
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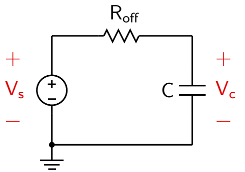
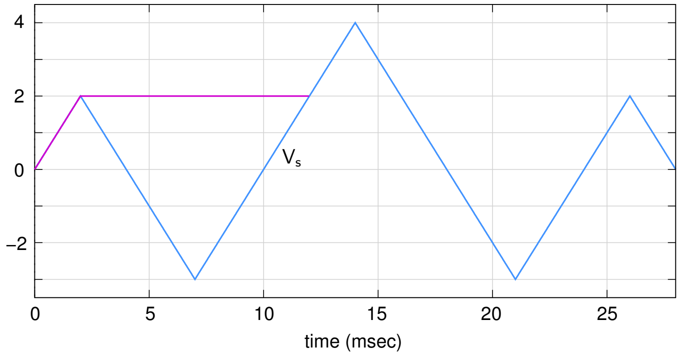


$V_s < V_c$
$\tau = R_{off} C$



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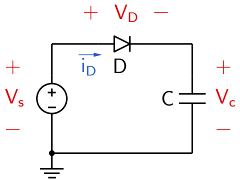

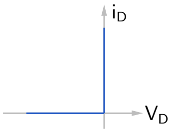

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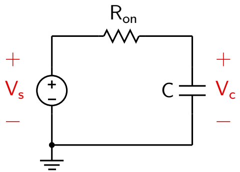

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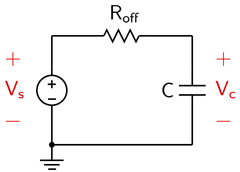
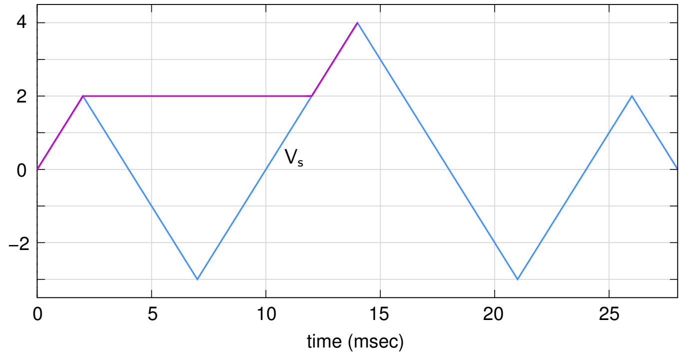
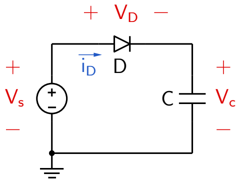


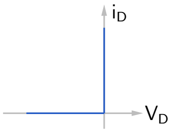
$V_s < V_c$

$\tau = R_{off} C$



Peak detector (with $V_{on} = 0\,\text{V}$)





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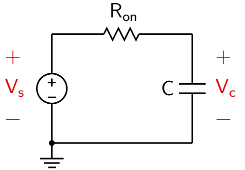

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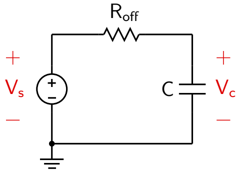
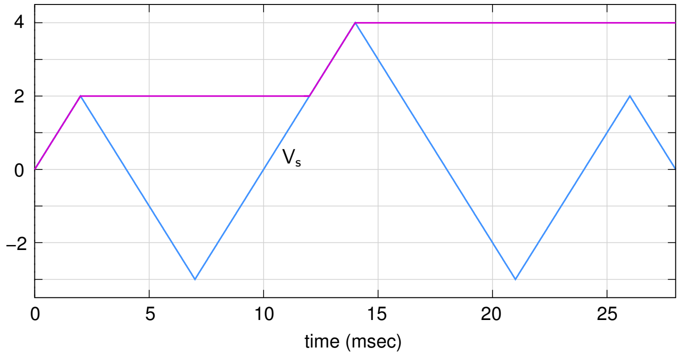
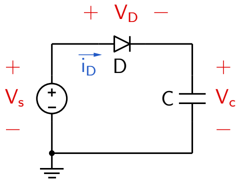


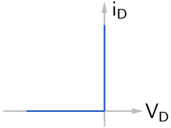
$V_s < V_c$

$\tau = R_{off} C$



Peak detector (with $V_{on} = 0\,V$)





$V_c(0) = 0\,V$

$V_{on} = 0\,V$

$R_{on} \rightarrow 0\,\Omega$

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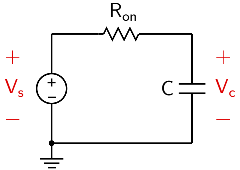

$V_s > V_c$

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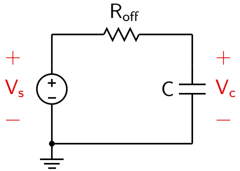

$V_s < V_c$

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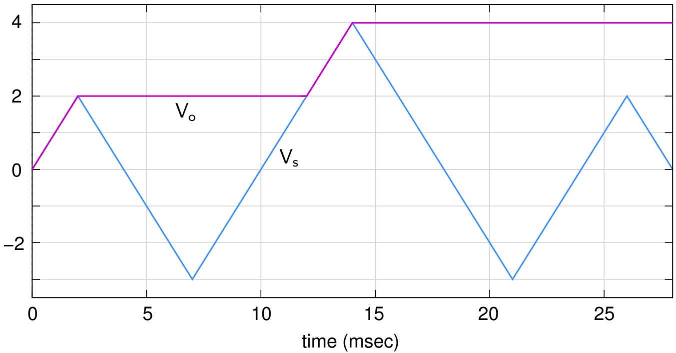

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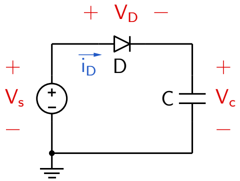
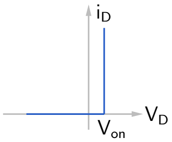
Peak detector (with $V_{on} = 0.7\,\text{V}$)



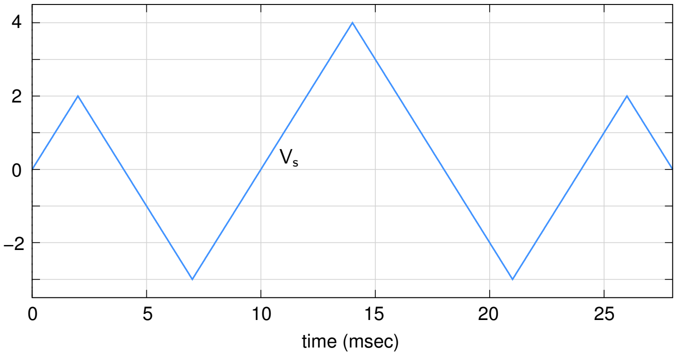


$V_c(0) = 0\,\text{V}$

$V_{on} = 0.7\,\text{V}$

$R_{on} \to 0\,\Omega$

$R_{off} \to \infty\,\Omega$



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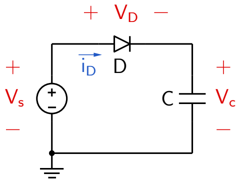

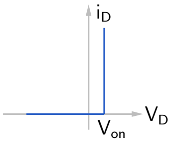

$V_c(0) = 0\,\mathrm{V}$

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$R_{on} \to 0\,\Omega$

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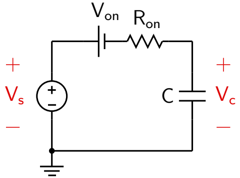
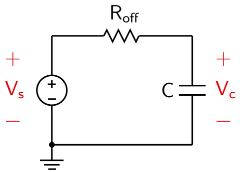


$V_s > V_c + V_{on}$

$\tau = R_{on}C$



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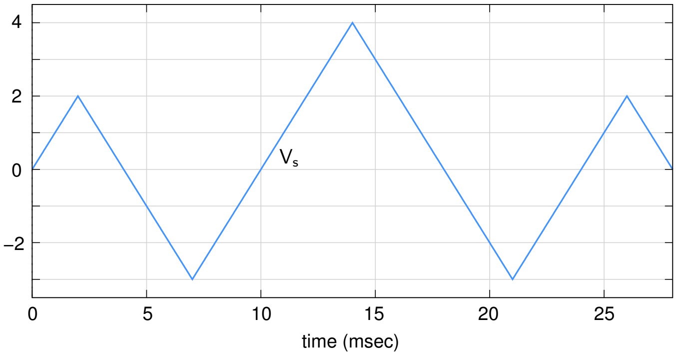

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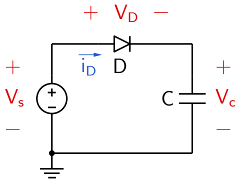

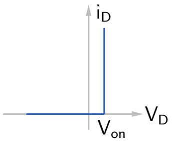

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$V_{on} = 0.7\,\mathrm{V}$

$R_{on} \to 0\,\Omega$

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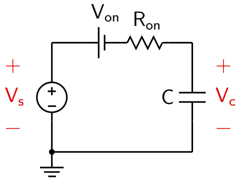

$V_s > V_c + V_{on}$

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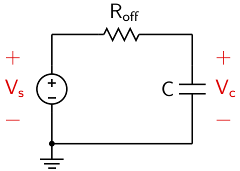

$V_s < V_c + V_{on}$

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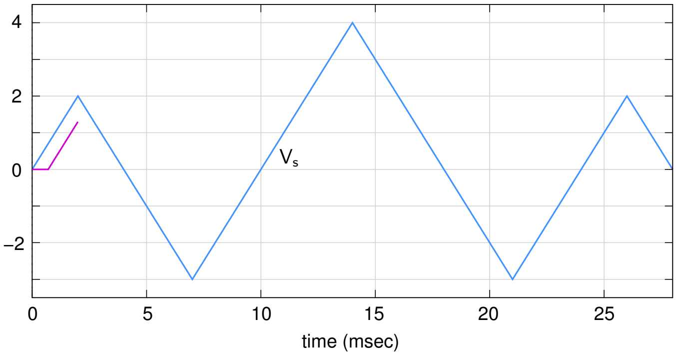
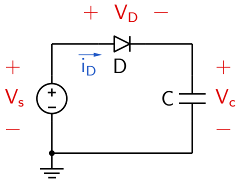
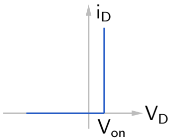


Peak detector (with $V_{on} = 0.7\,V$)





$V_c(0) = 0\,V$

$V_{on} = 0.7\,V$

$R_{on} \rightarrow 0\,\Omega$

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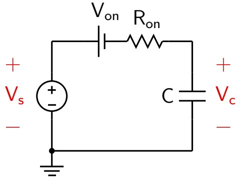

$V_s > V_c + V_{on}$

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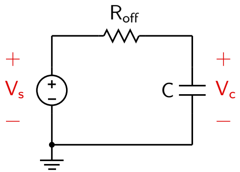

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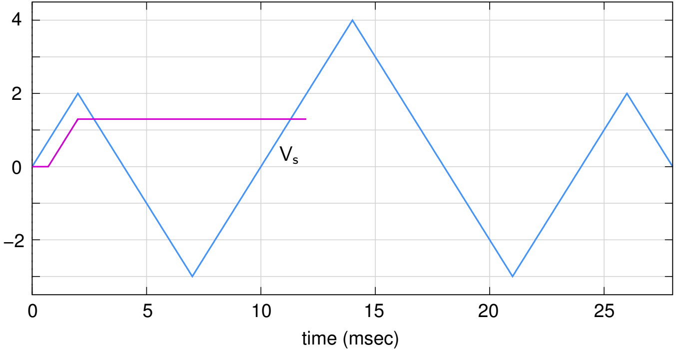
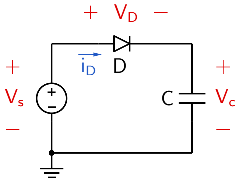
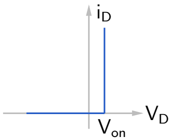


Peak detector (with $V_{on} = 0.7\,\mathrm{V}$)





$V_c(0) = 0\,\mathrm{V}$

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$R_{on} \to 0\,\Omega$

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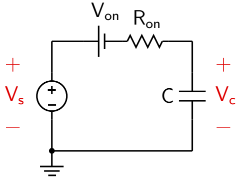

$V_s > V_c + V_{on}$

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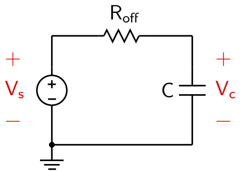

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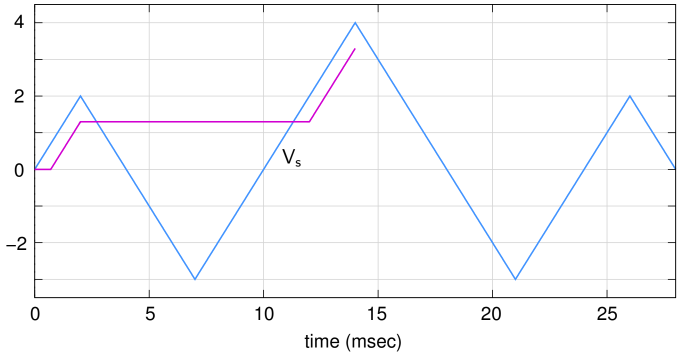

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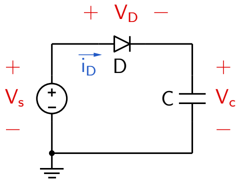

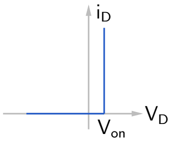

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$R_{on} \to 0\,\Omega$

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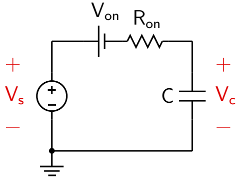

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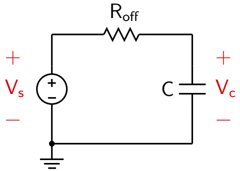

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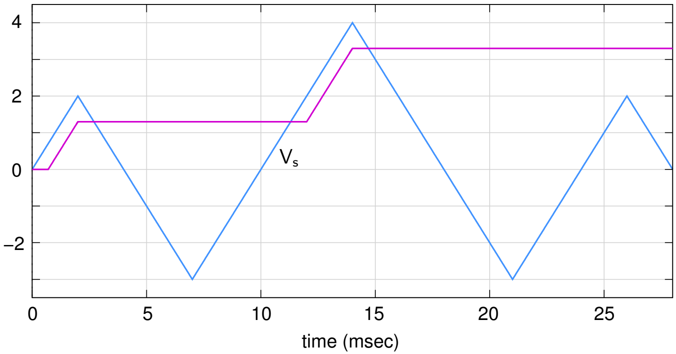

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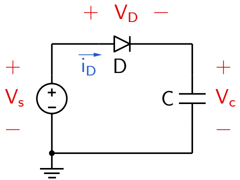
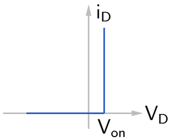


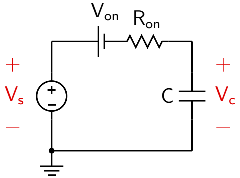


$V_c(0) = 0\,V$

$V_{on} = 0.7\,V$

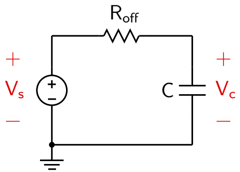
$R_{on} \to 0\,\Omega$

$R_{off} \to \infty\,\Omega$



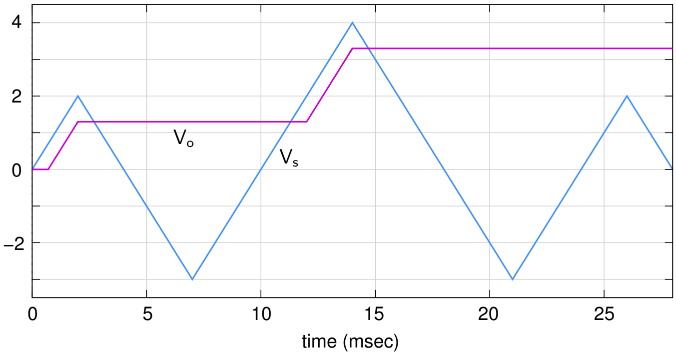
$V_s > V_c + V_{on}$

$\tau = R_{on}C$



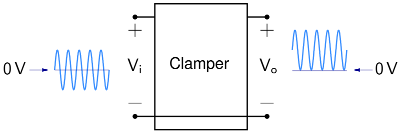
$V_s < V_c + V_{on}$

$\tau = R_{off}C$



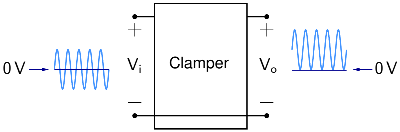
SEQUEL file: `ee101_peak_detector_1.sqproj`

Clamper circuits



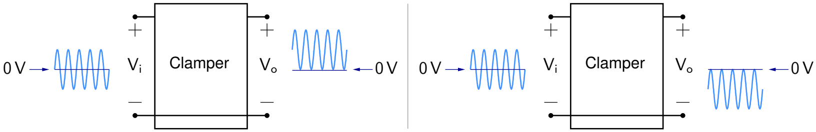
* A clamper circuit provides a "level shift." (The shape of the input signal is not altered.)

Clamper circuits



* A clamper circuit provides a "level shift." (The shape of the input signal is not altered.)
* The shift could be positive or negative.

Clamper circuits



* A clamper circuit provides a "level shift." (The shape of the input signal is not altered.)
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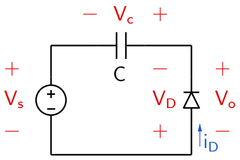
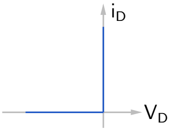
Clamper circuits





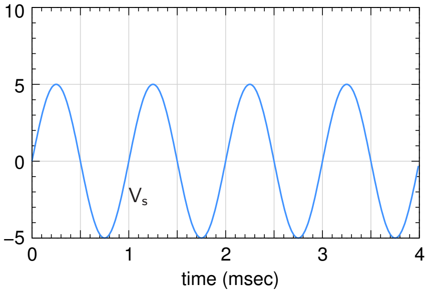
$V_s(t) = V_m \sin \omega t$

$V_c(0) = 0\,\mathrm{V}$

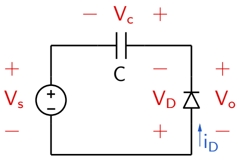
$V_{on} = 0\,\mathrm{V}$

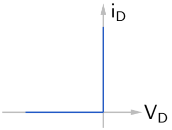
$R_{on} \to 0\,\Omega$

$R_{off} \to \infty\,\Omega$



Clamper circuits





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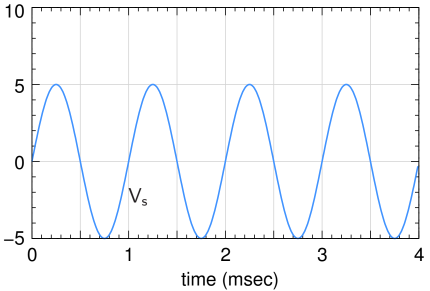
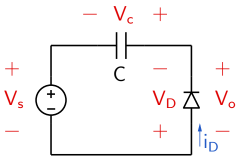
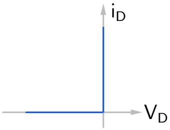


* When D conducts, the capacitor charges instantaneously since R_{on} is small. In this phase, $V_D = 0 \rightarrow V_c + V_s = 0 \rightarrow V_c = -V_s$.





