

# SEMICONDUCTOR DEVICES

## Bipolar Junction Transistors: Part 1

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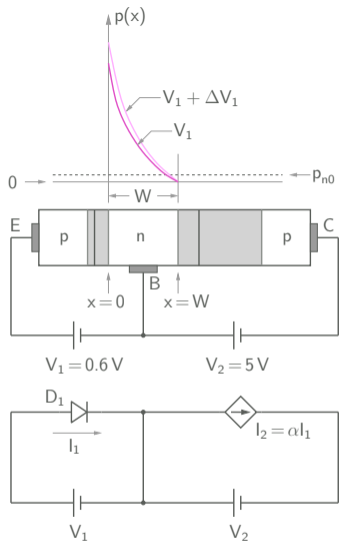
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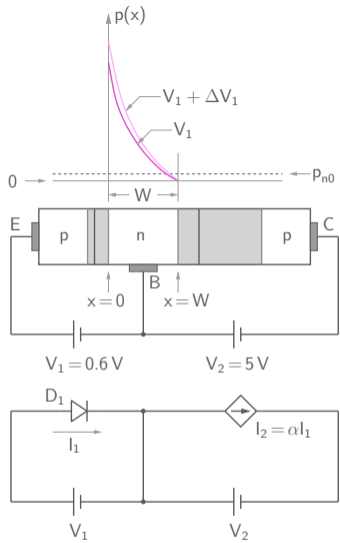
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- \* The actual device construction is different than the above schematic diagram (to be discussed).
- \* For the device to work as a transistor (rather than two independent diodes), the two junctions must be “close.”

Basic operation



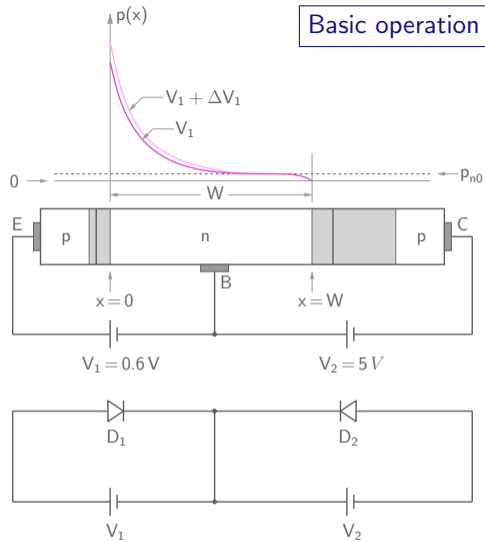
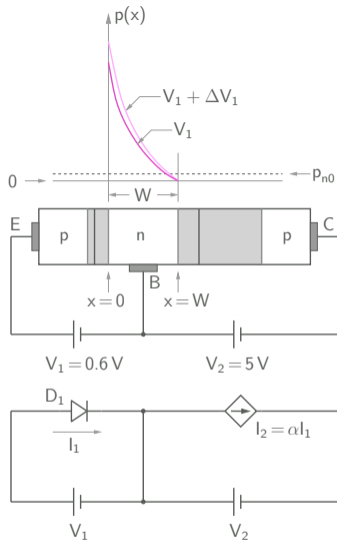


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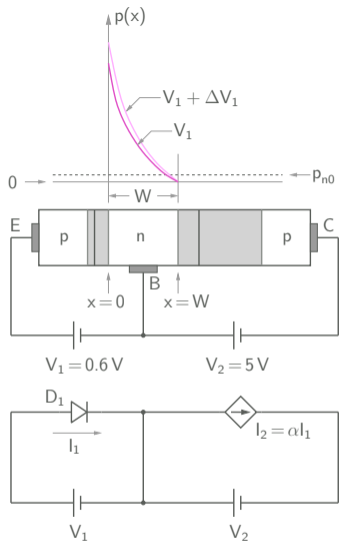


\* If  $V_1$  is varied,  $p(x)$  varies  $\rightarrow I_p(W) \propto \frac{dp}{dx}(W)$  varies, i.e., by changing  $V_1$ ,  $I_2$  can be controlled. This is the basic transistor action.

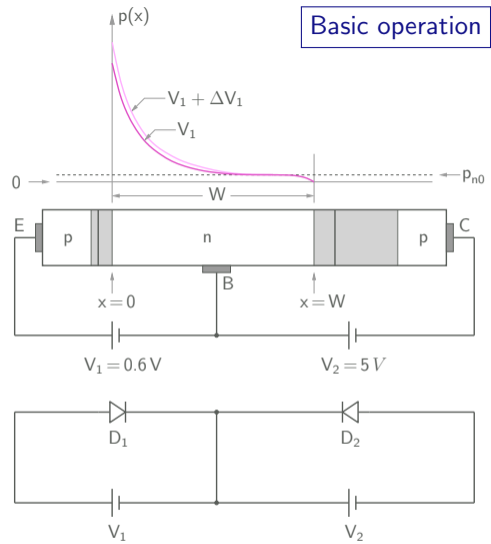
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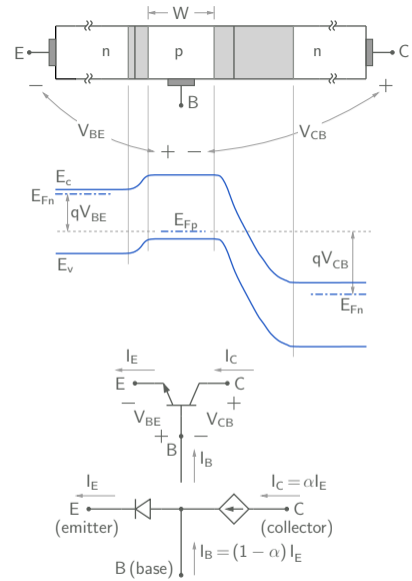
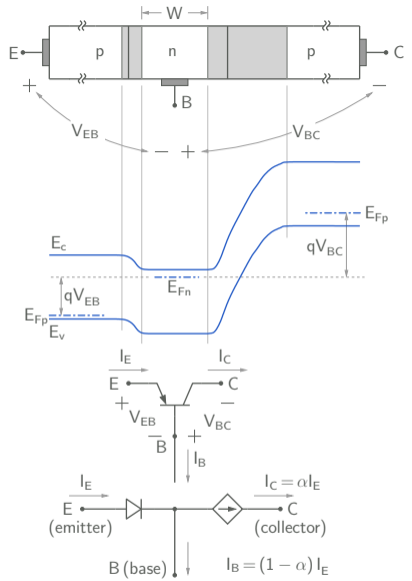


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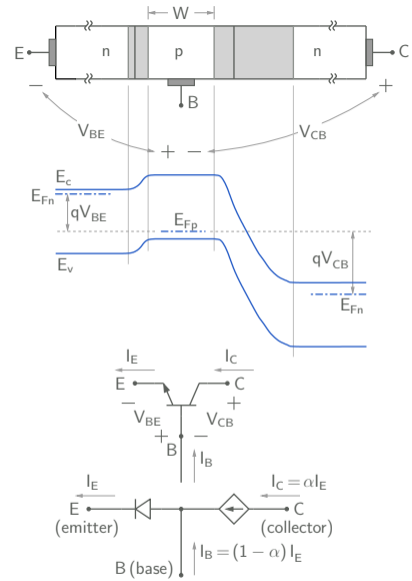
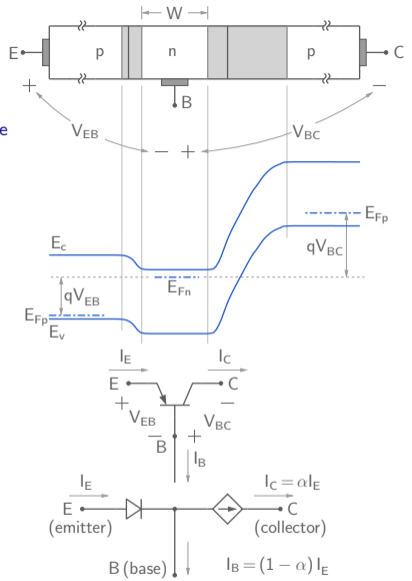
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- \* If the two junctions are not sufficiently close, the device behaves like two independent diodes connected back-to-back, and there is no transistor action.