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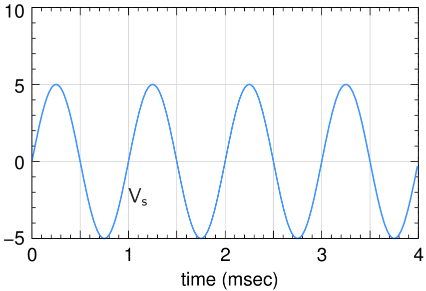

* When D conducts, the capacitor charges instantaneously since R_{on} is small. In this phase, $V_D = 0 \rightarrow V_c + V_s = 0 \rightarrow V_c = -V_s$.
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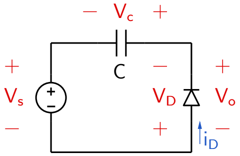
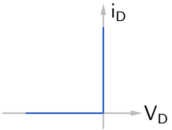
Clamper circuits





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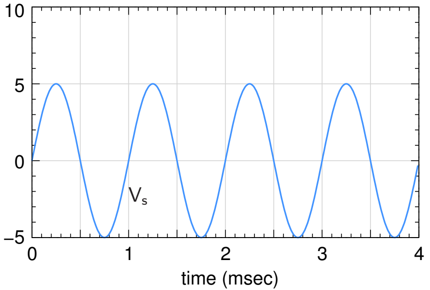

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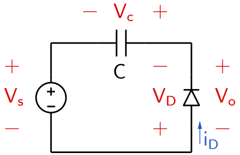
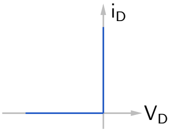
* After V_c reaches its maximum value (V_m), it cannot change any more. We then have $V_o(t) = V_s(t) + V_c(t) = V_s(t) + V_m$, i.e., a positive level shift.

Clamper circuits





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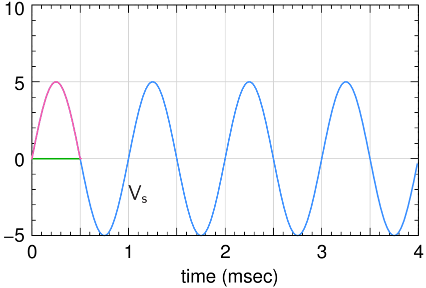

- ∗ When D conducts, the capacitor charges instantaneously since R_{on} is small. In this phase, $V_D = 0 \rightarrow V_c + V_s = 0 \rightarrow V_c = -V_s$.
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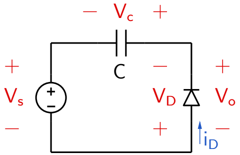
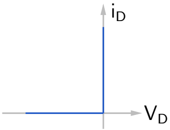
Clamper circuits





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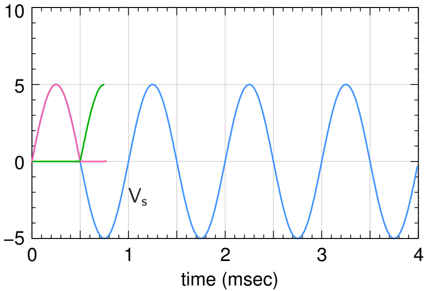

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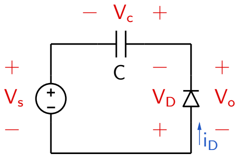
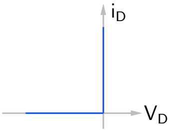
Clamper circuits





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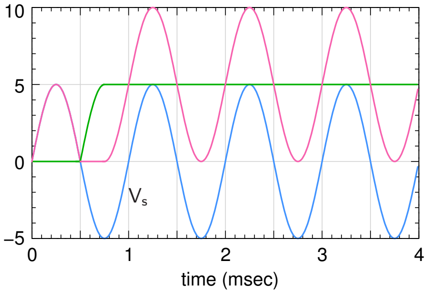

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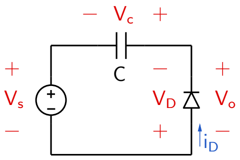
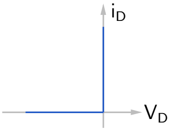
Clamper circuits





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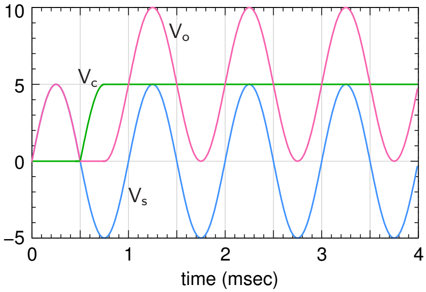

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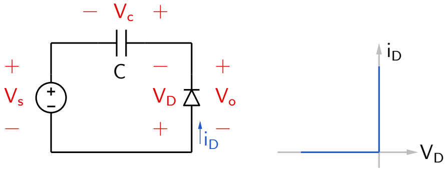

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$R_{on} \rightarrow 0\,\Omega$
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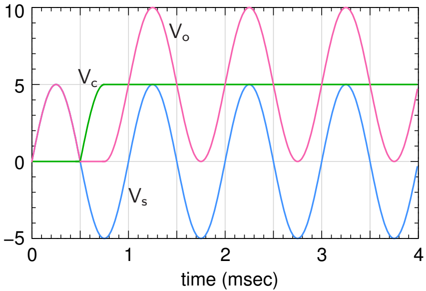

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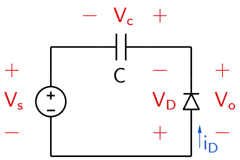
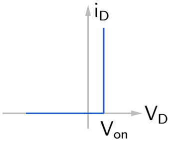
* Note that we are generally interested only in the steady-state behaviour and not in the transient at the beginning.





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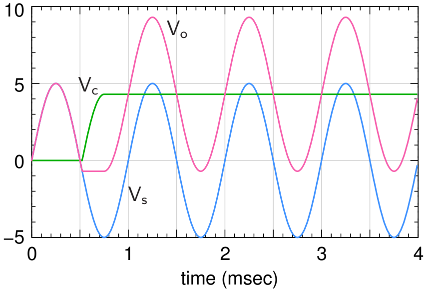

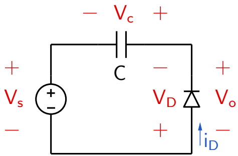

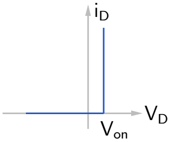

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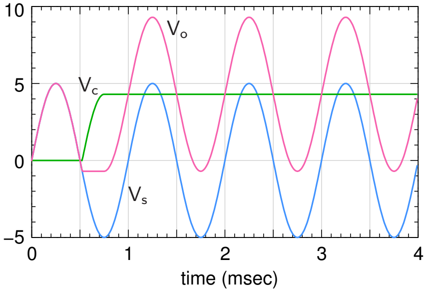
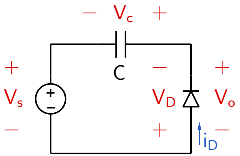
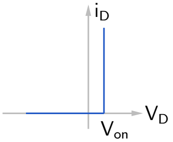


* When D conducts, the capacitor charges instantaneously since R_{on} is small (as in the last circuit). In this phase,
  $V_c + V_s + V_{on} = 0 \rightarrow V_c = -V_s - V_{on}$.

Clamper circuits





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$R_{on} \rightarrow 0\,\Omega$

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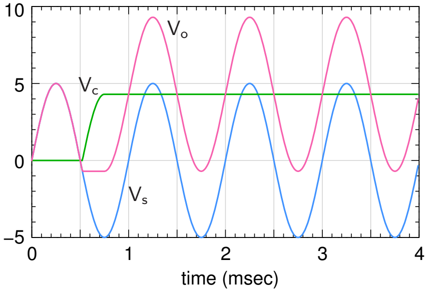

* When D conducts, the capacitor charges instantaneously since R_{on} is small (as in the last circuit).
  In this phase,
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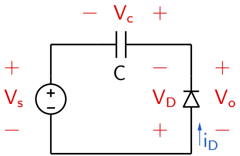
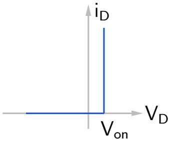
Clamper circuits





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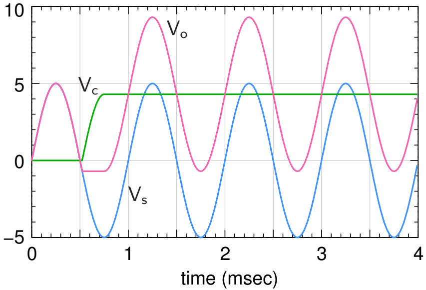
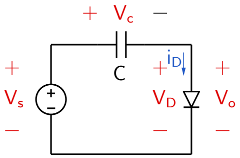
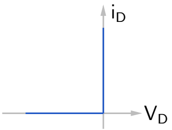


* When D conducts, the capacitor charges instantaneously since R_{on} is small (as in the last circuit). In this phase,
  $V_c + V_s + V_{on} = 0 \to V_c = -V_s - V_{on}$.
* V_c can only increase since a decrease in V_c would require the diode to conduct in the reverse direction.
* After V_c reaches its maximum value $(V_m - V_{on})$, it cannot change any more. We then have
  $V_o(t) = V_s(t) + V_c(t) = V_s(t) + V_m - V_{on}$. In this case, V_o gets clamped at $-0.7\,V$.

Clamper circuits





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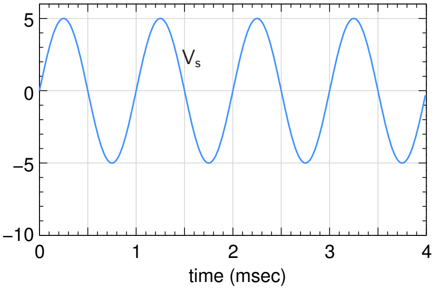
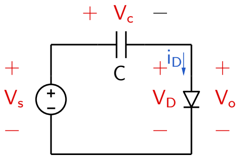
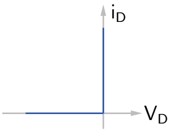


Clamper circuits





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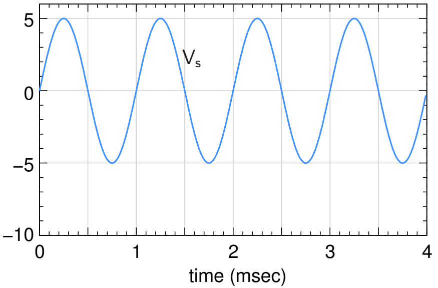
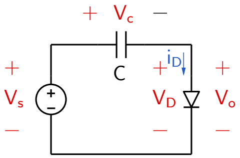
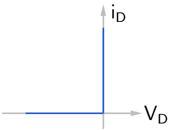


* When D conducts, the capacitor charges instantaneously since R_{on} is small. In this phase, $V_D = 0 \to V_c - V_s = 0 \to V_c = V_s$.

Clamper circuits





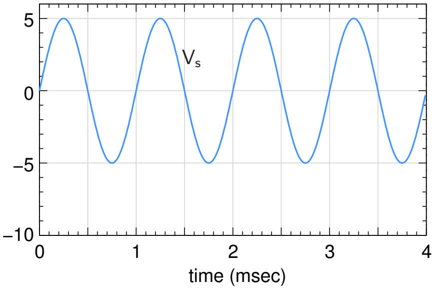
$V_s(t) = V_m \sin \omega t$

$V_c(0) = 0\,V$

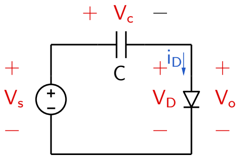
$V_{on} = 0\,V$

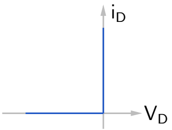
$R_{on} \rightarrow 0\,\Omega$

$R_{off} \rightarrow \infty\,\Omega$



* When D conducts, the capacitor charges instantaneously since R_{on} is small. In this phase, $V_D = 0 \rightarrow V_c - V_s = 0 \rightarrow V_c = V_s$.
* V_c can only increase since a decrease in V_c would require the diode to conduct in the reverse direction.





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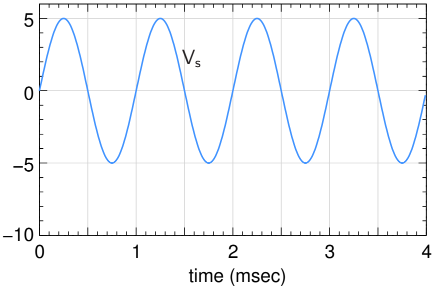
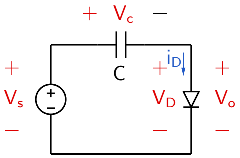
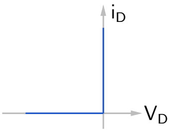


* When D conducts, the capacitor charges instantaneously since R_{on} is small. In this phase, $V_D = 0 \rightarrow V_c - V_s = 0 \rightarrow V_c = V_s$.
* V_c can only increase since a decrease in V_c would require the diode to conduct in the reverse direction.
* After V_c reaches its maximum value (V_m), it cannot change any more. We then have $V_o(t) = V_s(t) - V_c(t) = V_s(t) - V_m$, i.e., a negative level shift.

Clamper circuits





$V_s(t) = V_m \sin \omega t$

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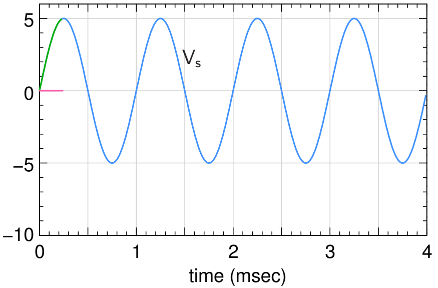

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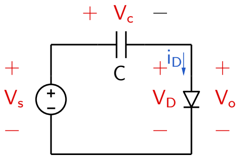
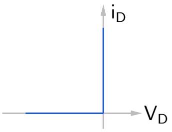
Clamper circuits





$V_s(t) = V_m \sin \omega t$

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$V_{on} = 0\,\mathrm{V}$

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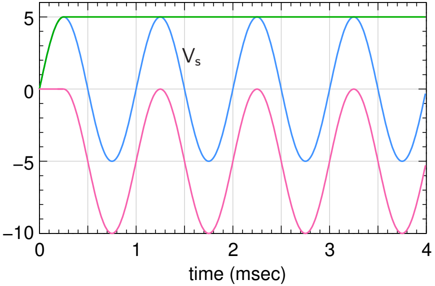

* When D conducts, the capacitor charges instantaneously since R_{on} is small. In this phase, $V_D = 0 \rightarrow V_c - V_s = 0 \rightarrow V_c = V_s$.
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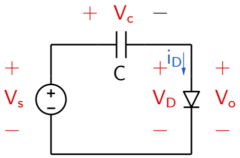

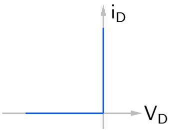

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$R_{on} \rightarrow 0\,\Omega$

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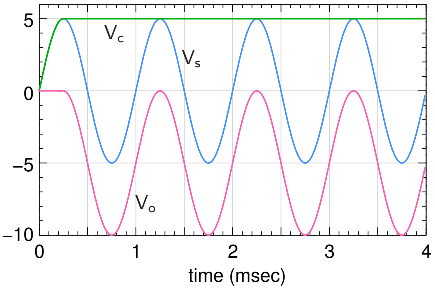

* When D conducts, the capacitor charges instantaneously since R_{on} is small. In this phase, $V_D = 0 \rightarrow V_c - V_s = 0 \rightarrow V_c = V_s$.
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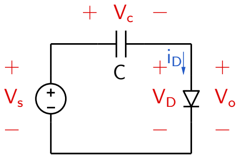
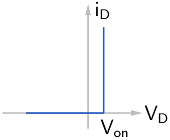




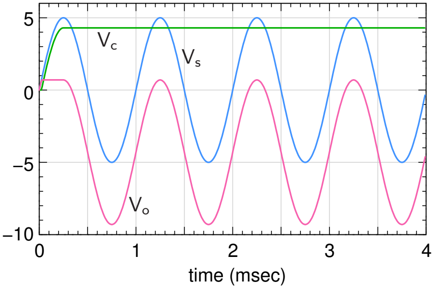
$V_s(t) = V_m \sin \omega t$

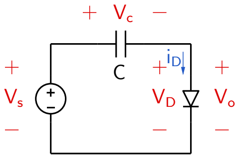
$V_c(0) = 0\,\text{V}$

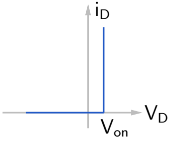
$V_{on} = 0.7\,\text{V}$

$R_{on} \to 0\,\Omega$

$R_{off} \to \infty\,\Omega$







$V_s(t) = V_m \sin \omega t$

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$V_{on} = 0.7\,\mathrm{V}$

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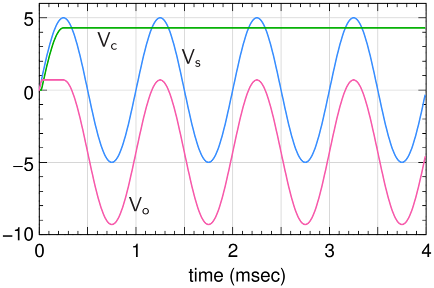
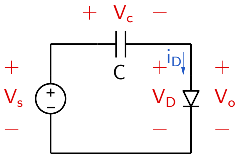
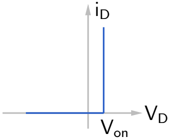


* When D conducts, the capacitor charges instantaneously since R_{on} is small (as in the last circuit). In this phase,
  $V_c + V_{on} - V_s = 0 \rightarrow V_c = V_s - V_{on}$.





$V_s(t) = V_m \sin \omega t$

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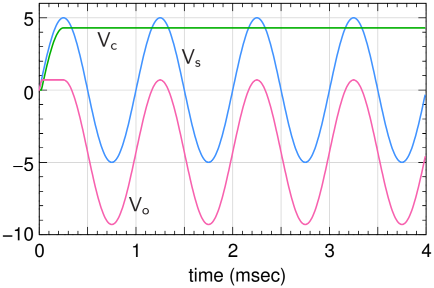
$R_{on} \rightarrow 0\,\Omega$

$R_{off} \rightarrow \infty\,\Omega$



* When D conducts, the capacitor charges instantaneously since R_{on} is small (as in the last circuit).
  In this phase,
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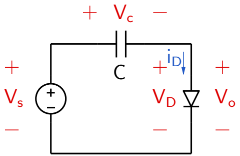
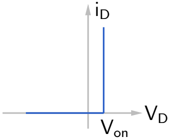
Clamper circuits





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$V_{on} = 0.7\,\text{V}$

$R_{on} \rightarrow 0\,\Omega$

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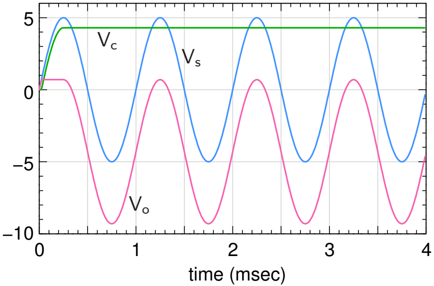
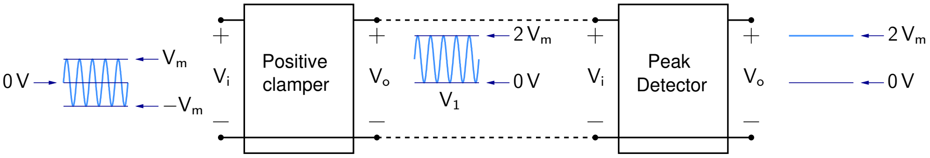


* When D conducts, the capacitor charges instantaneously since R_{on} is small (as in the last circuit). In this phase,
  $V_c + V_{on} - V_s = 0 \rightarrow V_c = V_s - V_{on}$.
* V_c can only increase since a decrease in V_c would require the diode to conduct in the reverse direction.
* After V_c reaches its maximum value $(V_m - V_{on})$, it cannot change any more. We then have $V_o(t) = V_s(t) - V_c(t) = V_s(t) - V_m + V_{on}$. In this case, V_o gets clamped at 0.7 V.

Voltage doubler (peak-to-peak detector)



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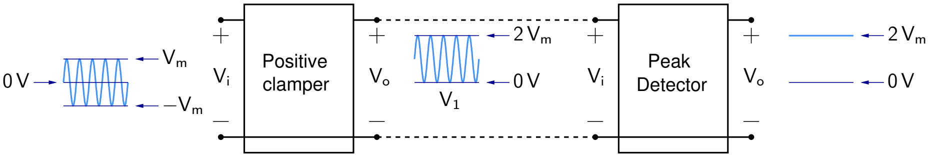
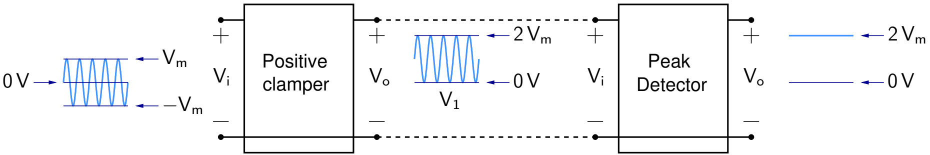
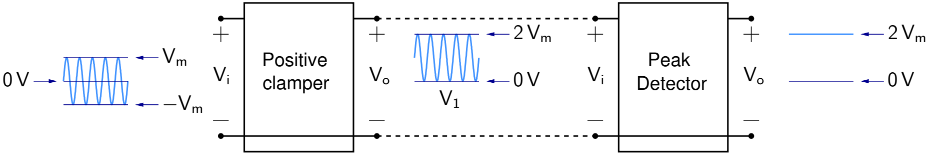


∗ Input voltage: $-V_m$ to V_m

Voltage doubler (peak-to-peak detector)

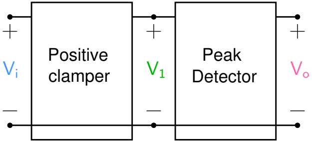


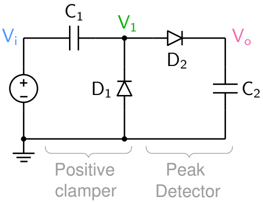
* Input voltage: $-V_m$ to V_m
* Output of positive clamper (V_1): 0 to 2 V_m



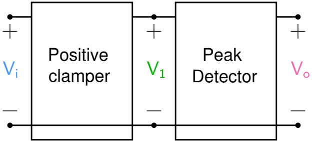
* Input voltage: $-V_m$ to V_m
* Output of positive clamper (V_1): 0 to 2 V_m
* The peak detector detects the peak of $V_1(t)$, i.e., 2 V_m (dc).

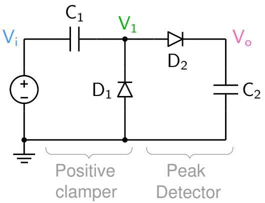
Voltage doubler (peak-to-peak detector)

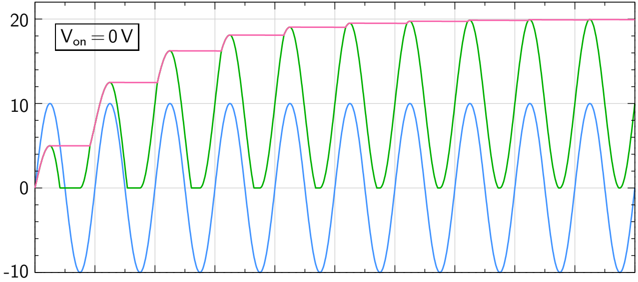




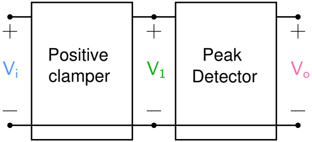
Voltage doubler (peak-to-peak detector)

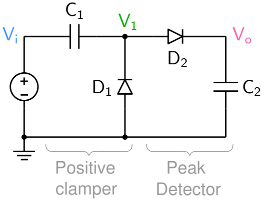


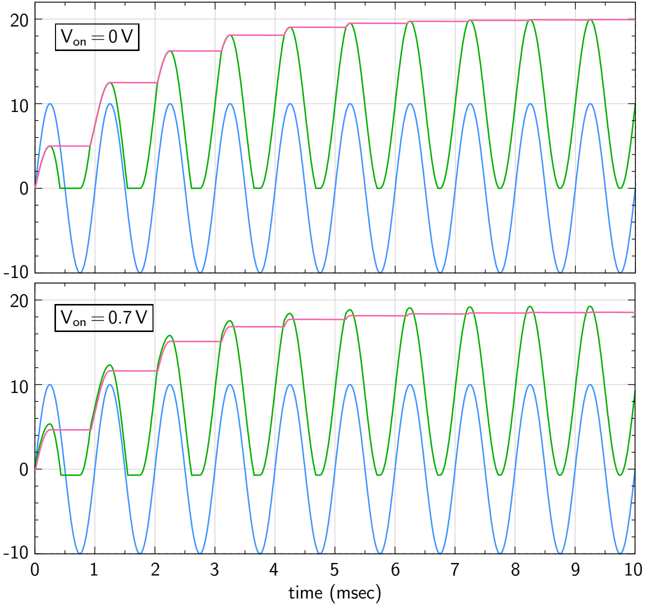




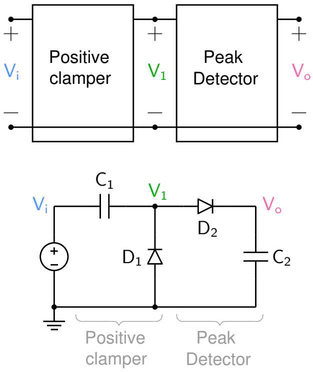
Voltage doubler (peak-to-peak detector)



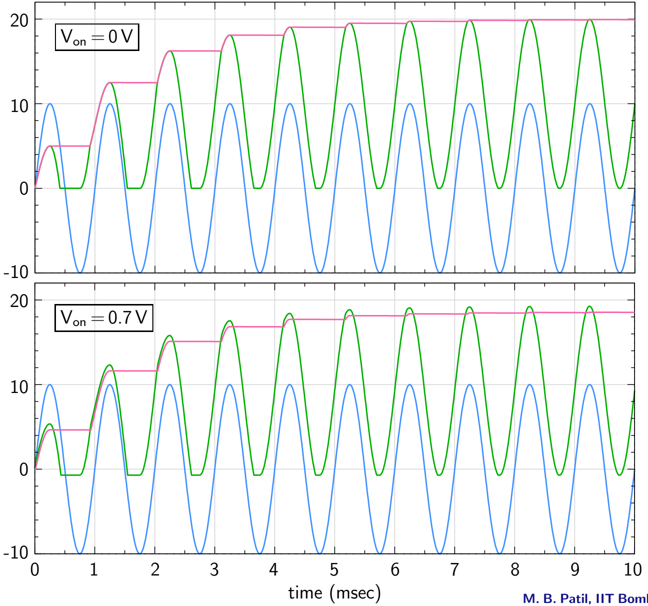




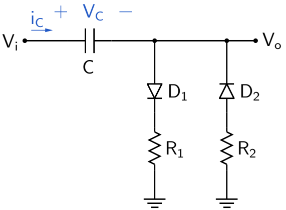
Voltage doubler (peak-to-peak detector)

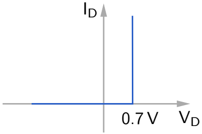


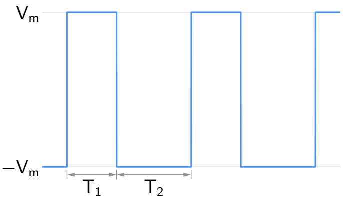
SEQUEL file: `ee101_voltage_doubler.sqproj`



Diode circuit example

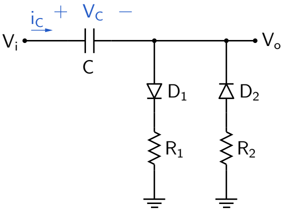


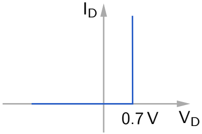


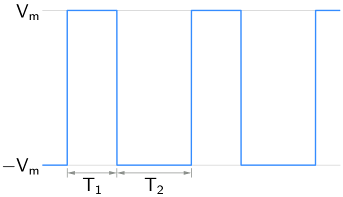


Assuming $R_1 C$ and $R_2 C$ to be large compared to T, find $V_o(t)$ in steady state.

Diode circuit example



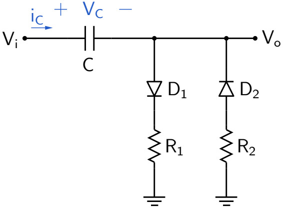


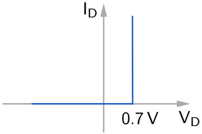


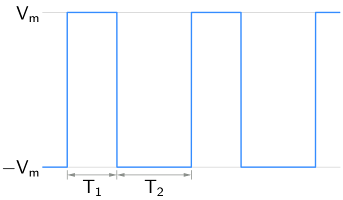
Assuming $R_1 C$ and $R_2 C$ to be large compared to T, find $V_o(t)$ in steady state.

* Charging time constant $\tau_1 = R_1 C$.

Diode circuit example



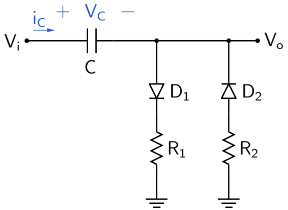


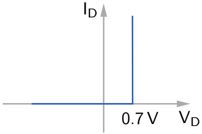


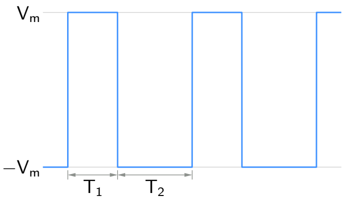
Assuming R_1C and R_2C to be large compared to T, find $V_o(t)$ in steady state.

- Charging time constant $\tau_1 = R_1C$.
- Discharging time constant $\tau_2 = R_2C$.

Diode circuit example



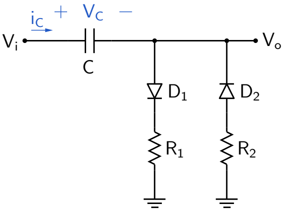


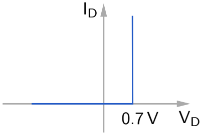


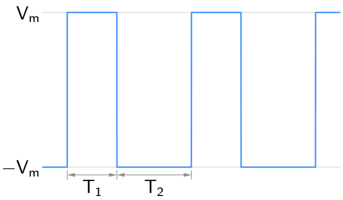
Assuming R_1C and R_2C to be large compared to T, find $V_o(t)$ in steady state.

* Charging time constant $\tau_1 = R_1C$.
* Discharging time constant $\tau_2 = R_2C$.
* Since $\tau_1 \gg T$ and $\tau_2 \gg T$, we expect V_C to be nearly constant in steady state,
  i.e., $V_C(t) \approx \text{constant} \equiv V_C^0$.

Diode circuit example







Assuming R_1C and R_2C to be large compared to T, find $V_o(t)$ in steady state.

* Charging time constant $\tau_1 = R_1C$.
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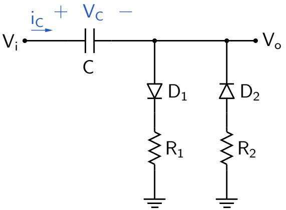
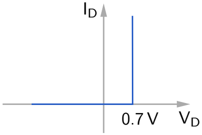
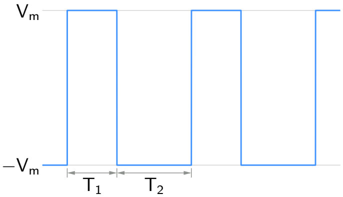
* Since $\tau_1 \gg T$ and $\tau_2 \gg T$, we expect V_C to be nearly constant in steady state,
  i.e., $V_C(t) \approx \text{constant} \equiv V_C^0$.
* $V_o(t) = V_i(t) - V_C(t) \approx V_i(t) - V_C^0$.

Diode circuit example



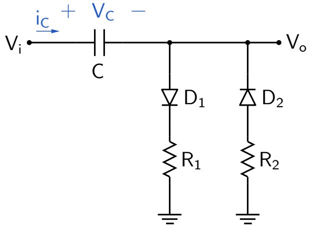




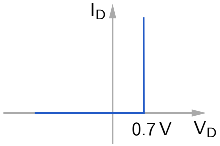
Assuming $R_1 C$ and $R_2 C$ to be large compared to T, find $V_o(t)$ in steady state.

* Charging time constant $\tau_1 = R_1 C$.
* Discharging time constant $\tau_2 = R_2 C$.
* Since $\tau_1 \gg T$ and $\tau_2 \gg T$, we expect V_C to be nearly constant in steady state,
  i.e., $V_C(t) \approx \text{constant} \equiv V_C^0$.
* $V_o(t) = V_i(t) - V_C(t) \approx V_i(t) - V_C^0$.

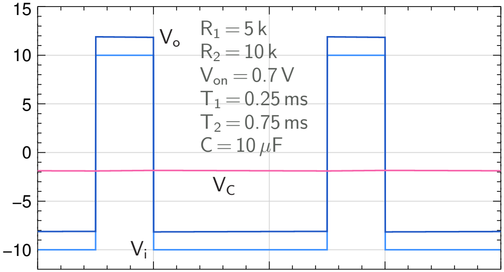
Let us look at an example.



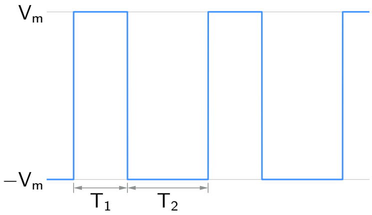
i_C
+ V_C −
V_i
V_o
C
D_1
D_2
R_1
R_2



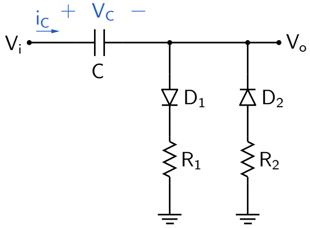
I_D
0.7 V
V_D



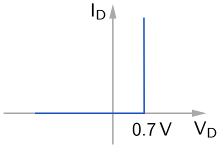
15
10
5
0
−5
−10
V_o
$R_1 = 5\,k$
$R_2 = 10\,k$
$V_{on} = 0.7\,V$
$T_1 = 0.25\,ms$
$T_2 = 0.75\,ms$
$C = 10\,\mu F$
V_C
V_i



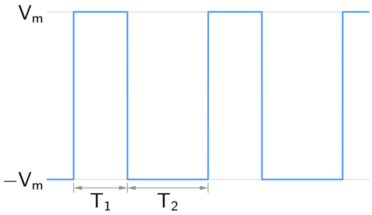
V_m
$-V_m$
T_1
T_2



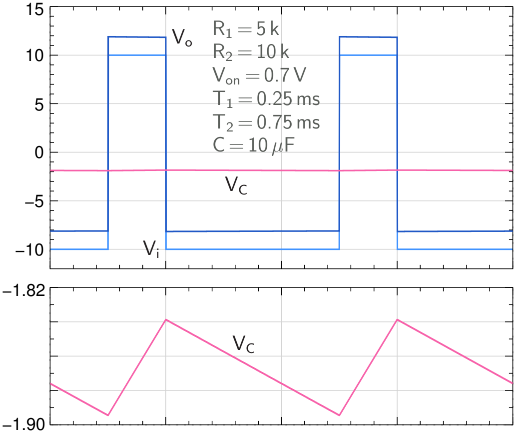
i_C
+ V_C −
V_i
V_o
C
D_1
D_2
R_1
R_2



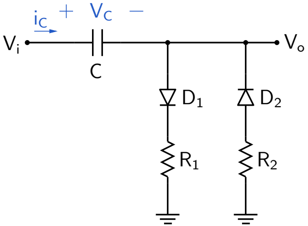
I_D
0.7 V
V_D

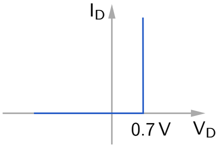


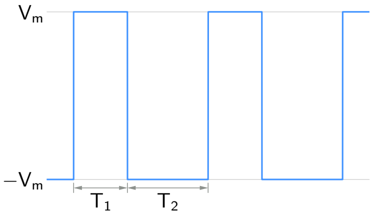
V_m
$-V_m$
T_1
T_2

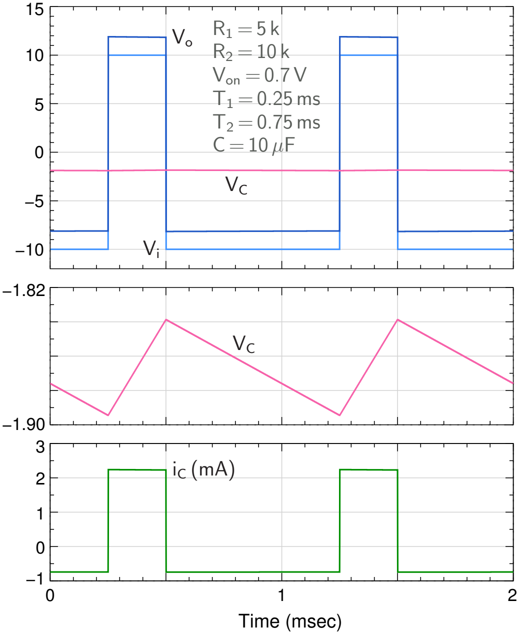


15
10
5
0
−5
−10
V_o
V_i
V_C
$R_1 = 5\,k$
$R_2 = 10\,k$
$V_{on} = 0.7\,V$
$T_1 = 0.25\,ms$
$T_2 = 0.75\,ms$
$C = 10\,\mu F$
−1.82
−1.90
V_C