*pnp and npn transistors*



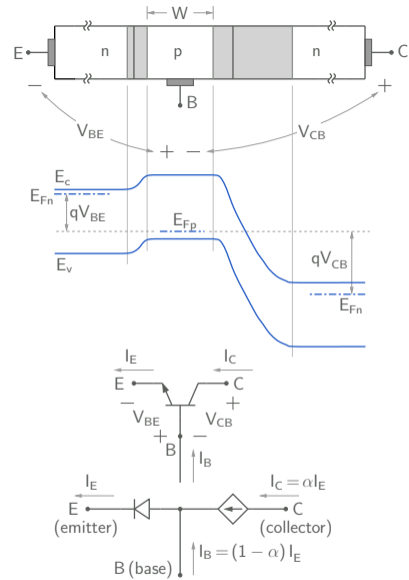
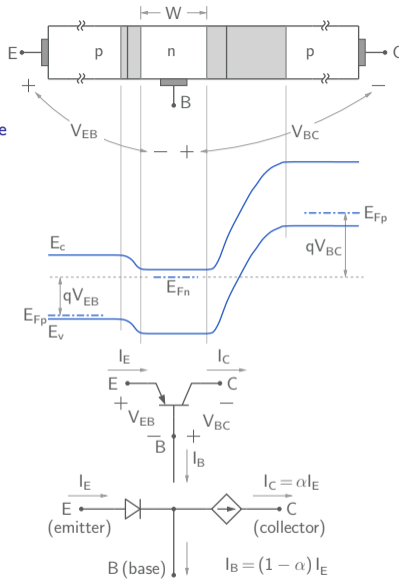
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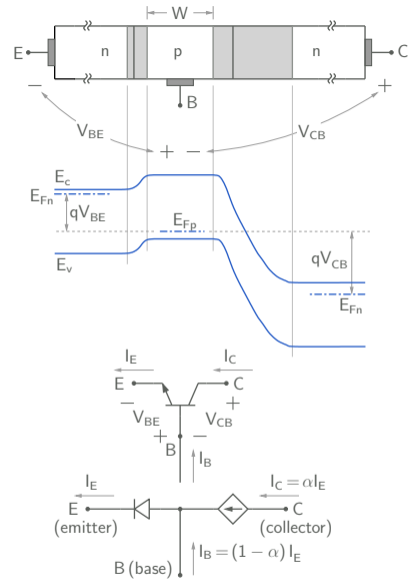
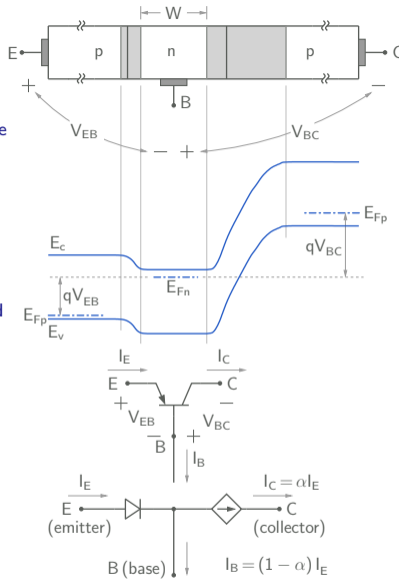
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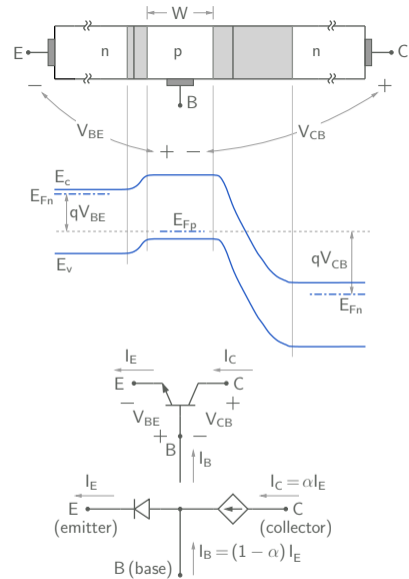
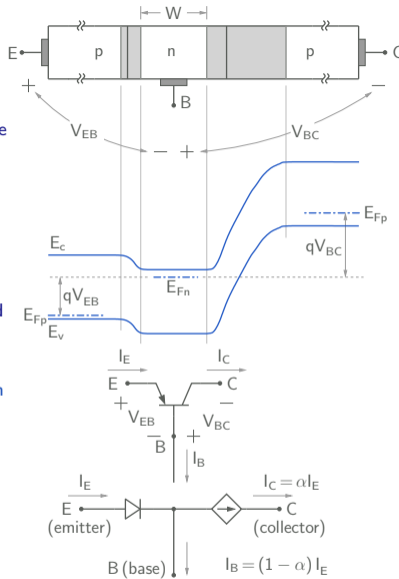
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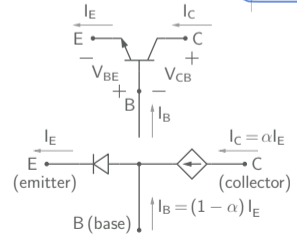
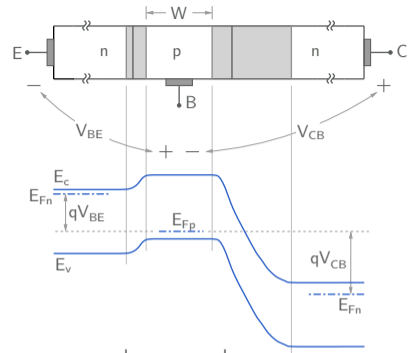
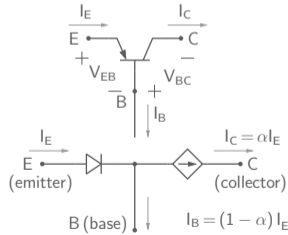
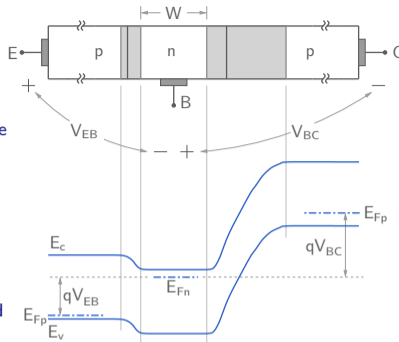
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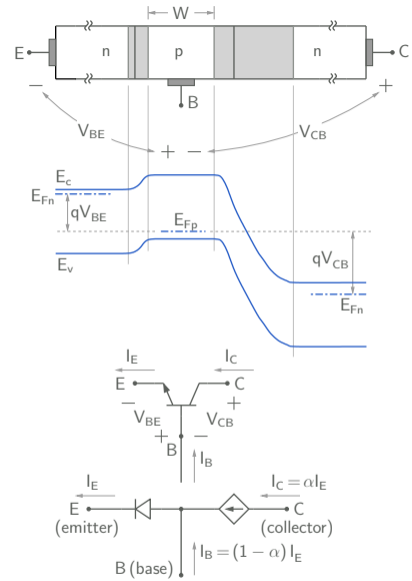
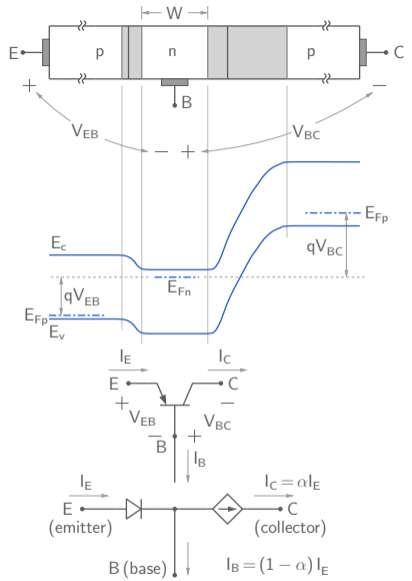


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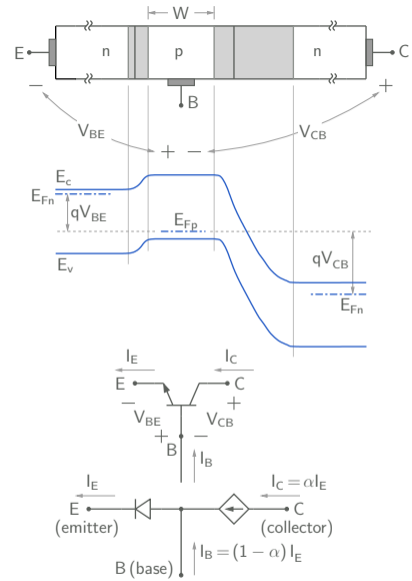
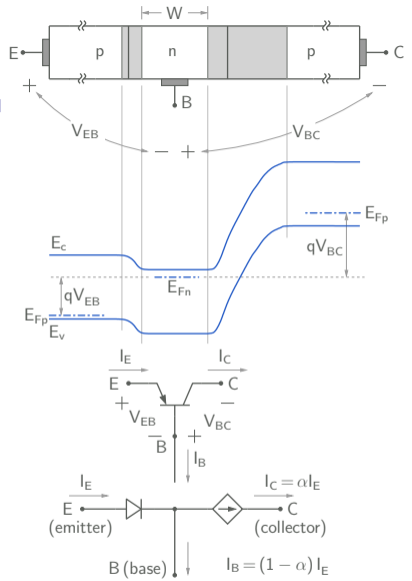