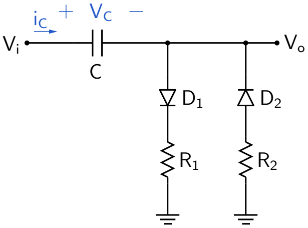


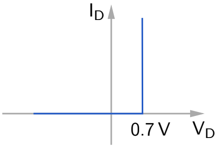


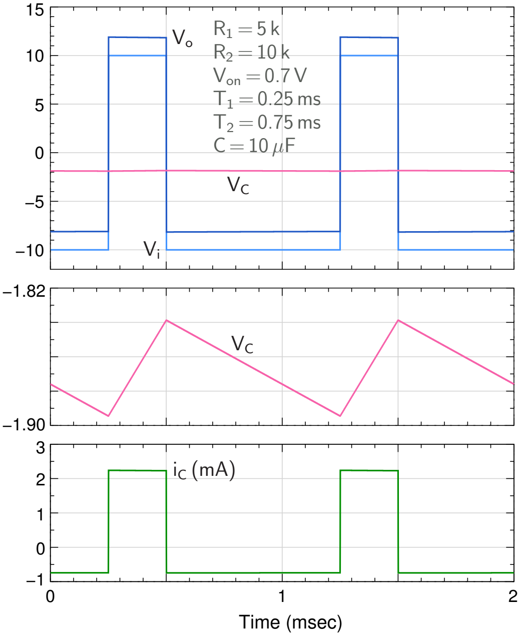


SEQUEL file:

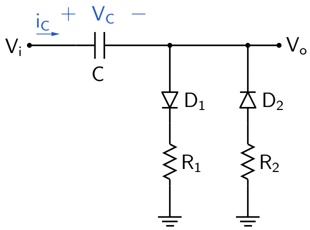
`ee101_diode_circuit_13.sqproj`

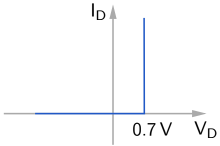


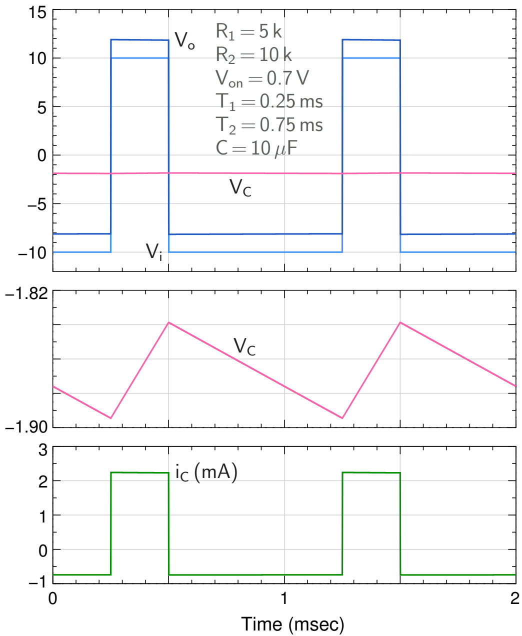




Charge conservation:

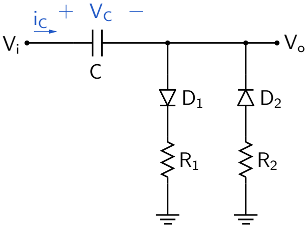


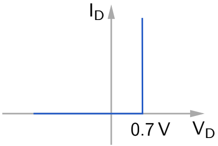




Charge conservation:

$$\Delta Q = \int_0^T i_C\,dt = \int_0^{T_1} i_C\,dt + \int_{T_1}^{T_1+T_2} i_C\,dt = 0.$$

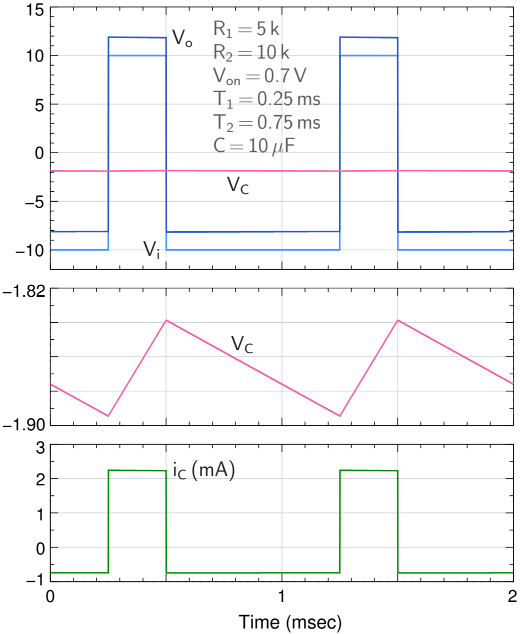


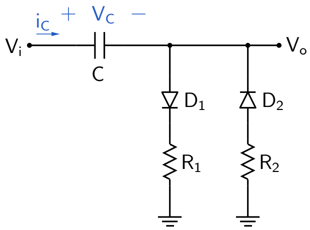


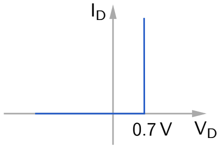
Charge conservation:

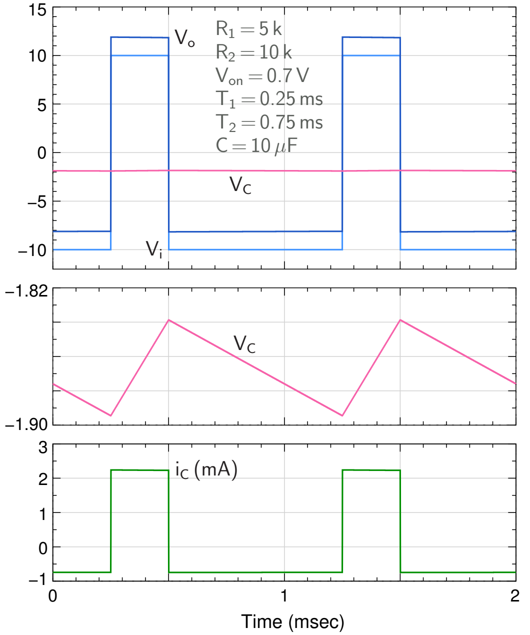
$$\Delta Q = \int_0^T i_C\,dt = \int_0^{T_1} i_C\,dt + \int_{T_1}^{T_1+T_2} i_C\,dt = 0.$$

$$T_1\left(\frac{V_m - V_C - V_{on}}{R_1}\right) - T_2\left(\frac{0-(-V_m - V_C) - V_{on}}{R_2}\right) = 0.$$









Charge conservation:

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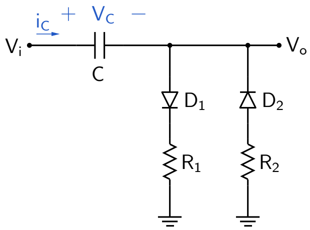
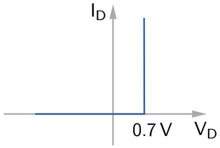
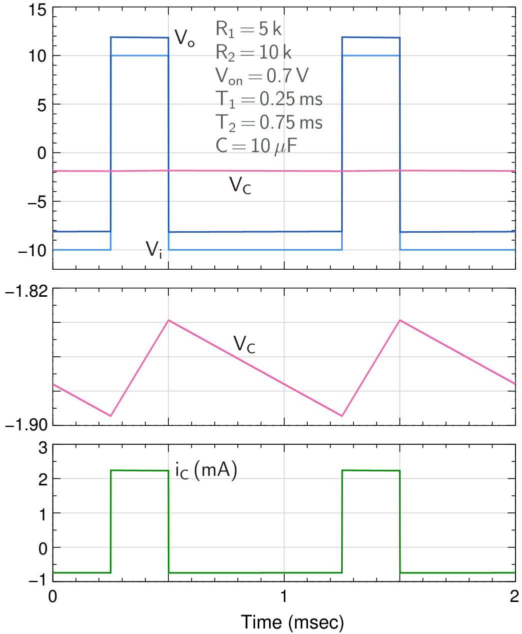
$$\left(\frac{T_1}{R_1} - \frac{T_2}{R_2}\right)(V_m - V_{on}) = V_C\left(\frac{T_1}{R_1} + \frac{T_2}{R_2}\right).$$







Charge conservation:

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$$\left(\frac{T_1}{R_1} - \frac{T_2}{R_2}\right)(V_m - V_{on}) = V_C\left(\frac{T_1}{R_1} + \frac{T_2}{R_2}\right).$$

$$\rightarrow V_C = \frac{\left(\dfrac{T_1}{R_1} - \dfrac{T_2}{R_2}\right)}{\left(\dfrac{T_1}{R_1} + \dfrac{T_2}{R_2}\right)}(V_m - V_{on}) = -1.86\,\text{V}.$$

Rectifiers

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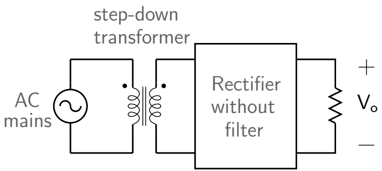
Rectifiers



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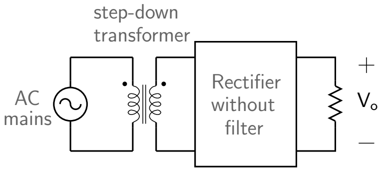
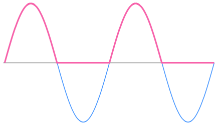
Rectifiers





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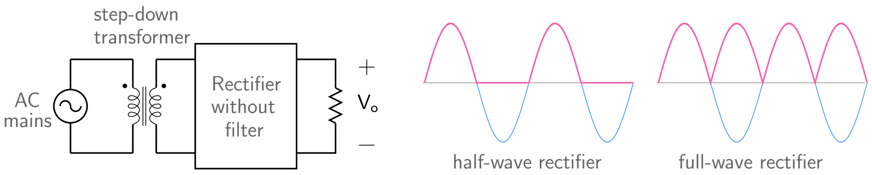
Rectifiers



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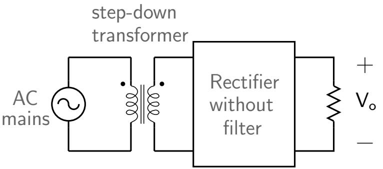
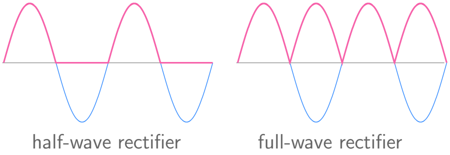
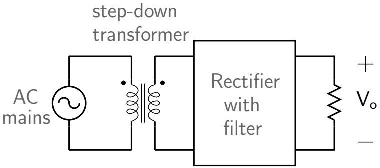
Rectifiers







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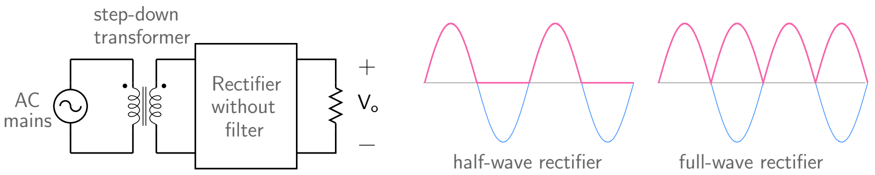
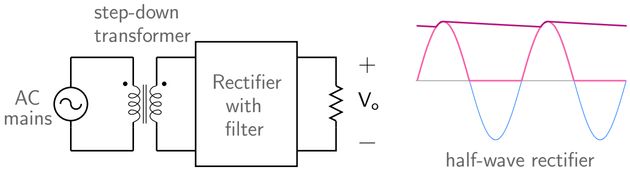
Rectifiers





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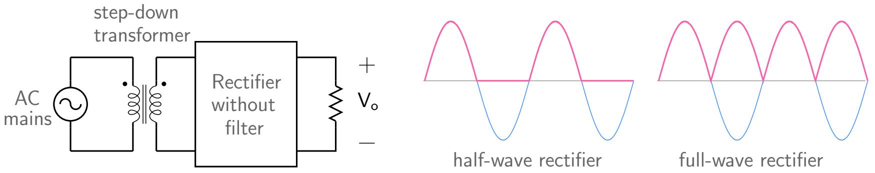
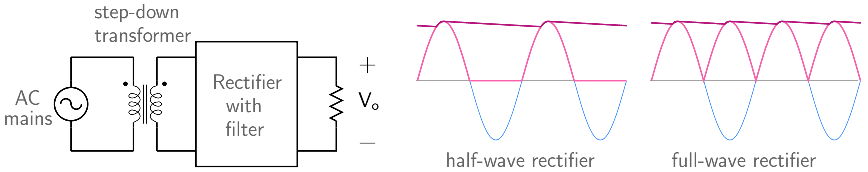
Rectifiers





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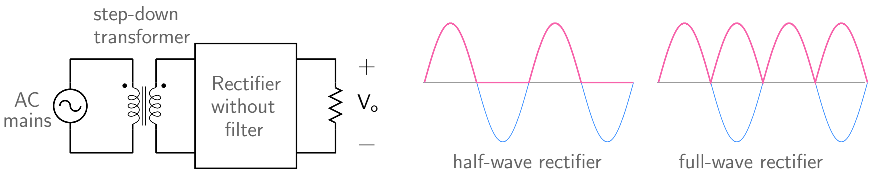
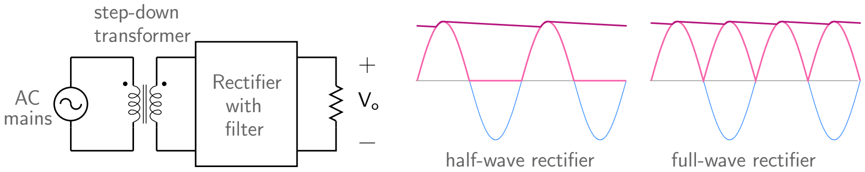
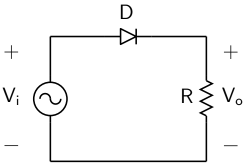
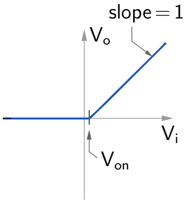
Rectifiers

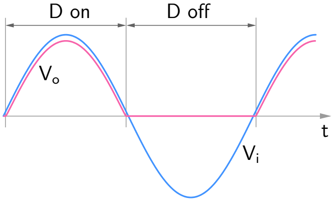




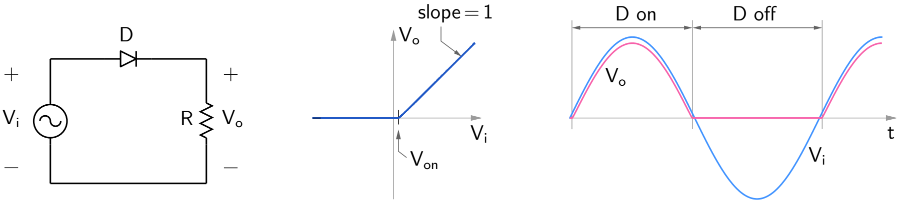
* A rectifier is used to convert an AC voltage to a DC voltage (typically 5 to 20 V), e.g., a mobile phone charger.
* AC mains → step-down transformer → DC voltage OR
  AC mains → DC voltage → lower DC voltage
* A voltage regulator would be typically used to remove the ripple riding on the DC output.


D
+
Vi
−
R
+
Vo
−


slope = 1
Vo
Vi
Von

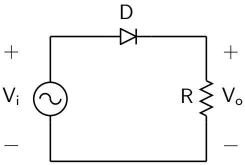

D on
D off
Vo
t
Vi

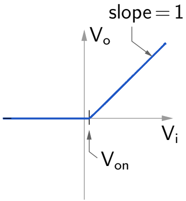
Half-wave rectifier without filter

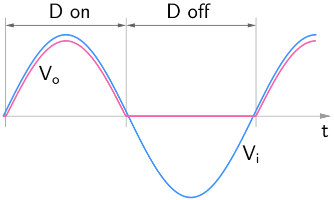


* D conducts only if $V_i > V_{\text{on}}$, and in that case $V_o = V_i - V_{\text{on}}$, a straight line with slope = 1.

Half-wave rectifier without filter

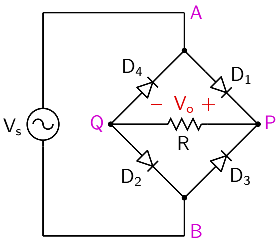




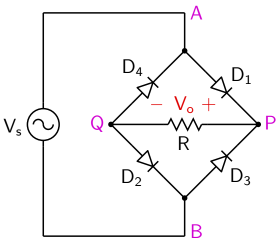


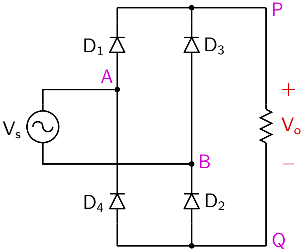
* D conducts only if $V_i > V_{\text{on}}$, and in that case $V_o = V_i - V_{\text{on}}$, a straight line with slope $= 1$.
* If $V_i < V_{\text{on}}$, D does not conduct $\rightarrow$ $V_o = 0$.