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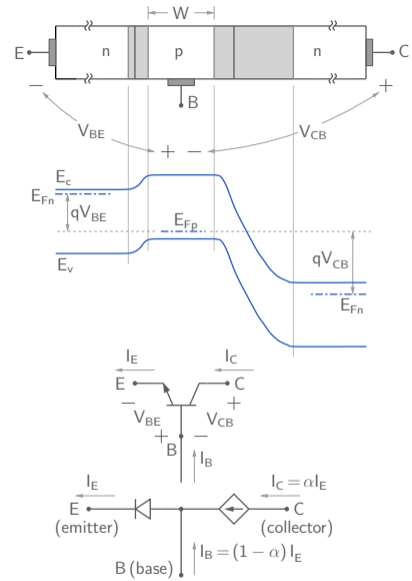
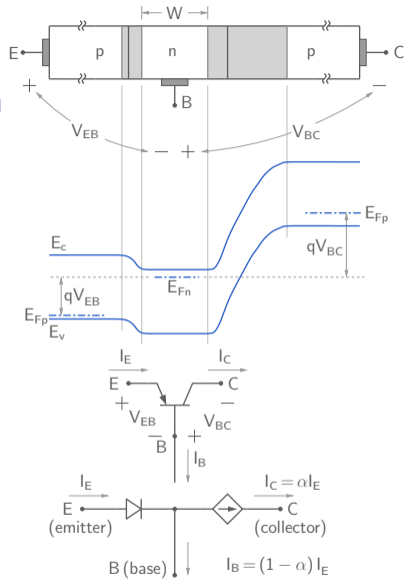
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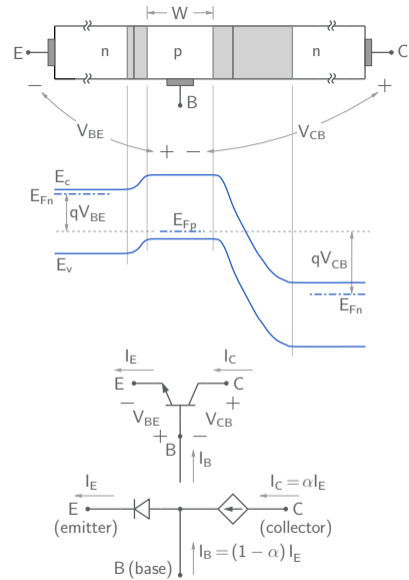
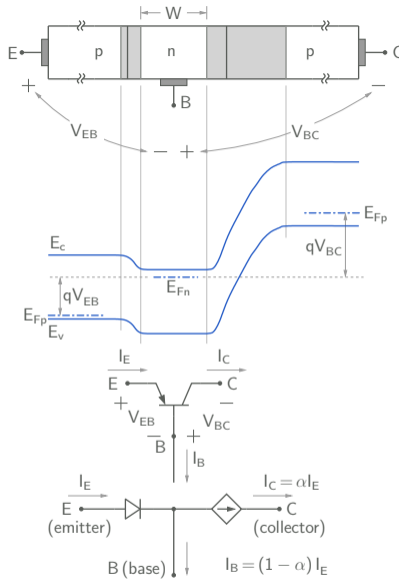
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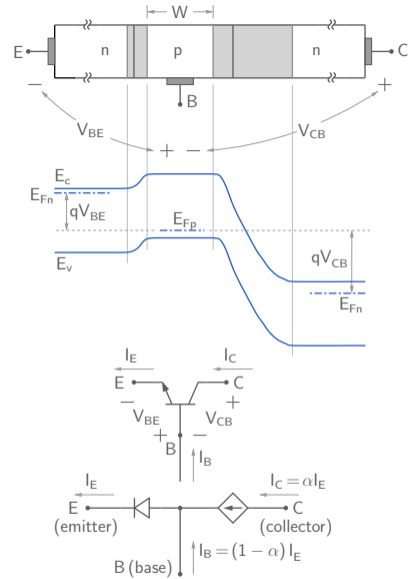
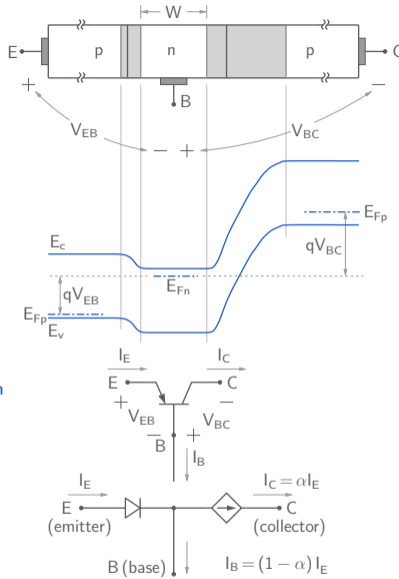
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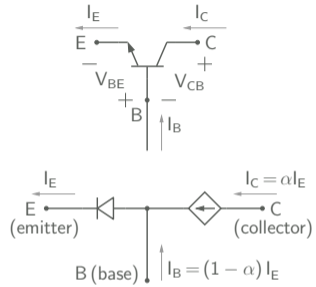
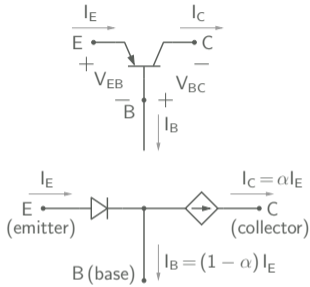


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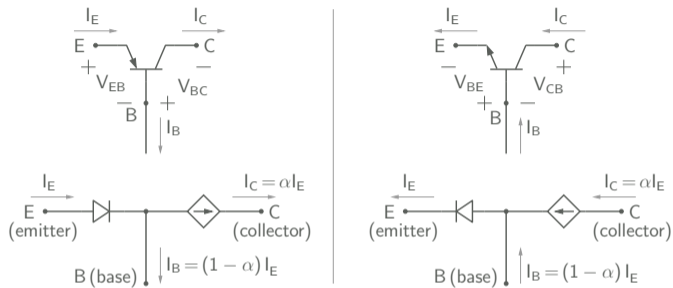
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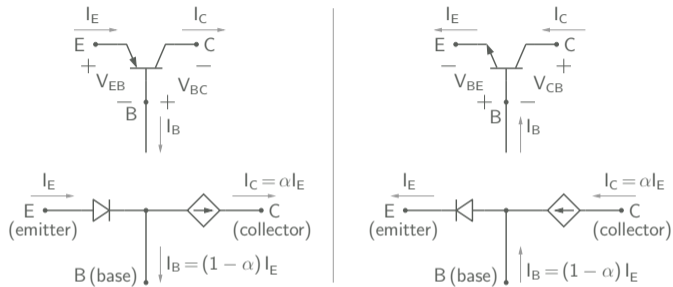
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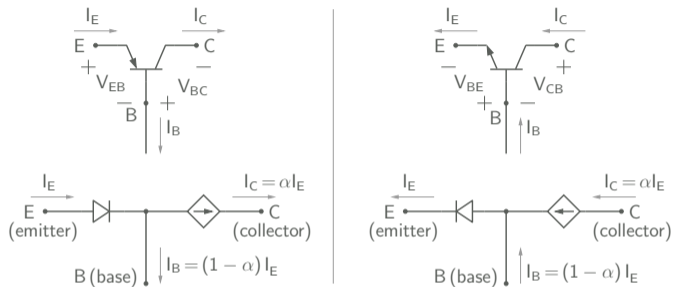
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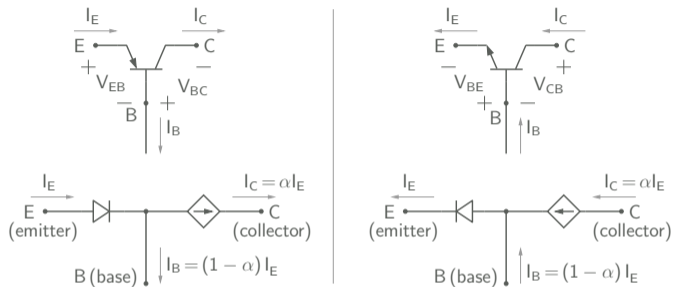


$\alpha$	$1 - \alpha$	$\beta = \alpha / (1 - \alpha)$
0.9	0.1	9
0.95	0.05	19
0.99	0.01	99
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$$\beta = \frac{I_C}{I_B} = \frac{\alpha I_E}{(1 - \alpha) I_E} = \frac{\alpha}{1 - \alpha}.$$
- \* For a typical discrete low-power transistor such as BC107A,  $\beta$  is in the range of 100 to 200.

# BJT modes of operation



Mode	<i>B-E</i> junction	<i>B-C</i> junction
Active (linear)	forward	reverse
Cutoff	reverse	reverse
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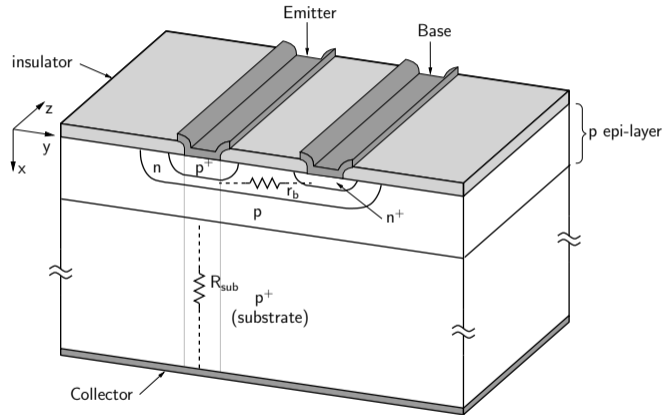
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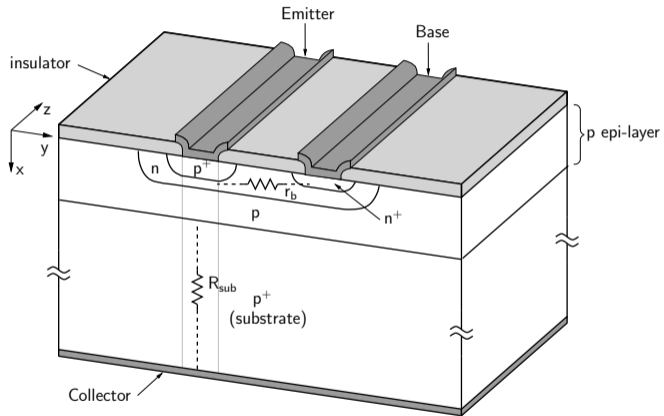
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- \* BJT as a switch:
  - Closed: saturation mode
  - Open: cutoff mode

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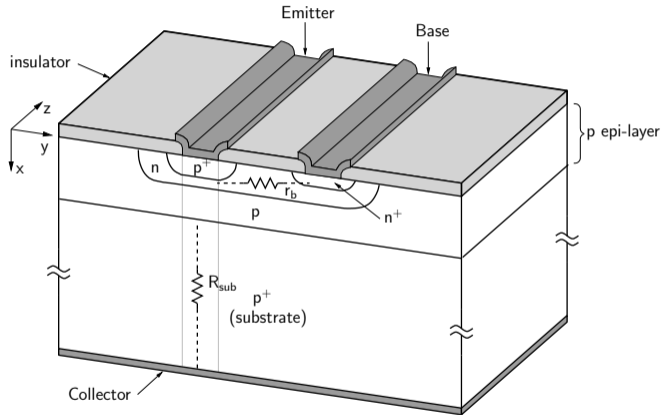
- \* The substrate thickness is hundreds of microns whereas the *p* epi-layer and the rest of the device structure is confined to a few microns.





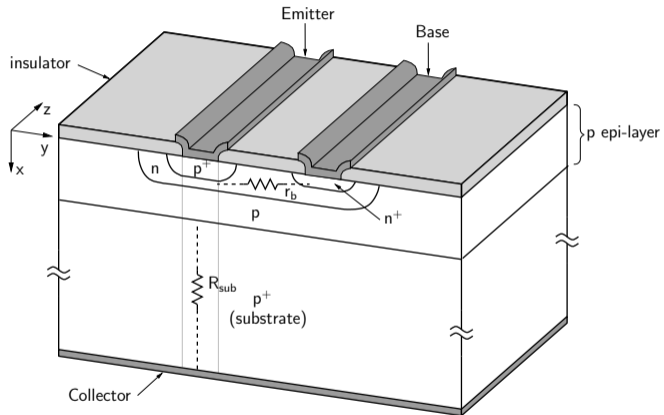
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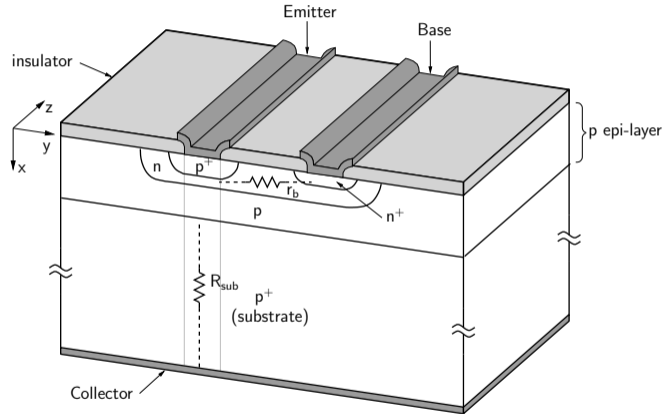
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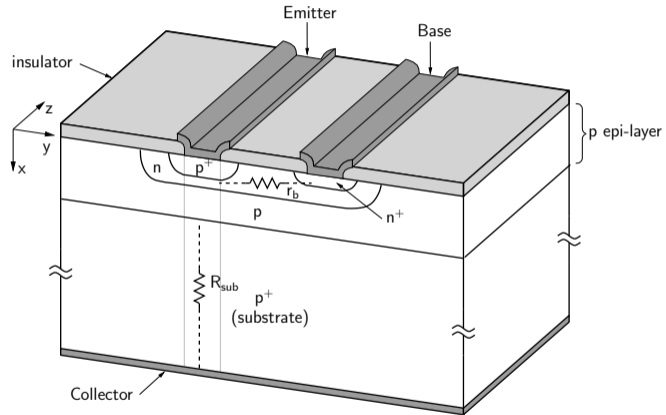
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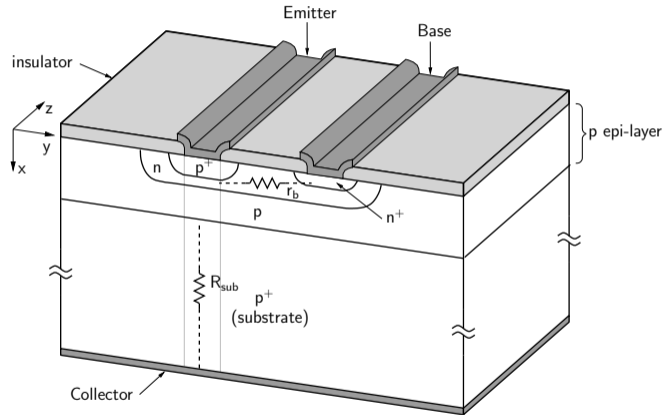
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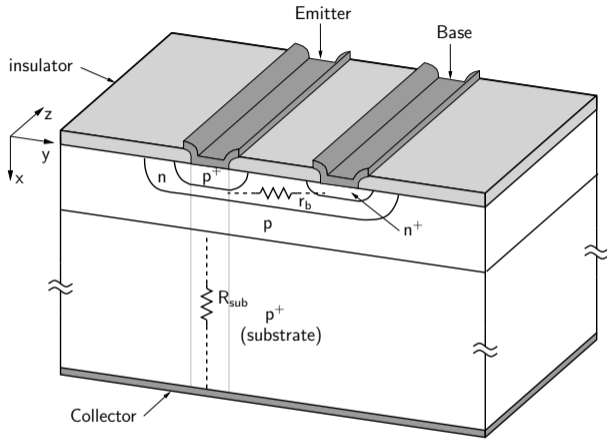


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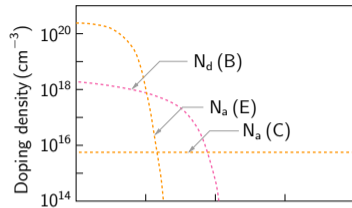
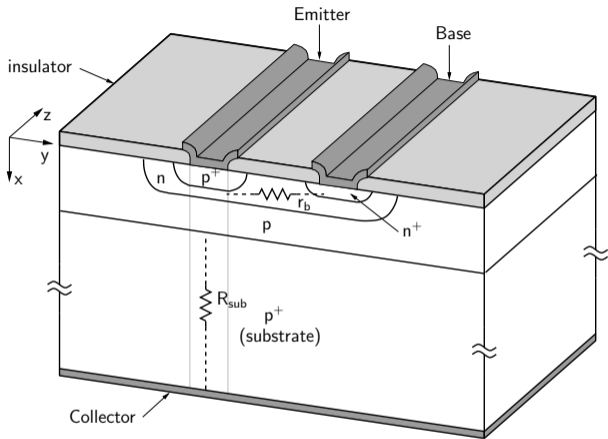
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- \* A “base resistance”  $r_b$  exists between the base region and the base contact. To keep  $r_b$  small, the base contact is made close to the emitter.



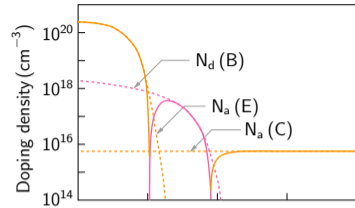
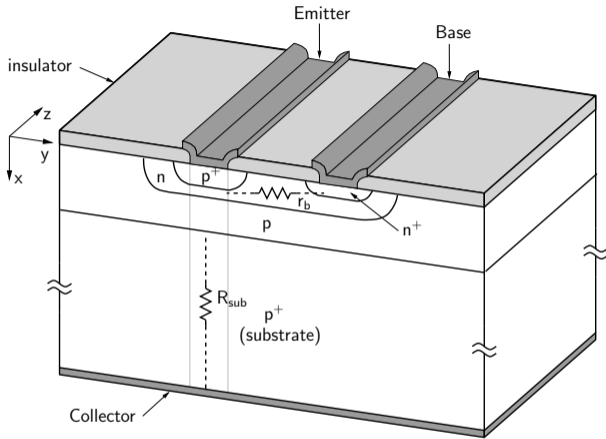
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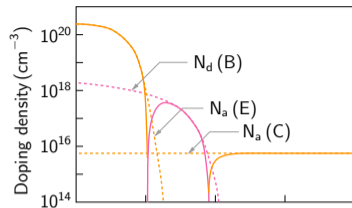
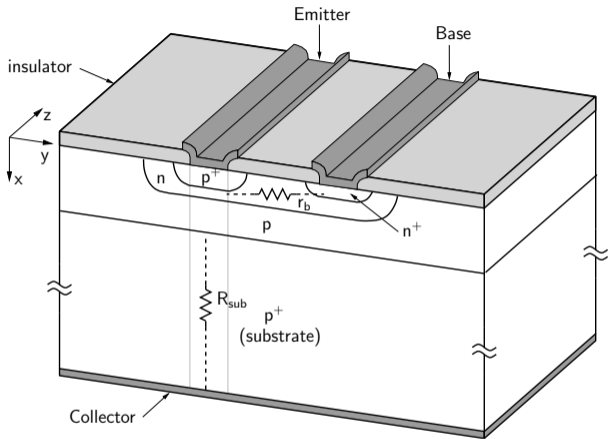


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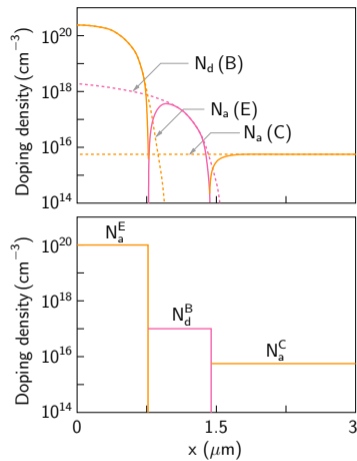
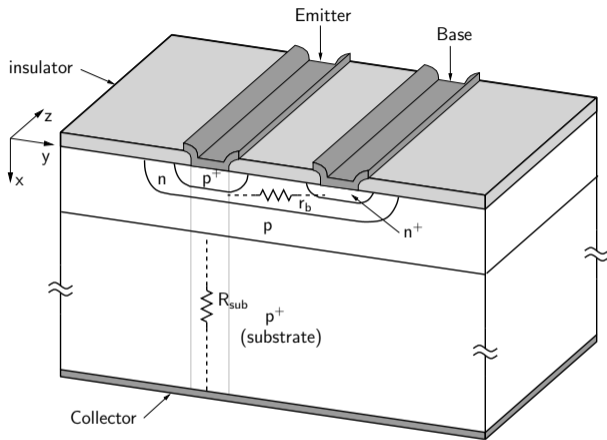


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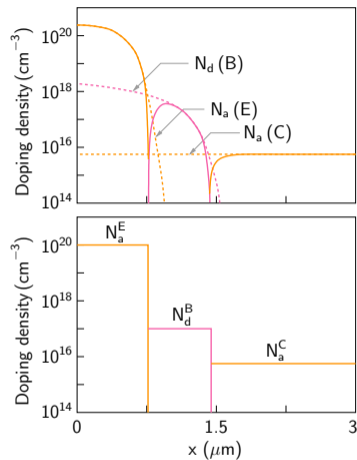
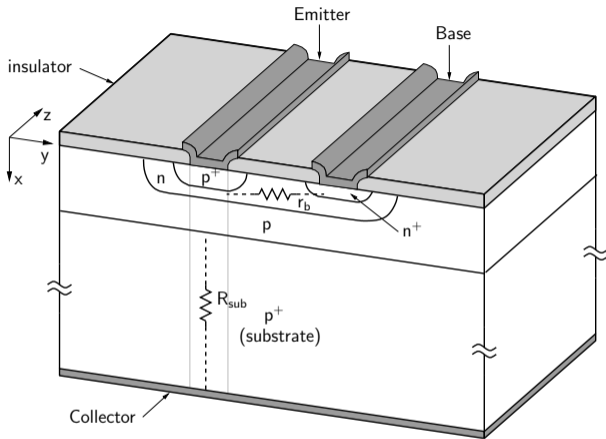
\* For simplicity, we will assume the doping densities to be constant in the emitter, base, and collector regions.