Full-wave (bridge) rectifier without filter

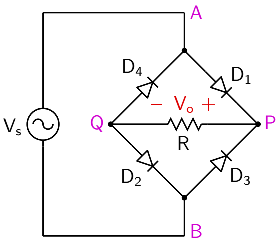


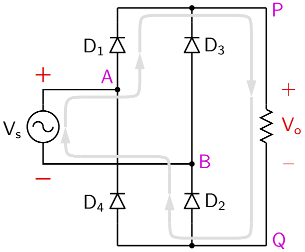
Full-wave (bridge) rectifier without filter





Full-wave (bridge) rectifier without filter





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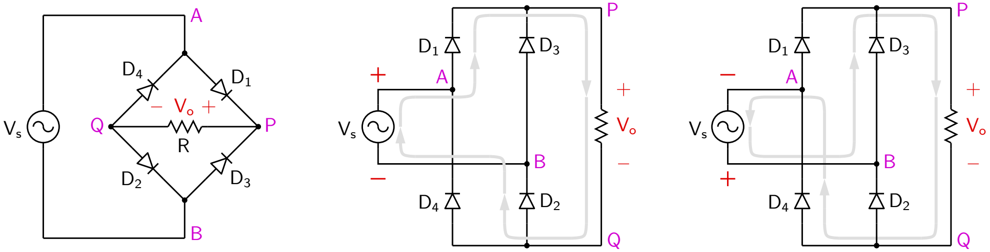

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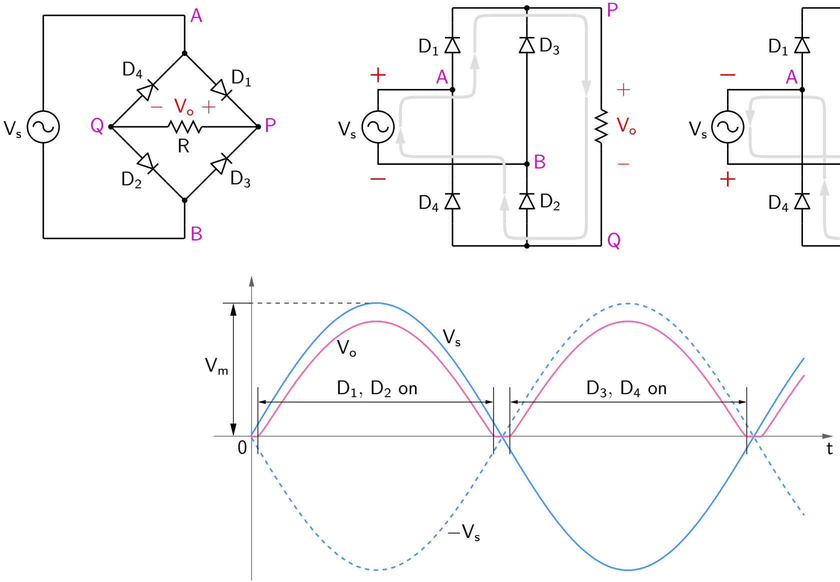
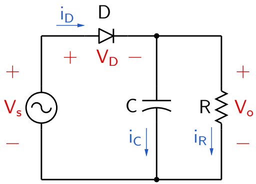
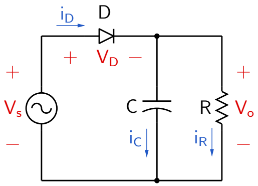


Half-wave rectifier with capacitor filter

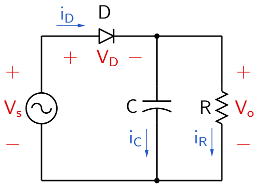


Half-wave rectifier with capacitor filter



* Similar to the peak detector except that the load resistance provides a discharge path for the capacitor in this case.

Half-wave rectifier with capacitor filter



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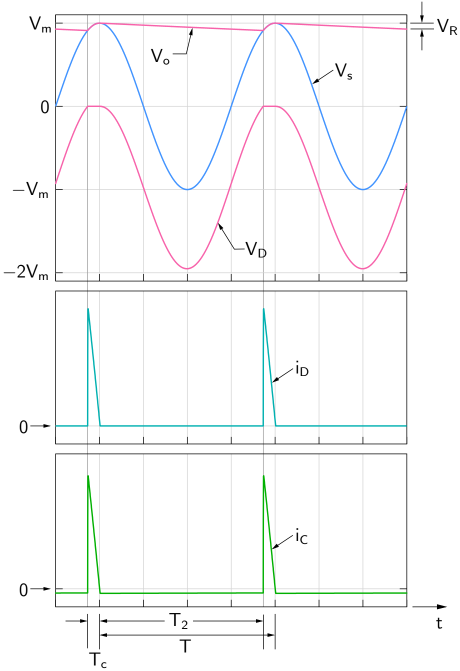
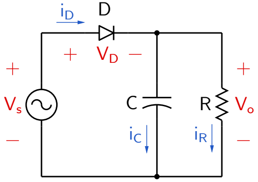
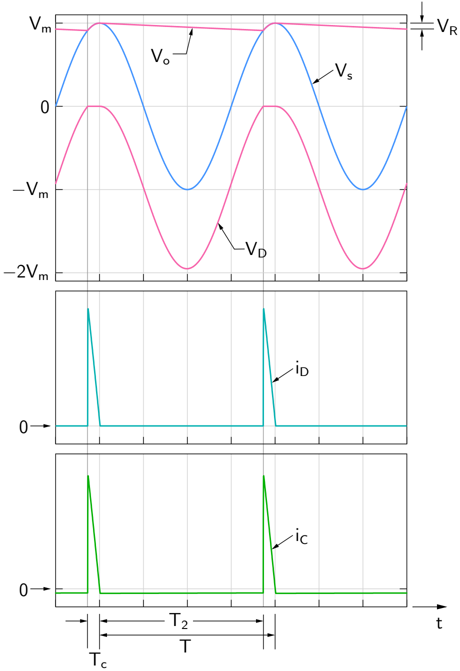


Half-wave rectifier with capacitor filter

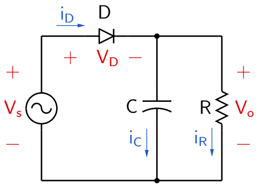


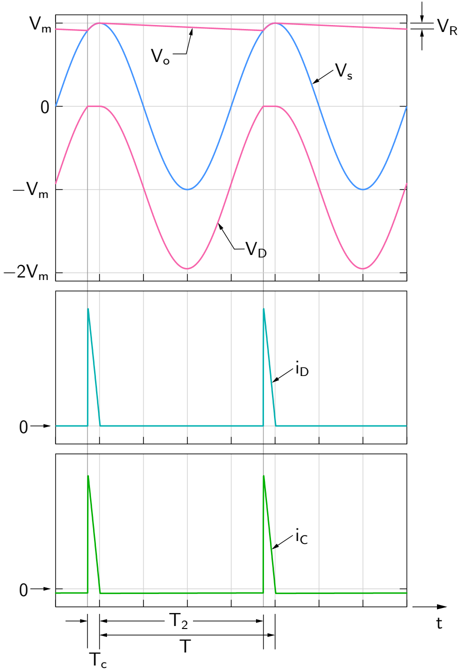


∗ Similar to the peak detector except that the load resistance provides a discharge path for the capacitor in this case.

∗ Because of the load current i_R, there is a drop in the output voltage → "ripple"

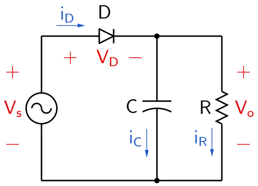
Half-wave rectifier with capacitor filter





- Similar to the peak detector except that the load resistance provides a discharge path for the capacitor in this case.
- Because of the load current i_R, there is a drop in the output voltage → "ripple"
- The peak diode current is much larger than the average load current.

Half-wave rectifier with capacitor filter



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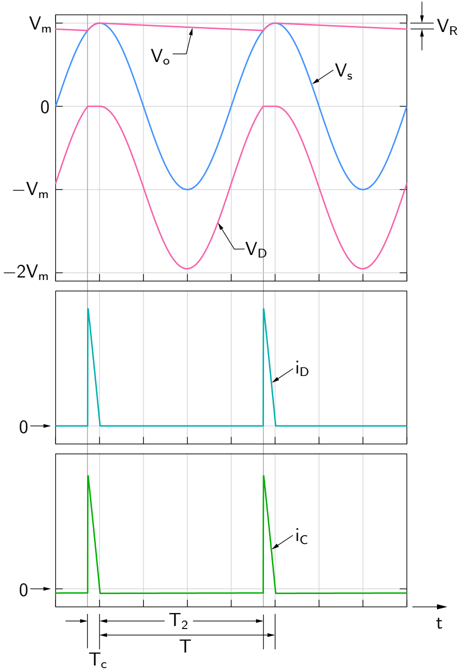
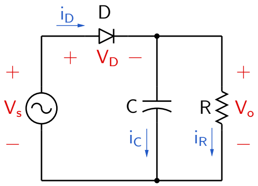
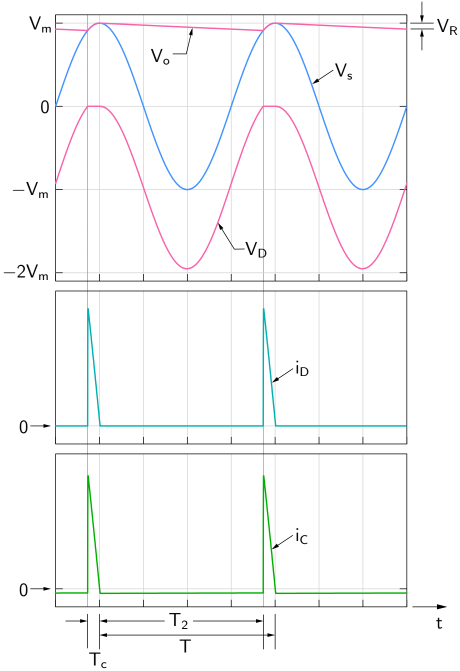
* Because of the load current i_R, there is a drop in the output voltage → “ripple”
* The peak diode current is much larger than the average load current.
* $V_D(t) = V_s(t) - V_o(t) \approx V_s(t) - V_m$
  → The maximum reverse bias (“Peak Inverse Voltage” or PIV) across the diode is 2 V_m.



Half-wave rectifier with capacitor filter





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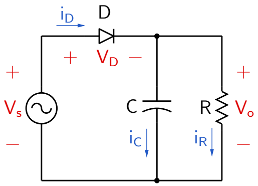
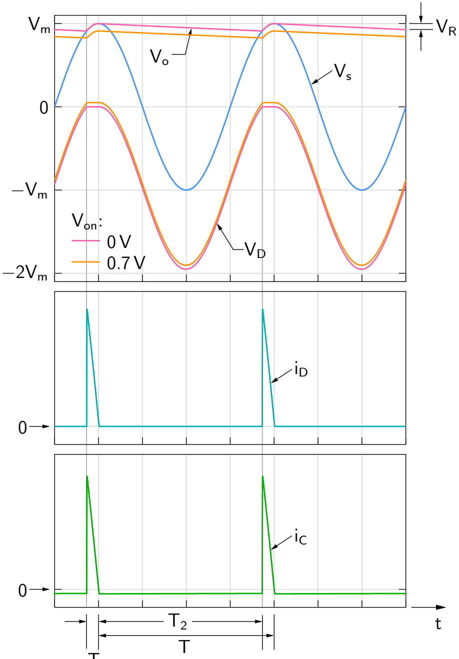
Half-wave rectifier with capacitor filter





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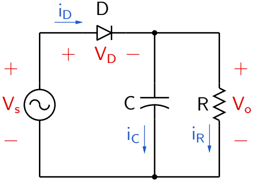
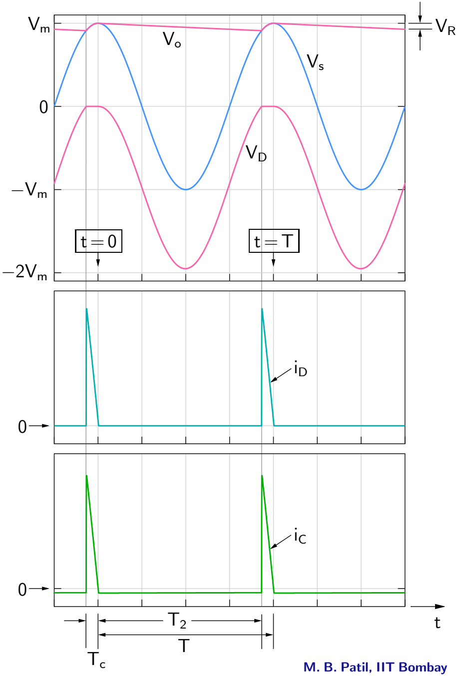
Half-wave rectifier with capacitor filter



$V_m = 16\,\text{V}$, $f = 50\,\text{Hz}$, $R = 100\,\Omega$. For a ripple voltage $V_R = 2\,\text{V}$, find (a) the filter capacitance C, (b) average and peak diode currents, (c) maximum reverse voltage across the diode.
(Let $V_{on} = 0\,\text{V}$.)



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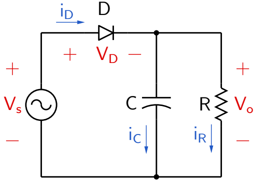
Half-wave rectifier with capacitor filter



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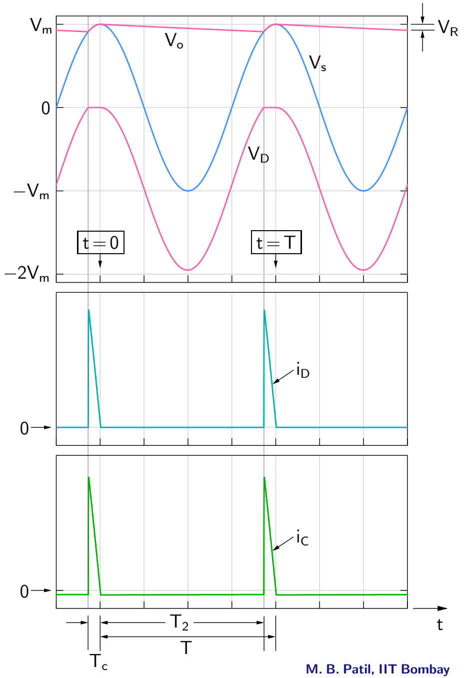
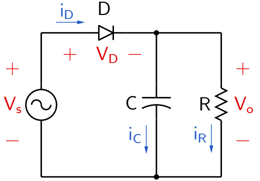
(a) filter capacitance

1. In the discharge phase,

$$V_o(t) = V_m e^{-t/\tau} \approx V_m \left(1 - \frac{t}{\tau}\right).$$



Half-wave rectifier with capacitor filter



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(a) filter capacitance

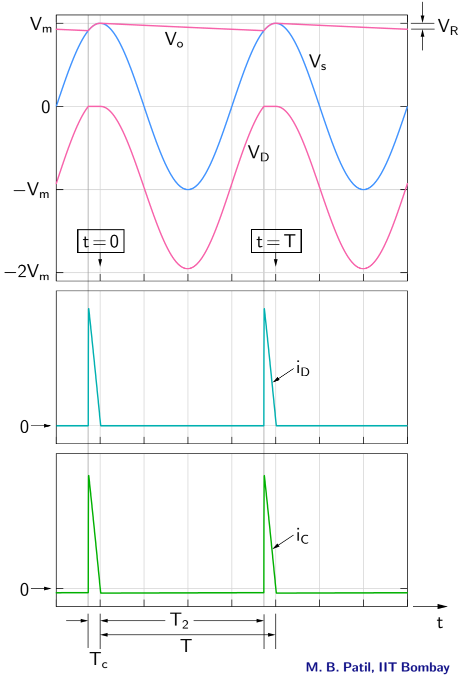
1. In the discharge phase,

   $$V_o(t) = V_m e^{-t/\tau} \approx V_m \left(1 - \frac{t}{\tau}\right).$$

   The drop in $V_o(t)$ is given by the second term.

   Using $T_2 \approx T$,

   $$V_R = V_m \frac{T}{\tau} = V_m \frac{T}{RC}.$$



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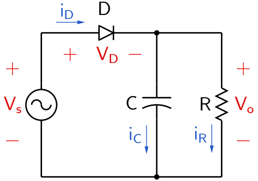
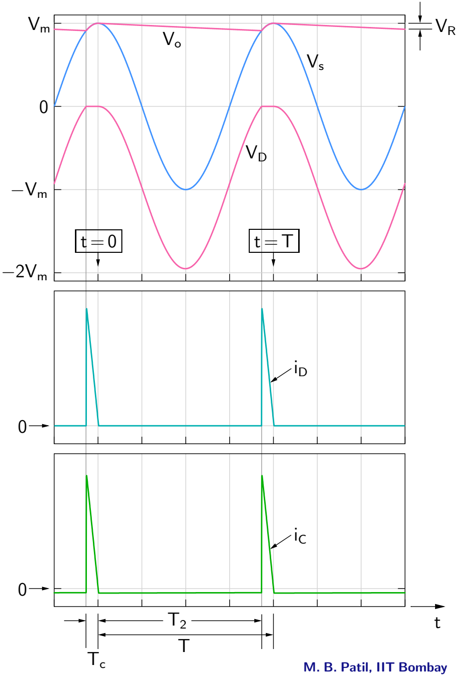
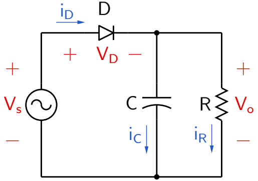
Half-wave rectifier with capacitor filter





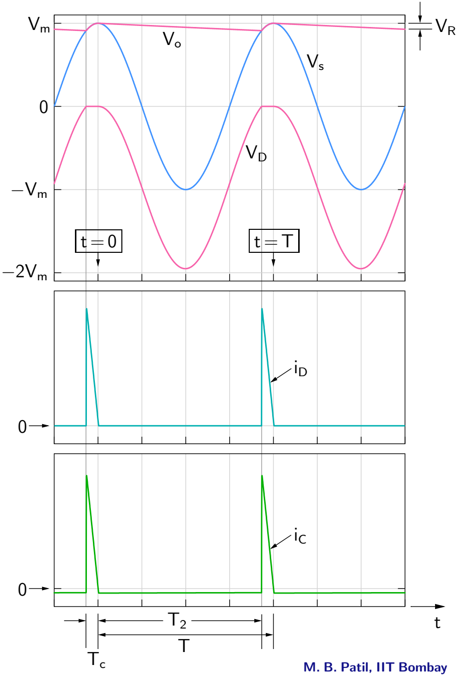
Half-wave rectifier with capacitor filter



(a) Ripple voltage V_R

2. Assuming $i_C = i_R = \dfrac{V_o}{R} \approx \dfrac{V_m}{R}$ in the discharge phase, we get

$$i_C = \frac{V_m}{R} = C\,\frac{\Delta V_o}{\Delta t} \approx C\,\frac{V_R}{T} \rightarrow V_R = V_m\,\frac{T}{RC}.$$