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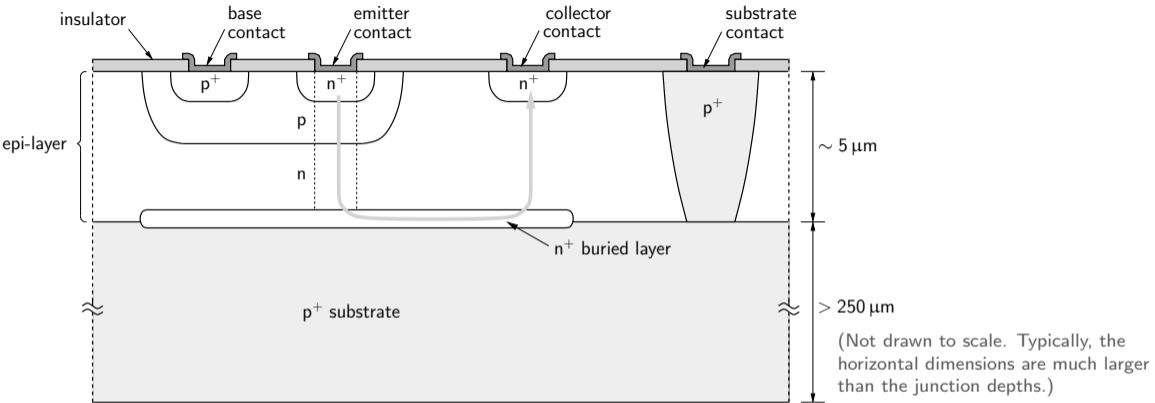
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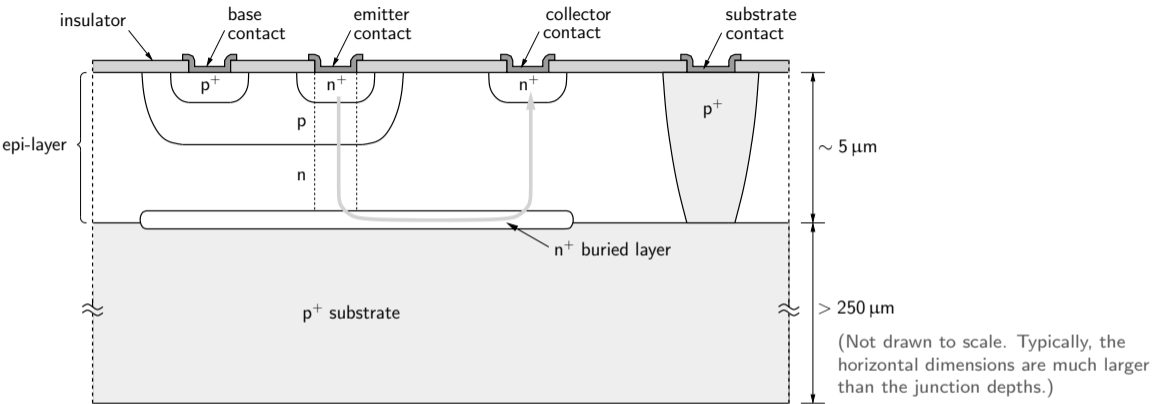


- \* For simplicity, we will assume the doping densities to be constant in the emitter, base, and collector regions.
- \* The relationship  $N_a^E > N_d^B > N_a^C$ , which is a consequence of the fabrication process, is also desirable from the device performance angle.

# BJT structure in integrated circuits (*npn* transistor)

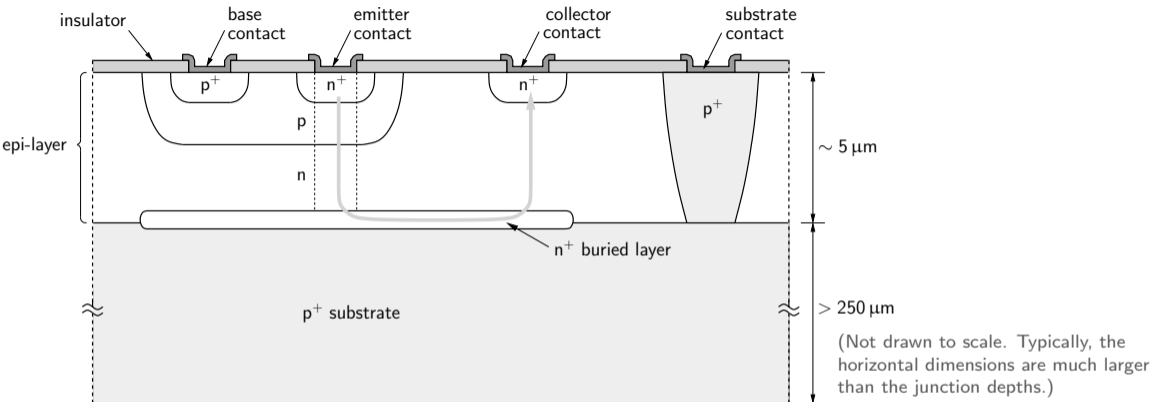


## BJT structure in integrated circuits (*npn* transistor)



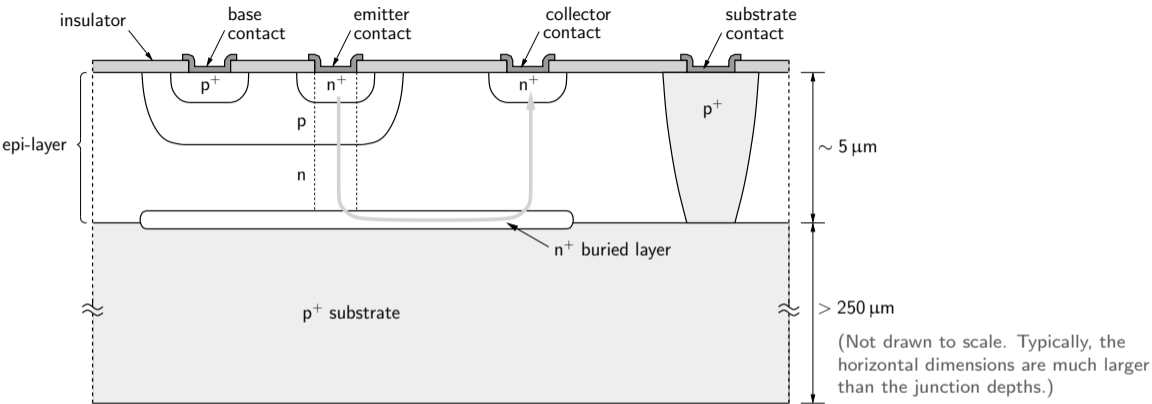
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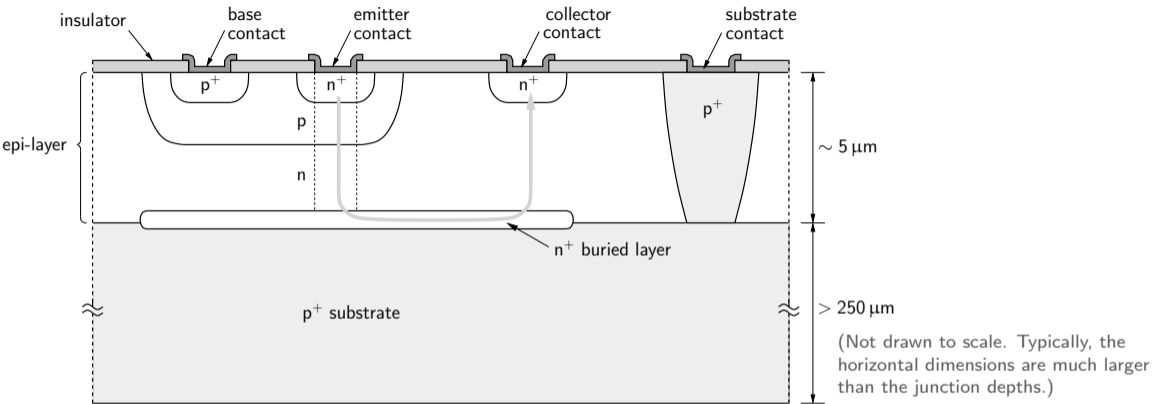
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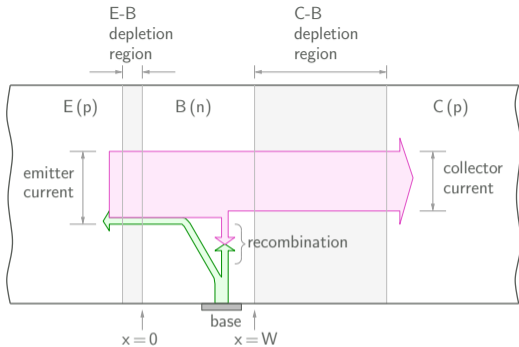
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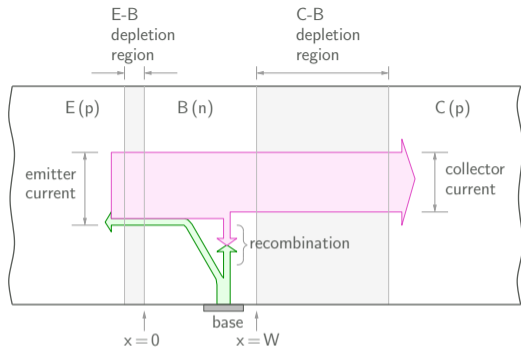
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- \* An  $n^+$  buried layer is used to provide a low-resistance path for the electron current.