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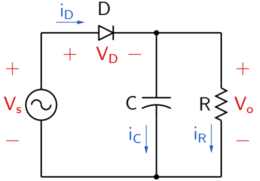
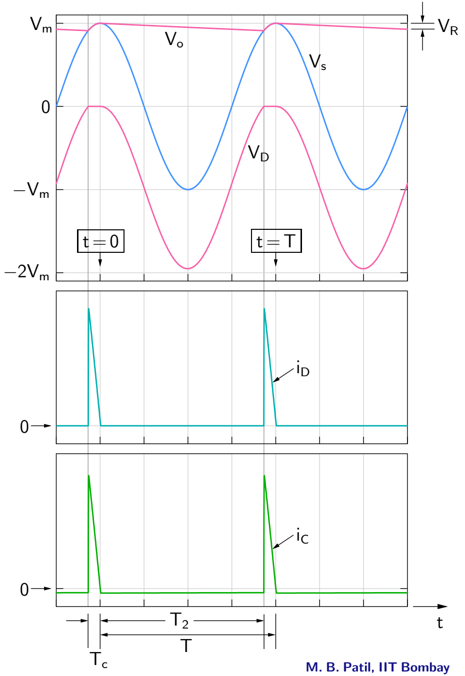
Half-wave rectifier with capacitor filter





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$$\rightarrow C = \frac{V_m}{V_R}\,\frac{T}{R} = \frac{16\,\text{V}}{2\,\text{V}}\,\frac{20\,\text{ms}}{100\,\Omega} = 1600\,\mu\text{F}.$$

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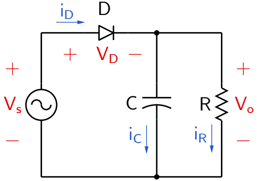
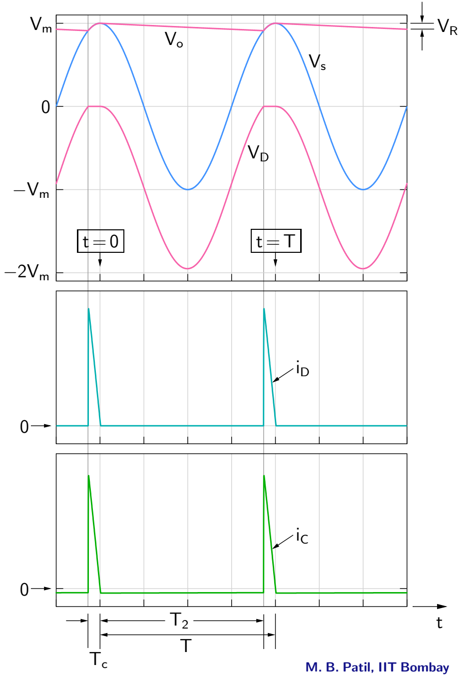
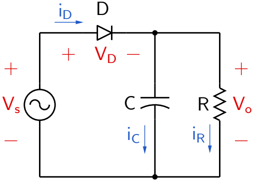
Half-wave rectifier with capacitor filter



(b) Average diode current



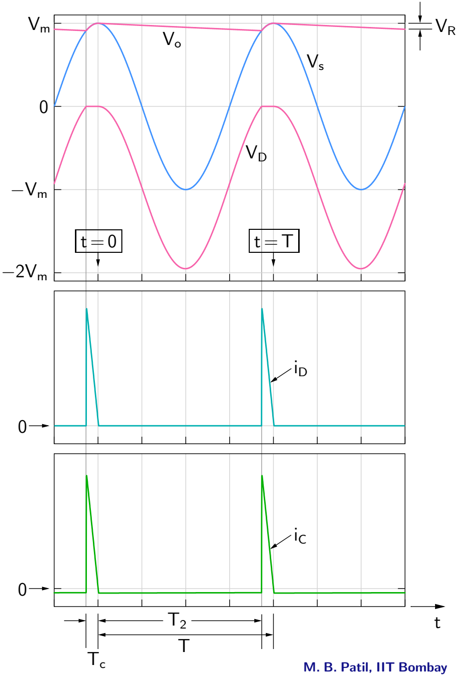
Half-wave rectifier with capacitor filter



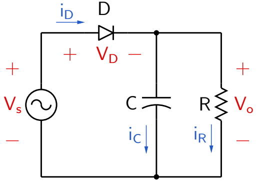
(b) Average diode current

Using charge balance,

$$\int_{T-T_c}^{T} (i_D - i_R)\, dt = \int_0^{T-T_c} i_R\, dt$$



Half-wave rectifier with capacitor filter



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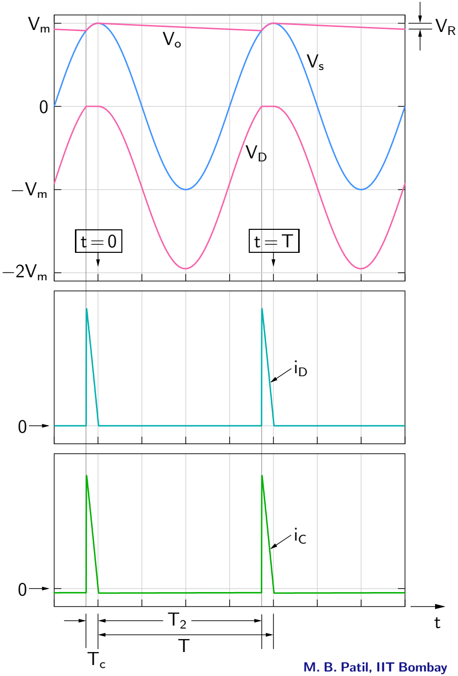
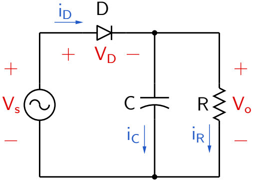


Half-wave rectifier with capacitor filter



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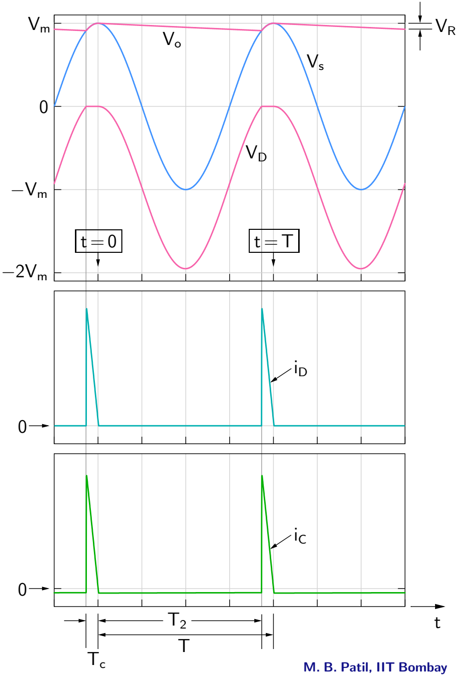
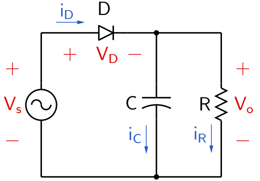
$$\rightarrow \int_{T-T_c}^{T} i_D\,dt = \int_0^{T} i_R\,dt.$$

$$i_D^{av} = \frac{1}{T}\int_0^{T} i_D\,dt = \frac{1}{T}\int_{T-T_c}^{T} i_D\,dt$$

$$= \frac{1}{T}\int_0^{T} i_R\,dt \approx \frac{V_m}{R}.$$



Half-wave rectifier with capacitor filter



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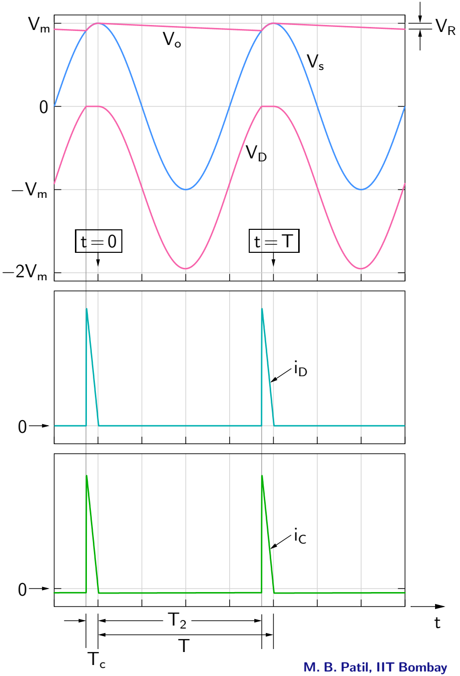
$$i_D^{av} \approx \frac{16\,\text{V}}{100\,\Omega} = 160\,\text{mA}.$$



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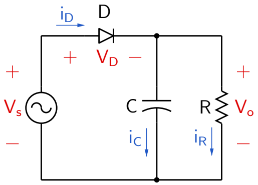
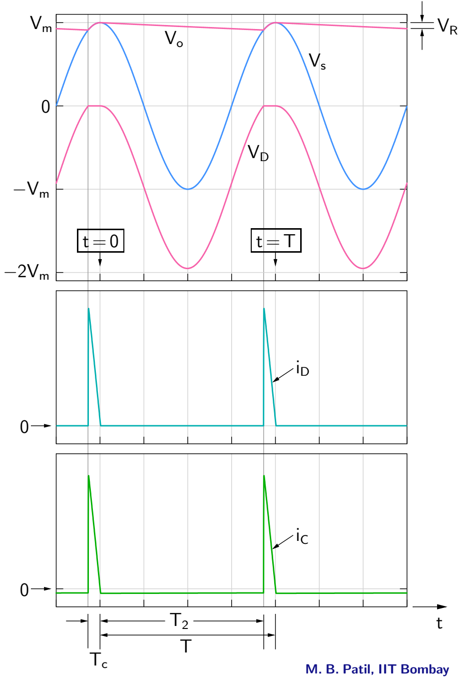


(b) Peak diode current



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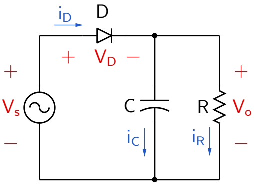
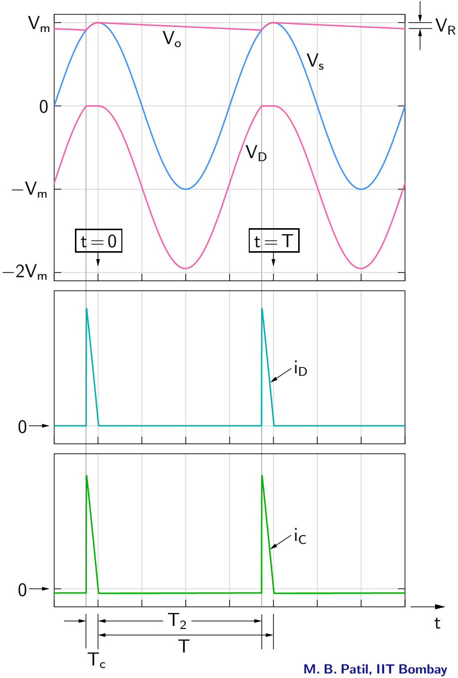
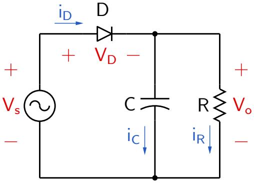


(b) Peak diode current

$$i_D^{\text{peak}} = C\,\frac{d}{dt}\,(V_m \cos\omega t)\Big|_{t=-T_c} + \frac{V_m}{R}$$





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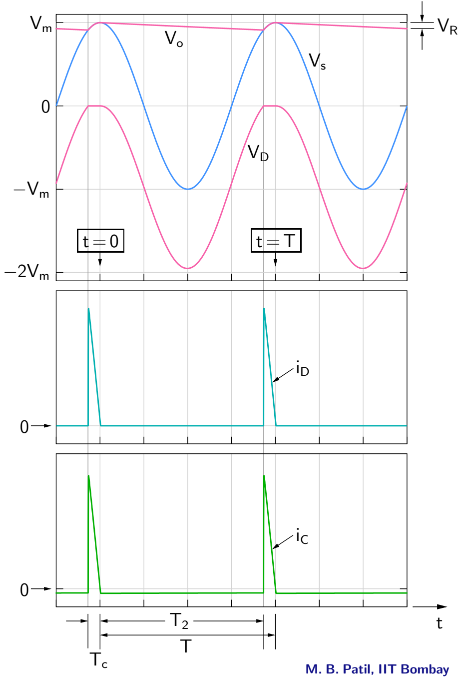
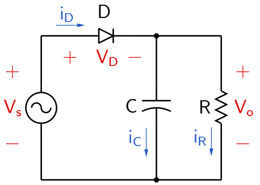
$$= -\omega C\, V_m \sin(-\omega T_c) + \frac{16\,\text{V}}{100\,\Omega}$$

$$= \omega C\, V_m \sin \omega T_c + 0.16$$





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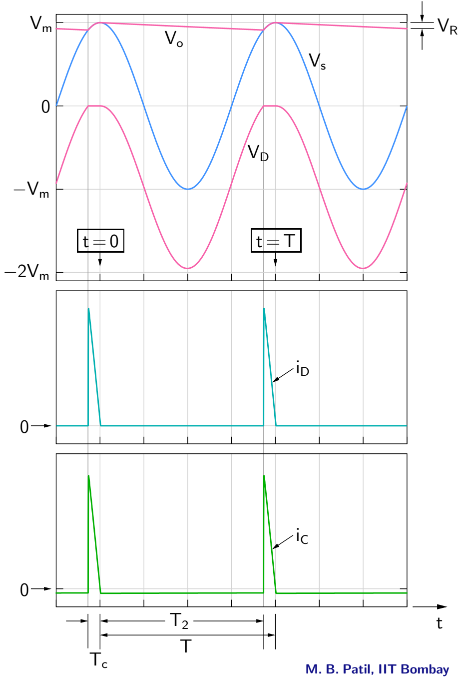
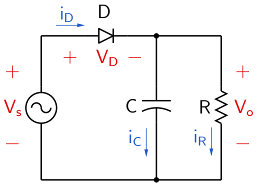
$V_m\ \cos(-\omega T_c) = V_m - V_R,$ giving

$$\omega T_c = \cos^{-1}\left(1 - \frac{V_R}{V_m}\right) = \cos^{-1}\left(1 - \frac{2}{16}\right) = 29^\circ.$$





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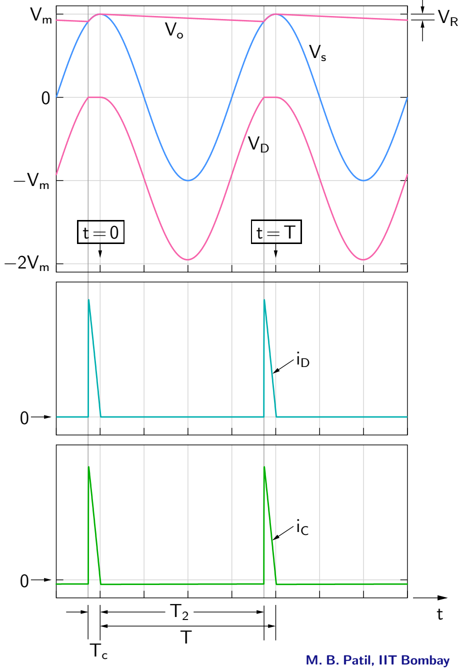
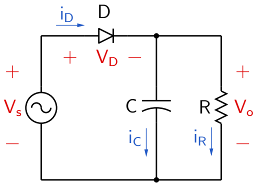
$$\omega T_c = \cos^{-1}\left(1-\frac{V_R}{V_m}\right) = \cos^{-1}\left(1-\frac{2}{16}\right) = 29^{\circ}.$$

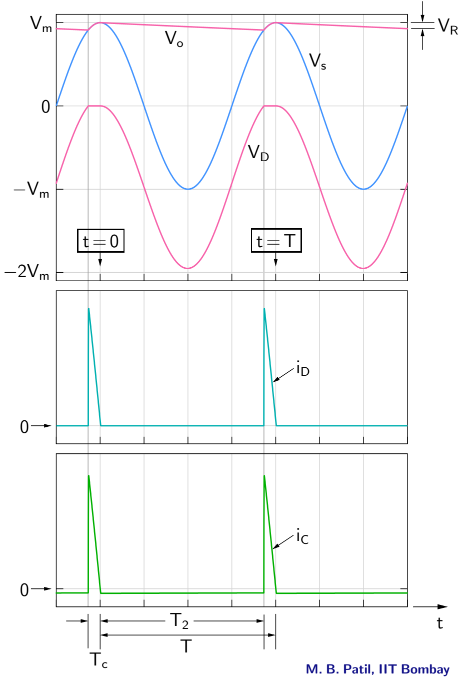
$$i_D^{\text{peak}} = 2\pi \times 50 \times 1600 \times 10^{-6} \times 16 \times \sin 29^{\circ} + 0.16$$

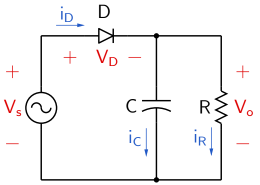
$$= 3.89 + 0.16 = 4.05\ \text{A}.$$





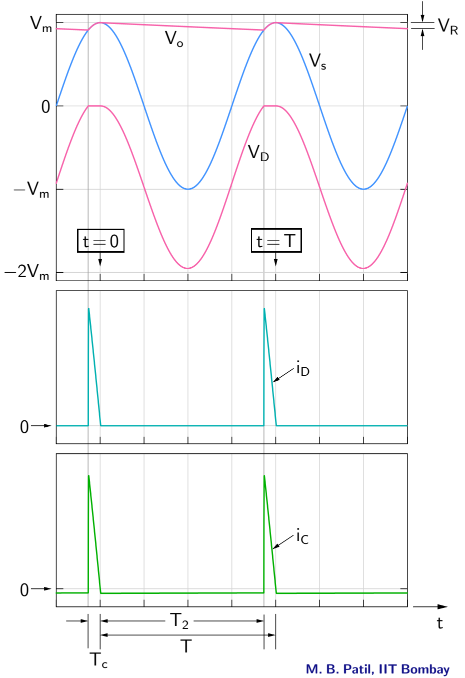
(b) Peak diode current: analytic expression





(b) Peak diode current: analytic expression

$$V_m \cos(-\omega T_c) = V_m - V_R \rightarrow \cos \omega T_c = 1 - \frac{V_R}{V_m} \equiv 1 - x$$



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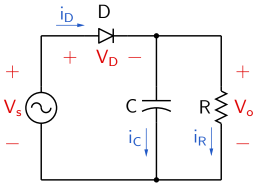
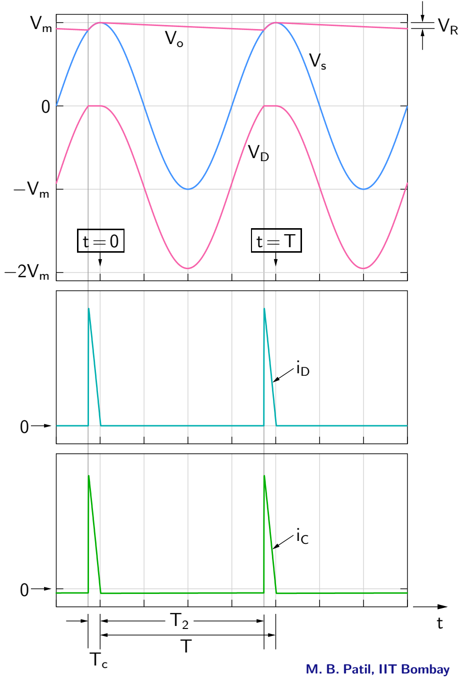


(b) Peak diode current: analytic expression

$$V_m \cos(-\omega T_c) = V_m - V_R \rightarrow \cos \omega T_c = 1 - \frac{V_R}{V_m} \equiv 1 - x$$

$$\sin \omega T_c = \sqrt{1 - \cos^2 \omega T_c} = \sqrt{1 - (1-x)^2}$$

$$= \sqrt{1 - (1 - 2x + x^2)} \approx \sqrt{2x} = \sqrt{\frac{2\,V_R}{V_m}}$$



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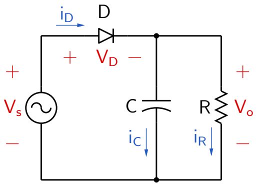

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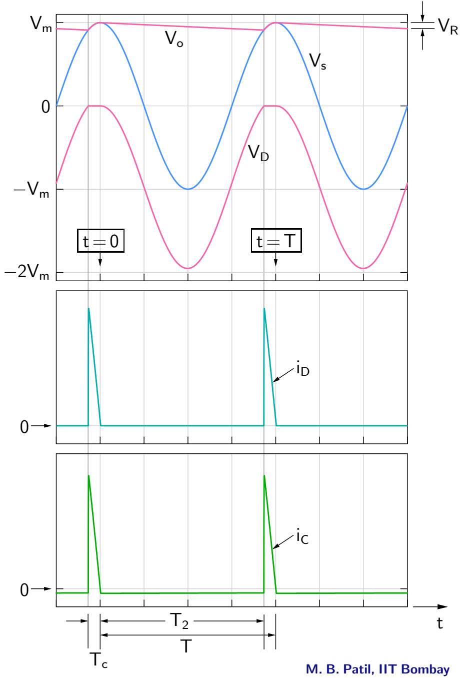

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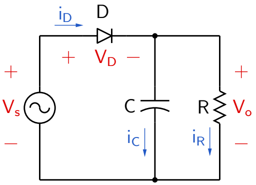

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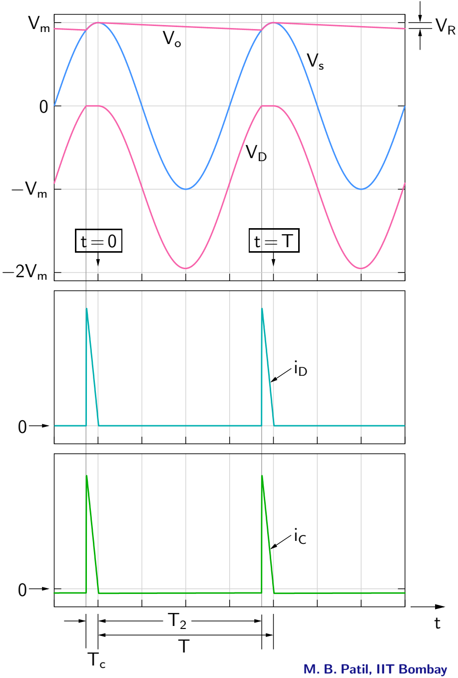
(c) Maximum reverse bias $\approx 2\,V_m = 32\,\text{V}$.



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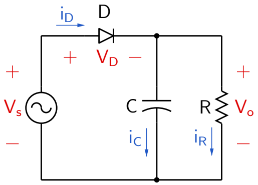

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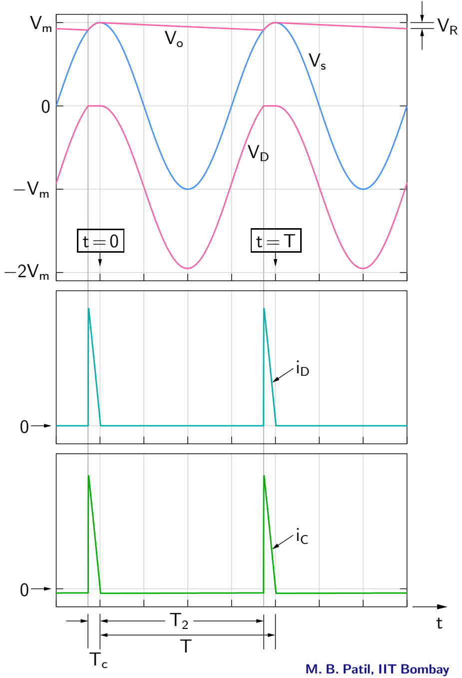
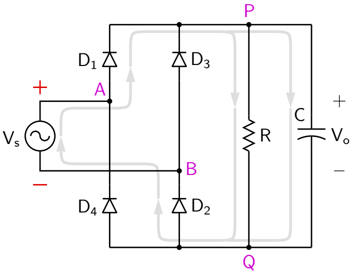
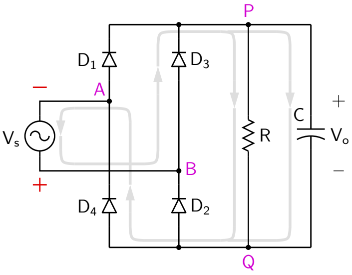
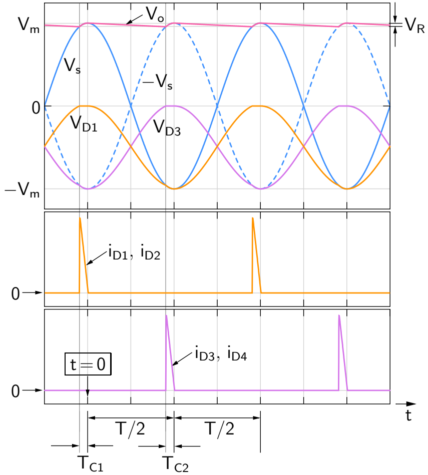
SEQUEL file: `ee101_half_rectifier.sqproj`



Full-wave (bridge) rectifier with capacitor filter







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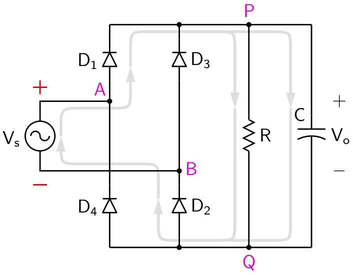
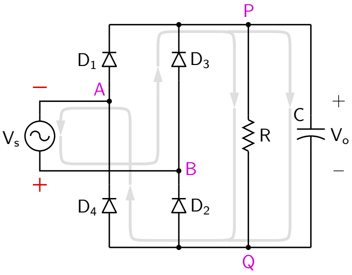
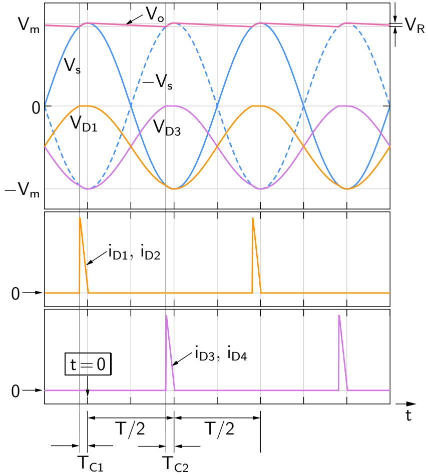
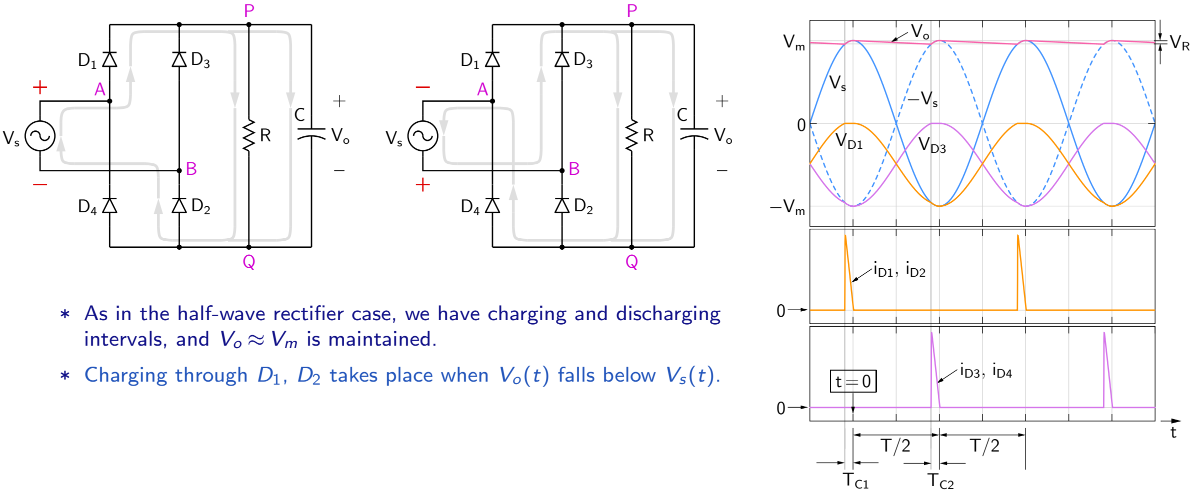






* As in the half-wave rectifier case, we have charging and discharging intervals, and $V_o \approx V_m$ is maintained.



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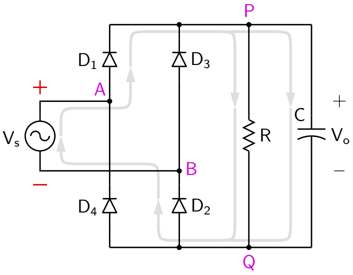

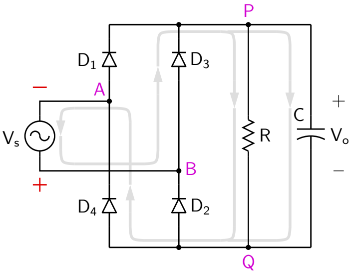

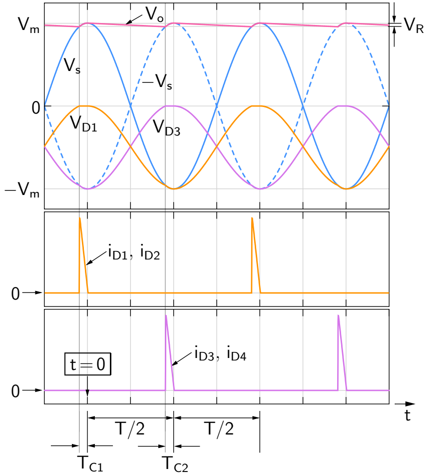

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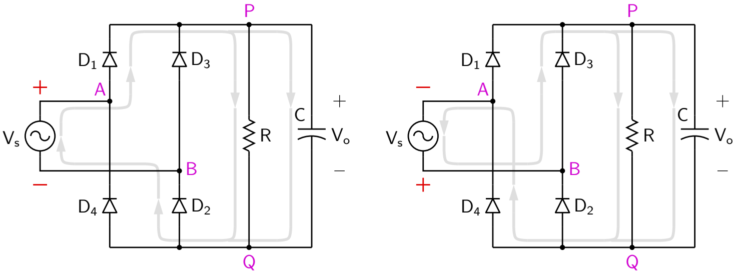
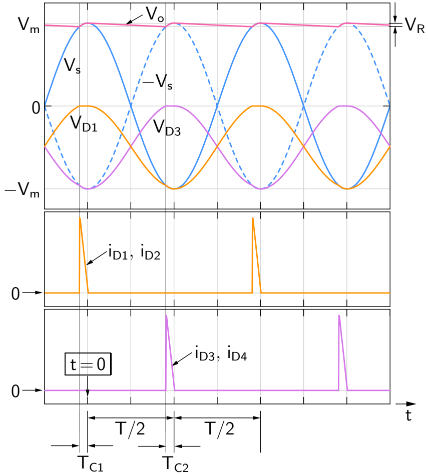
Full-wave (bridge) rectifier with capacitor filter





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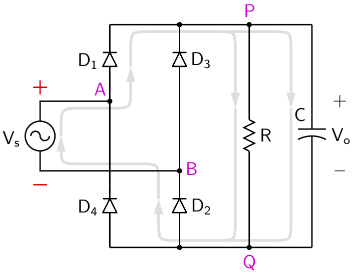
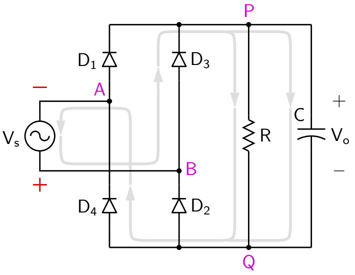
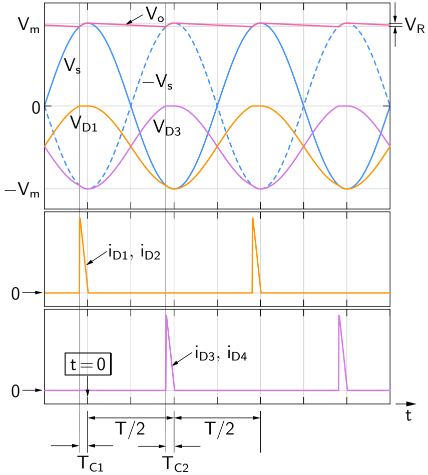
Full-wave (bridge) rectifier with capacitor filter







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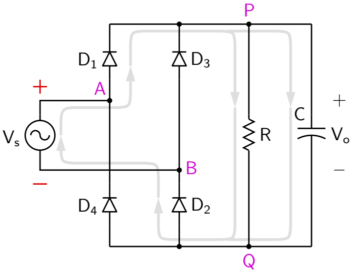
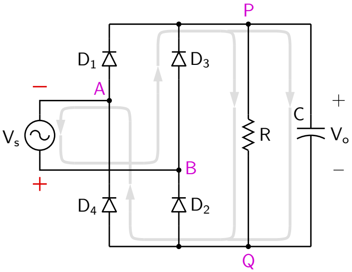
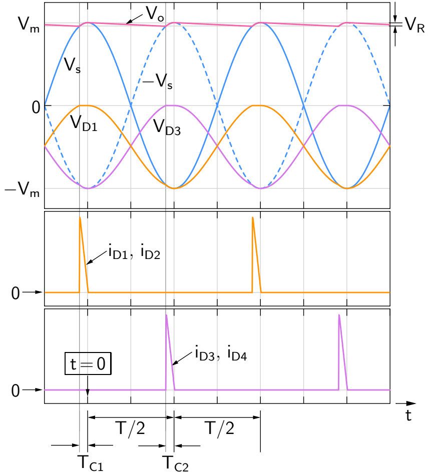
* The maximum reverse bias across any of the diodes is V_m.

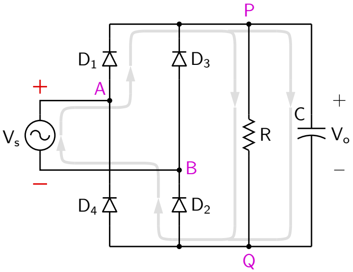
Full-wave rectifier with capacitor filter

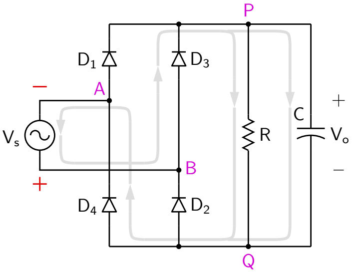


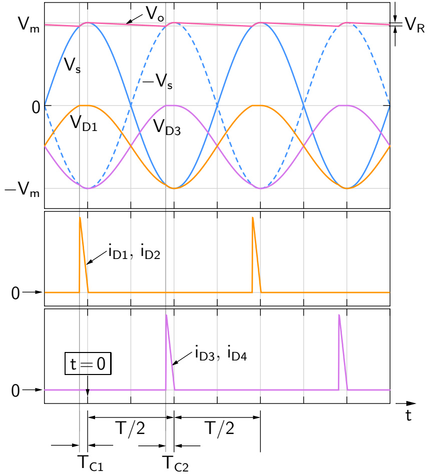




$V_m = 16\,\text{V}$, $f = 50\,\text{Hz}$, $R = 100\,\Omega$. For a ripple voltage $V_R = 2\,\text{V}$, find (a) the filter capacitance C, (b) average and peak diode currents, (c) maximum reverse voltage across the diode. (Let $V_{on} = 0\,\text{V}$.)