# Dependence of $\alpha$ on device parameters



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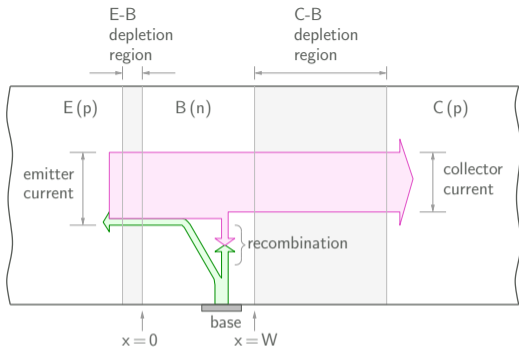


Consider a *pnp* transistor.

\*  $I_E$  has a hole component and an electron component. Of these, only the hole component contributes to  $I_C$ .

We define "emitter injection efficiency" (or simply "injection efficiency") as  $\gamma = \frac{I_{pE}}{I_E} = \frac{I_{pE}}{I_{pE} + I_{nE}}$ .

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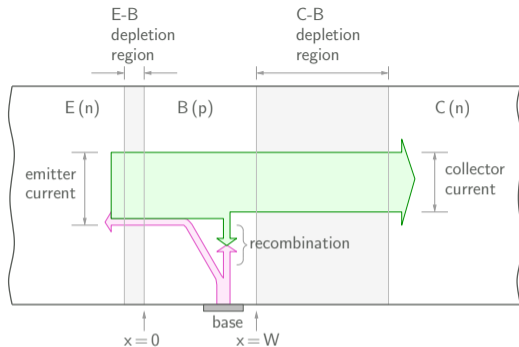
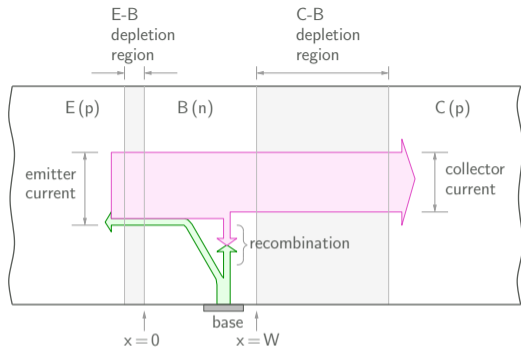
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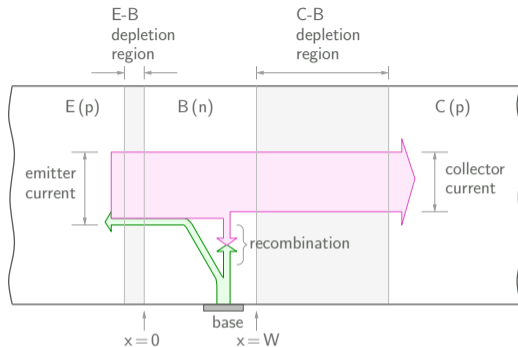
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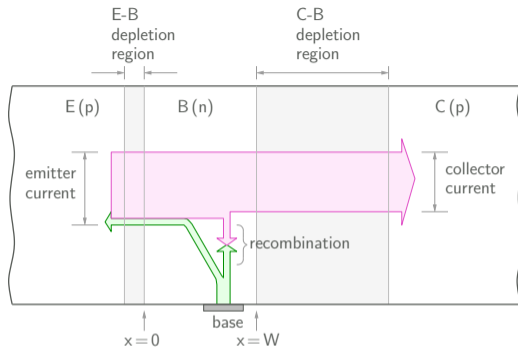
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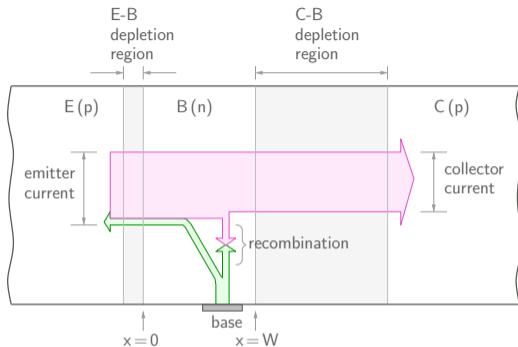


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→  $I_C$  is entirely due to the holes injected by the emitter which make it to the C-B depletion boundary ( $x = W$ ), i.e.,

$$I_C \approx I_{pC} = \alpha_T I_{pE} = \alpha_T (\gamma I_E) \rightarrow \alpha = \frac{I_C}{I_E} = \gamma \alpha_T.$$

## Dependence of $\alpha$ on device parameters

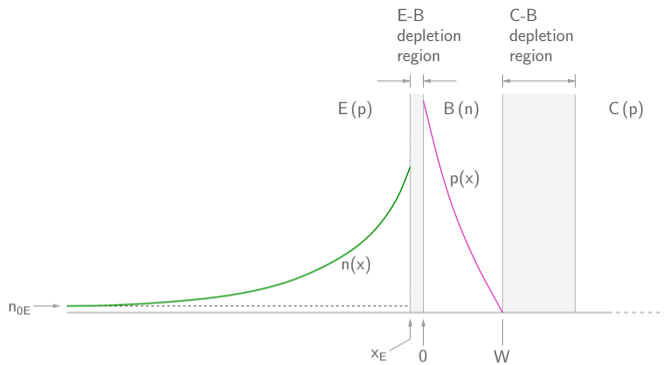


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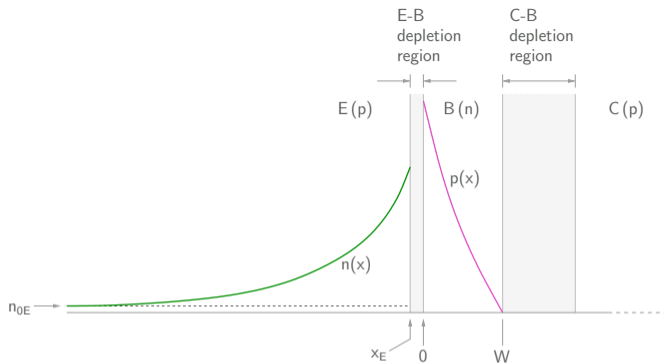
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→ For  $\alpha \approx 1$ , both  $\gamma$  and  $\alpha_T$  must be close to 1.



We assume that the emitter width is greater than  $5 L_n$ .





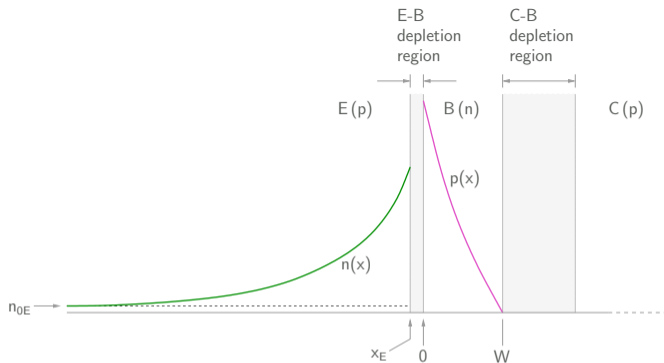
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Neglecting the drift components for minority carriers in the emitter and base neutral regions, we get

$$D_{nE} \frac{d^2 \Delta n}{dx^2} - \frac{\Delta n}{\tau_{nE}} = 0, \quad x < x_E, \text{ with}$$

$$\Delta n(x_E) = n_{0E} \left[ \exp\left(\frac{V_{EB}}{V_T}\right) - 1 \right],$$

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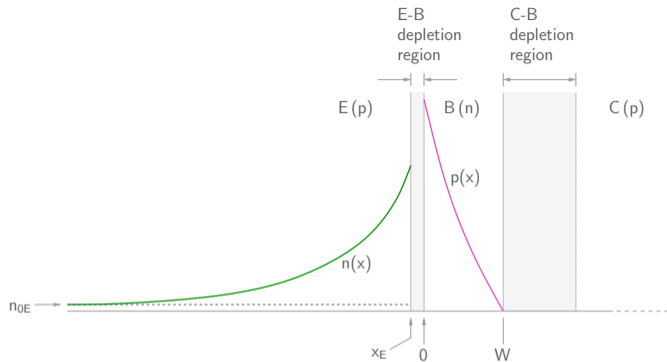


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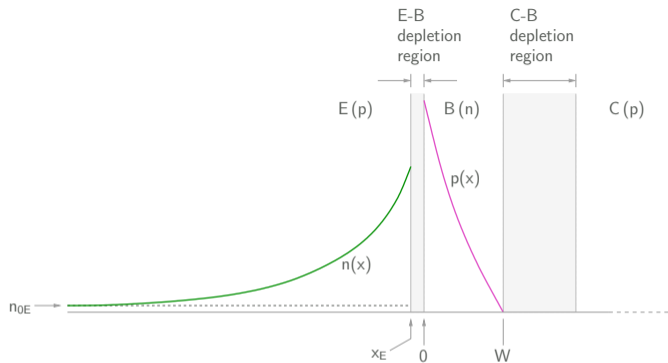
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 \end{array}
 \quad \left| \quad \begin{array}{l}
 D_{pB} \frac{d^2 \Delta p}{dx^2} - \frac{\Delta p}{\tau_{pB}} = 0, \quad 0 < x < W, \text{ with} \\
 \Delta p(0) = p_{0B} \left[ \exp\left(\frac{V_{EB}}{V_T}\right) - 1 \right], \\
 \Delta p(W) = p_{0B} \left[ \exp\left(\frac{V_{CB}}{V_T}\right) - 1 \right].
 \end{array}
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# Dependence of $\alpha$ on device parameters



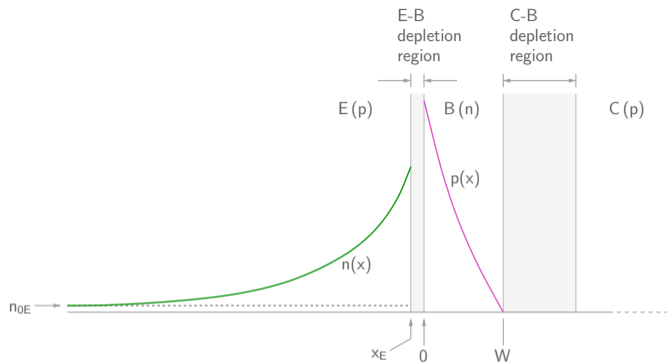
## Dependence of $\alpha$ on device parameters



Solution:

$$\Delta n(x) = \Delta n(x_E) e^{-(x_E - x)/L_{nE}}, \quad x < x_E,$$
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## Dependence of $\alpha$ on device parameters

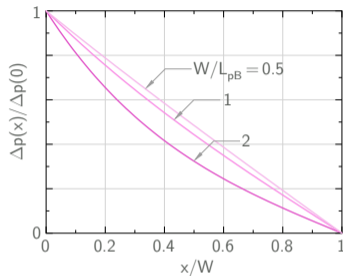
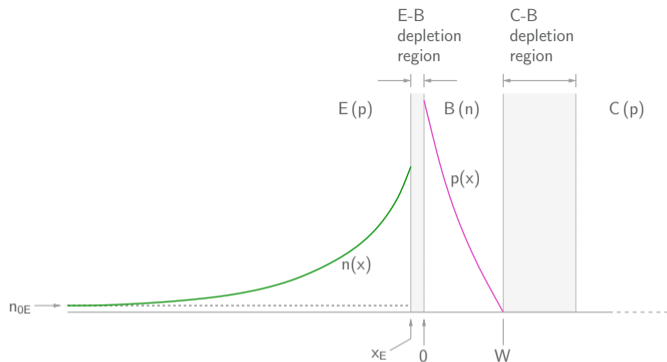


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Using the boundary conditions (last slide), we get

$$\Delta p(x) = \Delta p(0) \frac{\sinh\left(\frac{W-x}{L_{pB}}\right)}{\sinh\left(\frac{W}{L_{pB}}\right)} + \Delta p(W) \frac{\sinh\left(\frac{x}{L_{pB}}\right)}{\sinh\left(\frac{W}{L_{pB}}\right)}.$$

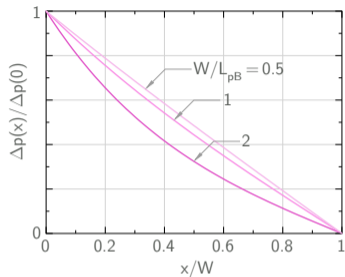
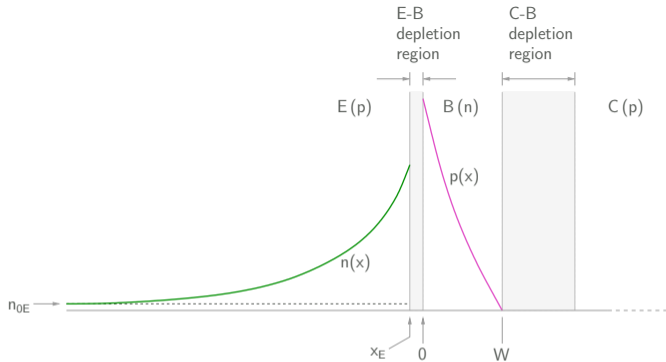
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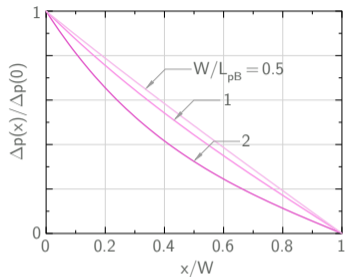
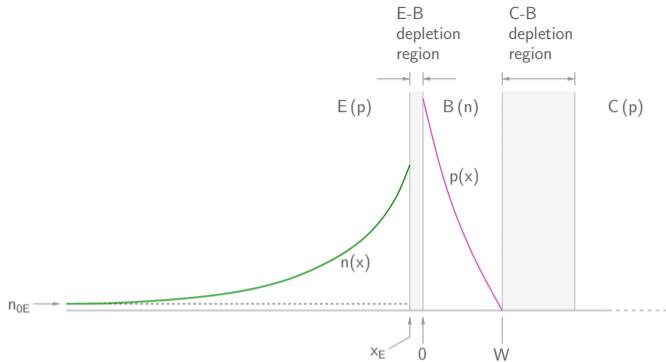
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$$\begin{aligned}
 I_{nE} &= qAD_{nE} \frac{dn}{dx}(x_E) = qAD_{nE} \frac{d\Delta n}{dx}(x_E) \\
 &= \frac{qAD_{nE}}{L_{nE}} \Delta n(x_E) = \frac{qAD_{nE}}{L_{nE}} n_{0E} \left( e^{V_{EB}/V_T} - 1 \right),
 \end{aligned}$$



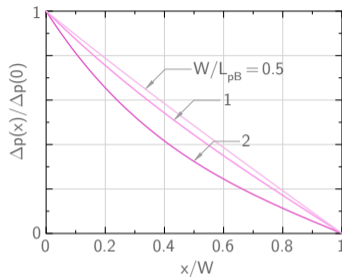
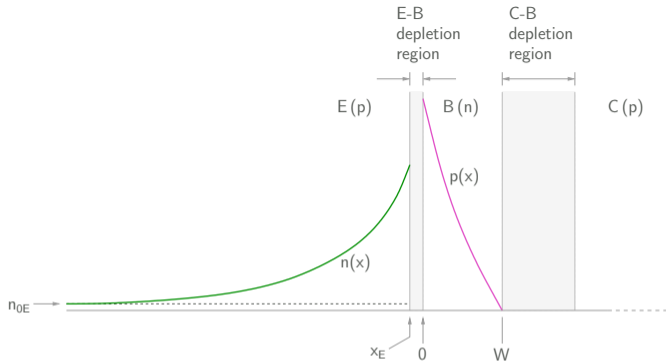
$$I_{nE} = qAD_{nE} \frac{dn}{dx}(x_E) = qAD_{nE} \frac{d\Delta n}{dx}(x_E)$$

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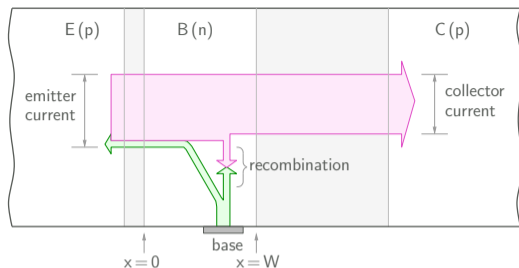
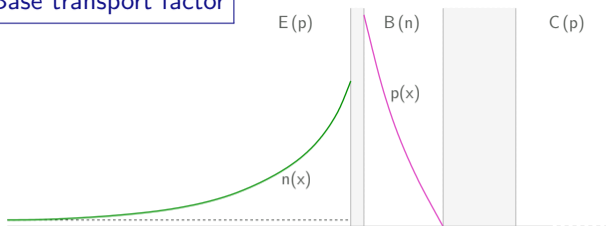
$$= \frac{qAD_{pB}}{L_{pB}} p_{0B} \left( e^{V_{EB}/V_T} - 1 \right) \frac{\cosh(W/L_{pB})}{\sinh(W/L_{pB})}.$$

$$\gamma = \frac{I_{pE}}{I_{pE} + I_{nE}} = \frac{1}{1 + (I_{nE}/I_{pE})}$$

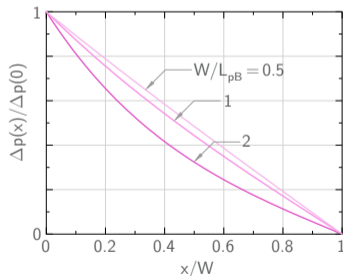
$$= \frac{1}{1 + \left( \frac{D_{nE}}{D_{pB}} \frac{L_{pB}}{L_{nE}} \frac{N_{dB}}{N_{aE}} \right) \frac{\sinh(W/L_B)}{\cosh(W/L_B)}}$$

since  $\frac{n_{0E}}{p_{0B}} = \frac{n_i^2}{N_{aE}} \times \frac{N_{dB}}{n_i^2} = \frac{N_{dB}}{N_{aE}}$ .