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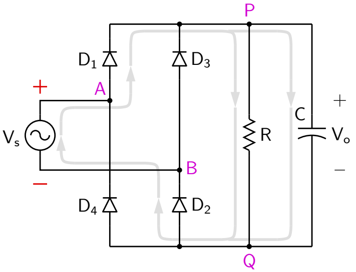

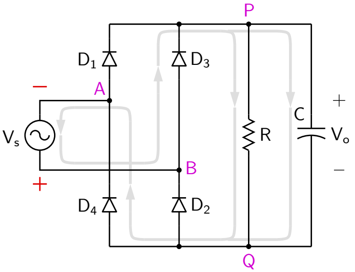

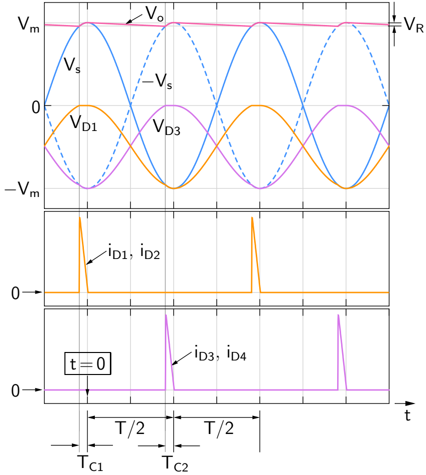

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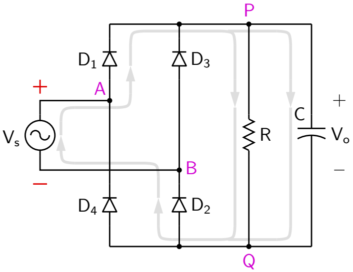
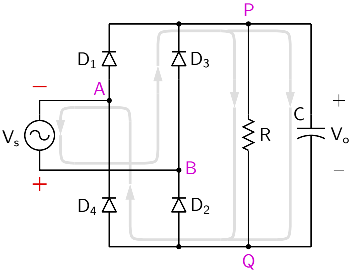
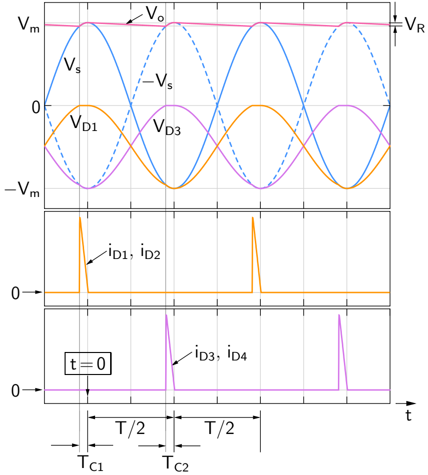
$$i_C = \frac{V_m}{R} = C\,\frac{\Delta V_o}{\Delta t} \approx C\,\frac{V_R}{T/2} \rightarrow V_R = V_m\,\frac{T}{2RC}.$$







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$$\rightarrow C = \frac{1}{2}\,\frac{V_m}{V_R}\,\frac{T}{R} = \frac{1}{2}\,\frac{16\,\text{V}}{2\,\text{V}}\,\frac{20\,\text{ms}}{100\,\Omega} = 800\,\mu\text{F}.$$

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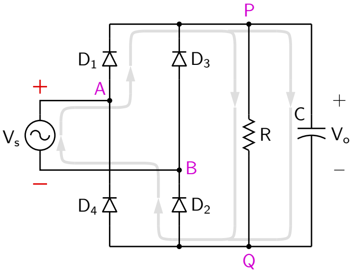
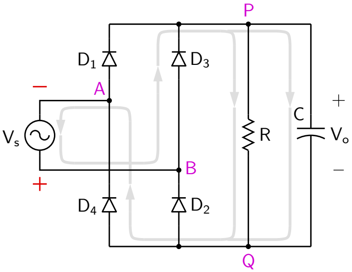
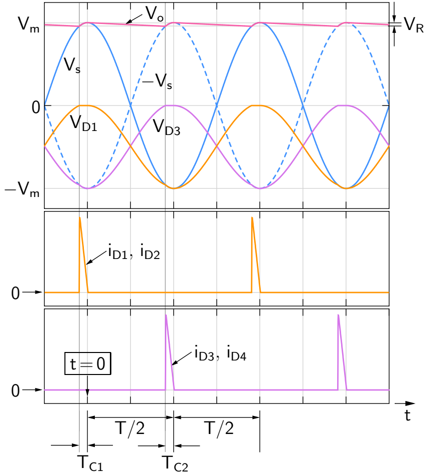
Full-wave rectifier with capacitor filter

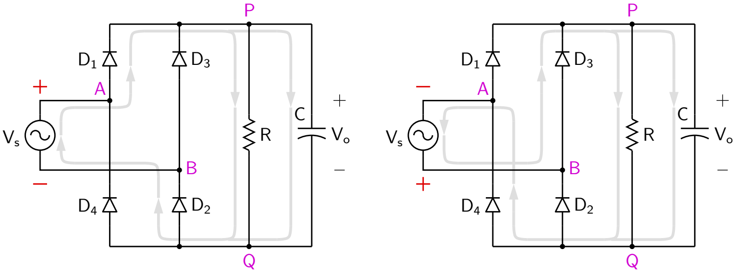


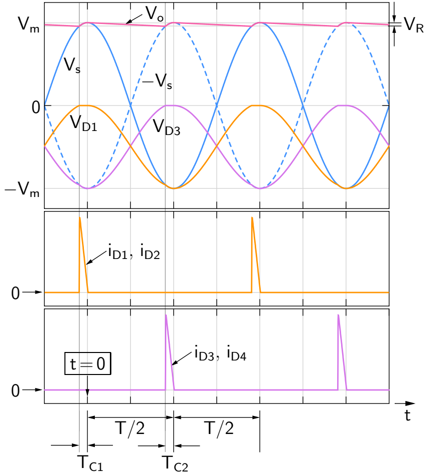




(b) Average diode current

Full-wave rectifier with capacitor filter

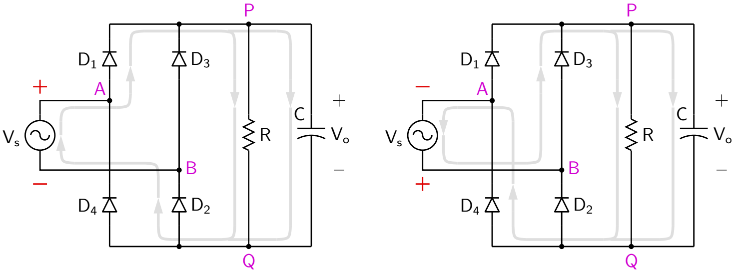


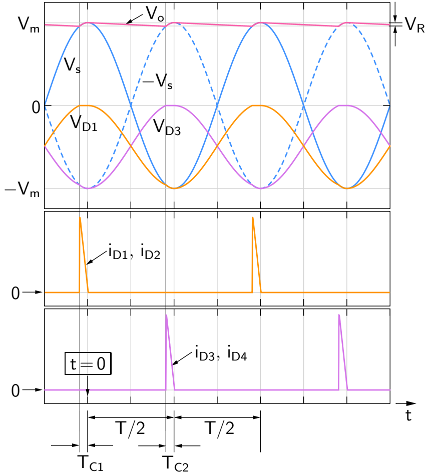


(b) Average diode current

Half of the charge lost by the capacitor is supplied by i_{D1} $(= i_{D2})$, and the other half by i_{D3} $(= i_{D4})$.

Full-wave rectifier with capacitor filter





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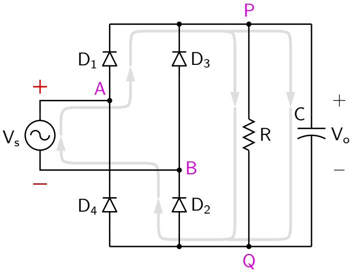
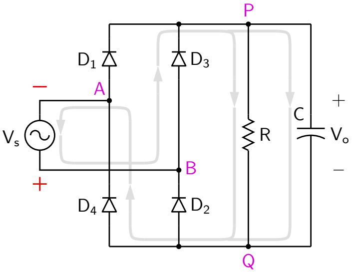
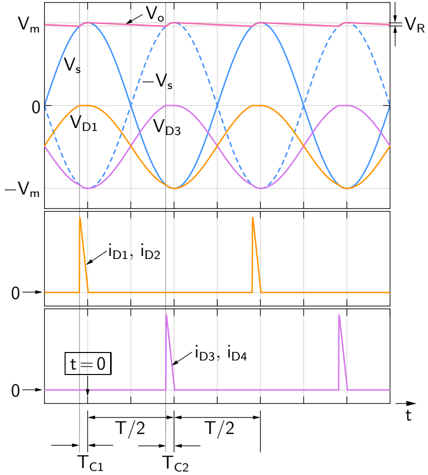
$$i_D^{av} = \frac{1}{T} \times \frac{1}{2} \times (\text{Charge lost in one cycle})$$







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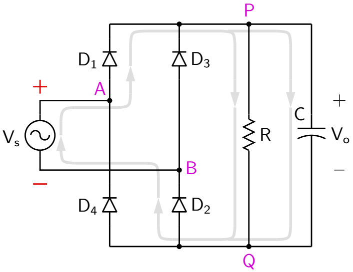
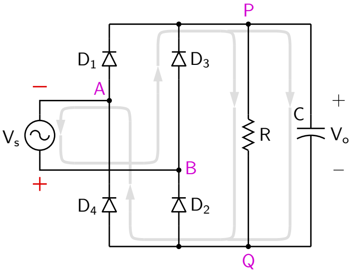
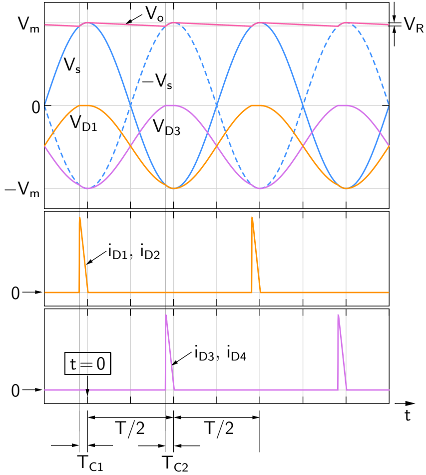
$$i_D^{av} = \frac{1}{T} \times \frac{1}{2} \times (\text{Charge lost in one cycle})$$

$$\approx \frac{1}{T} \times \frac{1}{2} \times \left(\frac{V_m}{R} \times T \right) = \frac{V_m}{2R} = \frac{16\,\text{V}}{2 \times 100\,\Omega} = 80\,\text{mA}.$$

Full-wave rectifier with capacitor filter

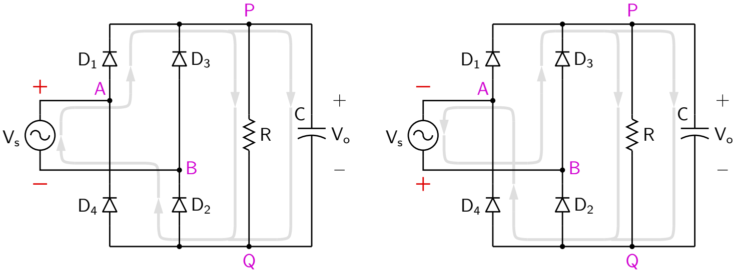


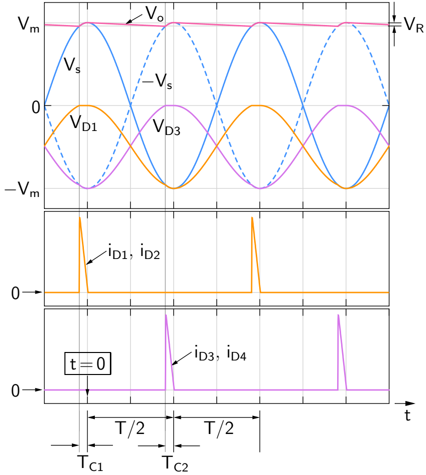




(b) Peak diode current

Full-wave rectifier with capacitor filter



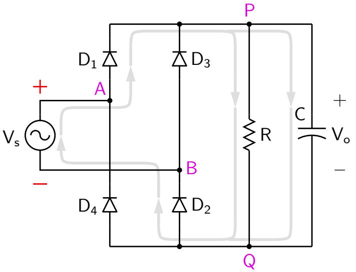


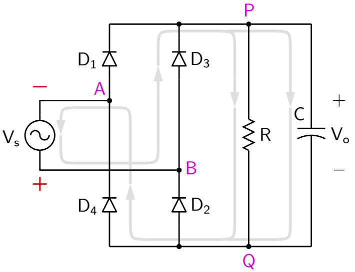
(b) Peak diode current

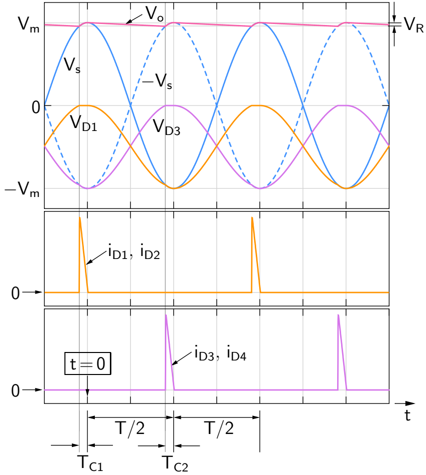
$$i_{D1}^{\text{peak}} = C\,\frac{d}{dt}\,(V_m \cos\omega t)\Big|_{t=-T_{C1}} + \frac{V_m}{R}$$

$$= -\omega C\, V_m \sin(-\omega T_{C1}) + \frac{16\,\text{V}}{100\,\Omega}$$

$$= \omega C\, V_m \sin\omega T_{C1} + 0.16$$







(b) Peak diode current

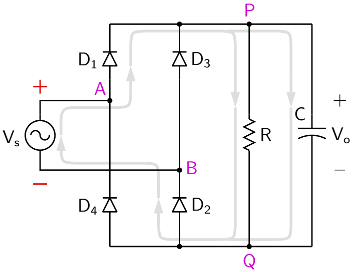
$$i_{D1}^{\text{peak}} = C\frac{d}{dt}\left(V_m \cos\omega t\right)\Big|_{t=-T_{C1}} + \frac{V_m}{R}$$

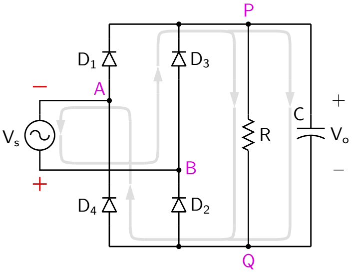
$$= -\omega C\, V_m \sin(-\omega T_{C1}) + \frac{16\,\text{V}}{100\,\Omega}$$

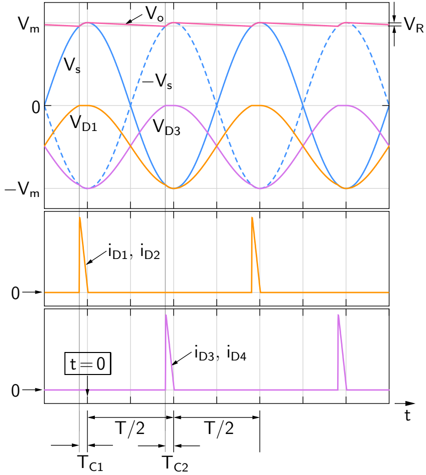
$$= \omega C\, V_m \sin\omega T_{C1} + 0.16$$

$$\omega T_{C1} = \cos^{-1}\left(1 - \frac{V_R}{V_m}\right) = \cos^{-1}\left(1 - \frac{2}{16}\right) = 29^\circ.$$

Full-wave rectifier with capacitor filter







(b) Peak diode current

$$i_{D1}^{\text{peak}} = C\,\frac{d}{dt}\left(V_m \cos\omega t\right)\Big|_{t=-T_{C1}} + \frac{V_m}{R}$$

$$= -\omega C\, V_m \sin(-\omega T_{C1}) + \frac{16\ \text{V}}{100\ \Omega}$$

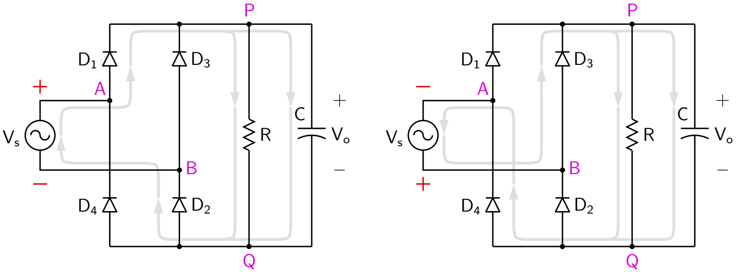
$$= \omega C\, V_m \sin\omega T_{C1} + 0.16$$

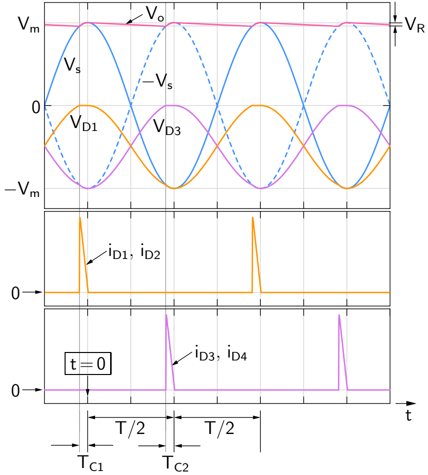
$$\omega T_{C1} = \cos^{-1}\left(1 - \frac{V_R}{V_m}\right) = \cos^{-1}\left(1 - \frac{2}{16}\right) = 29^\circ.$$

$$i_{D1}^{\text{peak}} = 2\pi \times 50 \times 800 \times 10^{-6} \times 16 \times \sin 29^\circ + 0.16$$

$$= 1.95 + 0.16 = 2.1\ \text{A}.$$

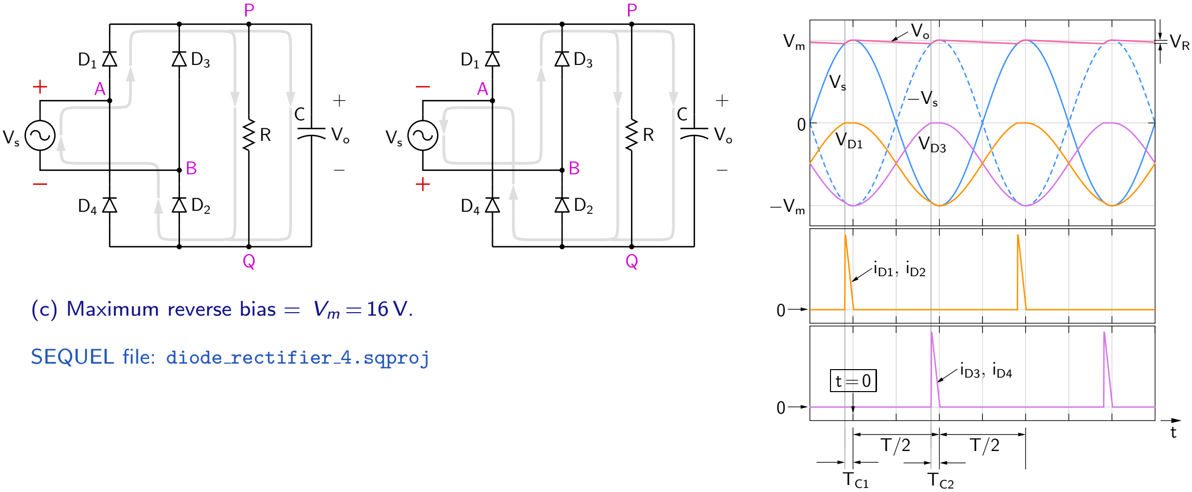
Full-wave rectifier with capacitor filter





(c) Maximum reverse bias $= V_m = 16\,\text{V}$.

Full-wave rectifier with capacitor filter



(c) Maximum reverse bias = $V_m = 16\,\text{V}$.

SEQUEL file: `diode_rectifier_4.sqproj`

For the same source voltage ($V_m \sin \omega t$), load (R), and ripple voltage (V_R), compare the half-wave and full-wave rectifiers.

Comparison of half-wave and full-wave (bridge) rectifiers with capacitive filter

For the same source voltage ($V_m \sin \omega t$), load (R), and ripple voltage (V_R), compare the half-wave and full-wave rectifiers.

Parameter	Half-wave	Full-wave
Number of diodes	1	4
Filter capacitance	C	$C/2$
Average diode current	i_D^{av}	$i_D^{\text{av}}/2$
Peak diode current	i_D^{peak}	$i_D^{\text{peak}}/2$
Maximum reverse voltage	$2\,V_m$	V_m