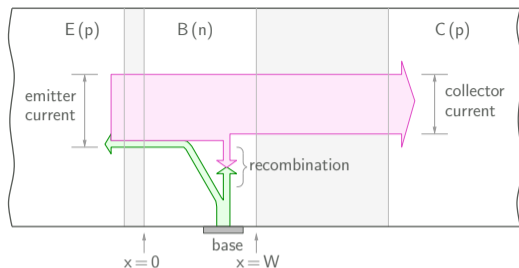
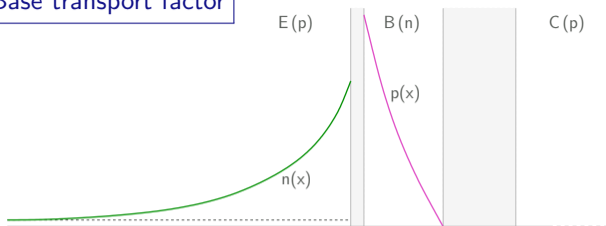
# Base transport factor



$$\Delta p(x) = \Delta p(0) \frac{\sinh\left(\frac{W-x}{L_{pB}}\right)}{\sinh\left(\frac{W}{L_{pB}}\right)} + \Delta p(W) \frac{\sinh\left(\frac{x}{L_{pB}}\right)}{\sinh\left(\frac{W}{L_{pB}}\right)}$$

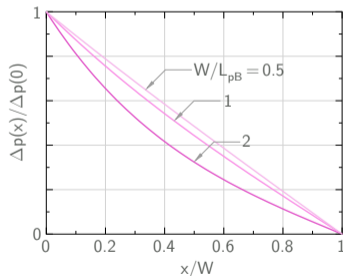


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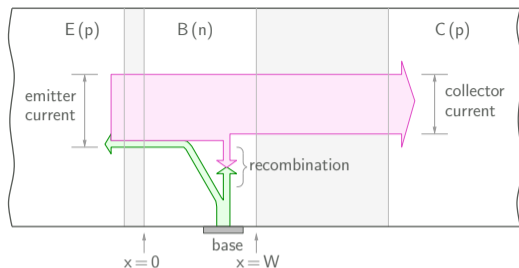
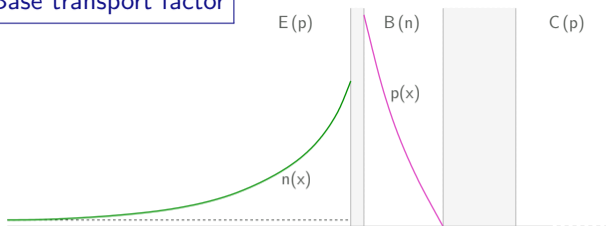


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$$I_C \approx I_{pC} = -qAD_{pB} \frac{d\Delta p}{dx}(W)$$



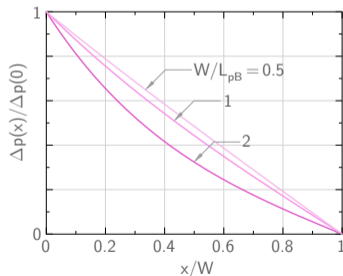
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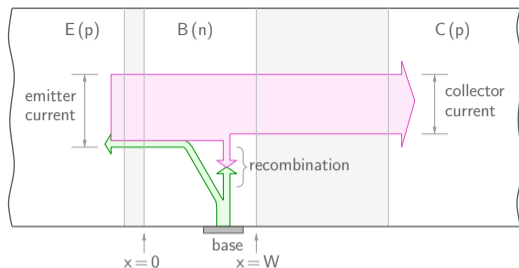
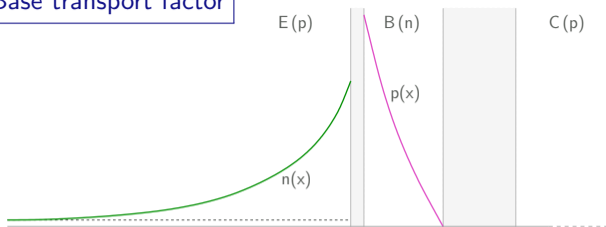
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## Base transport factor



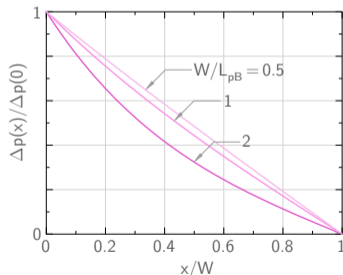
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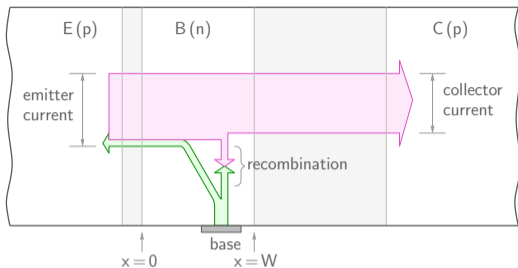
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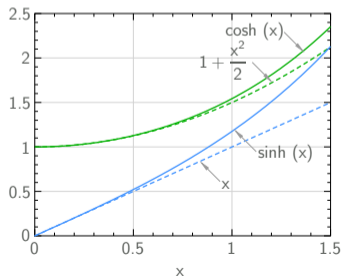
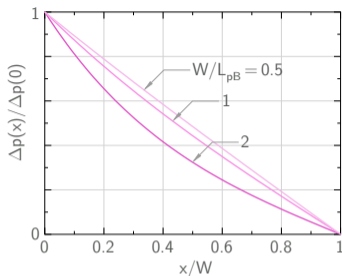
The base transport factor is (using  $I_{pE}$  from the last slide),

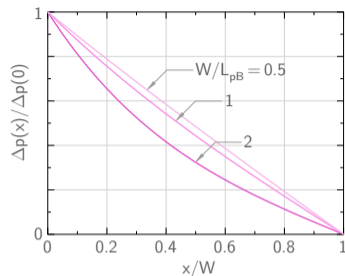
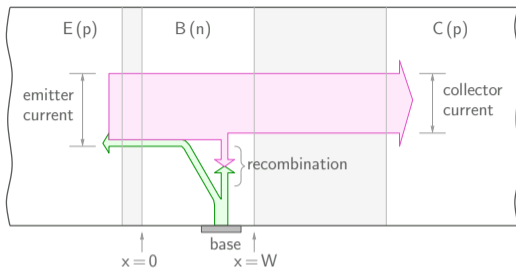
$$\alpha_T = \frac{I_{pC}}{I_{pE}} = \frac{1}{\cosh(W/L_{pB})}$$





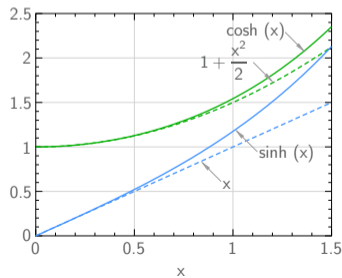
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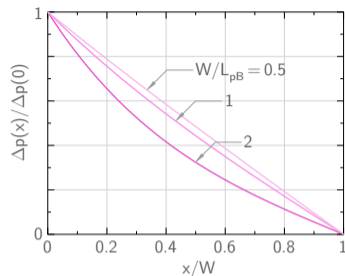
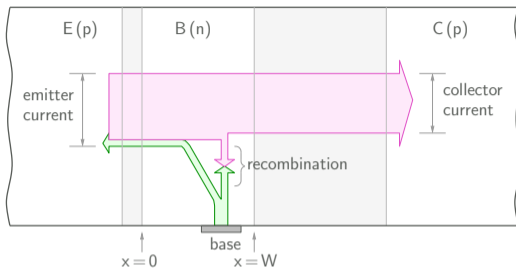




$$\Delta p(x) = \Delta p(0) \frac{\sinh\left(\frac{W-x}{L_{pB}}\right)}{\sinh\left(\frac{W}{L_{pB}}\right)} + \Delta p(W) \frac{\sinh\left(\frac{x}{L_{pB}}\right)}{\sinh\left(\frac{W}{L_{pB}}\right)}$$

$$\alpha_T = \frac{I_{pC}}{I_{pE}} = \frac{1}{\cosh(W/L_{pB})} \approx \frac{1}{1 + \frac{1}{2} \left(\frac{W}{L_{pB}}\right)^2}$$

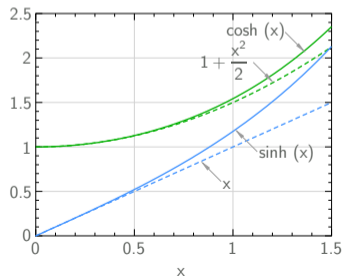




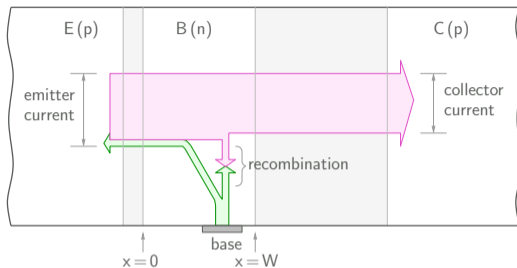
$$\Delta p(x) = \Delta p(0) \frac{\sinh\left(\frac{W-x}{L_{pB}}\right)}{\sinh\left(\frac{W}{L_{pB}}\right)} + \Delta p(W) \frac{\sinh\left(\frac{x}{L_{pB}}\right)}{\sinh\left(\frac{W}{L_{pB}}\right)}$$

$$\alpha_T = \frac{I_{pC}}{I_{pE}} = \frac{1}{\cosh(W/L_{pB})} \approx \frac{1}{1 + \frac{1}{2} \left(\frac{W}{L_{pB}}\right)^2}$$

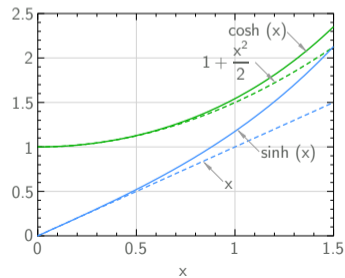
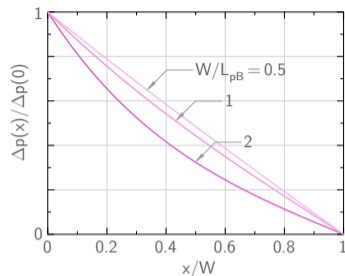
Remark:  $\alpha_T \rightarrow 1$  if the base width  $W$  is made small compared to  $L_{pB}$ .

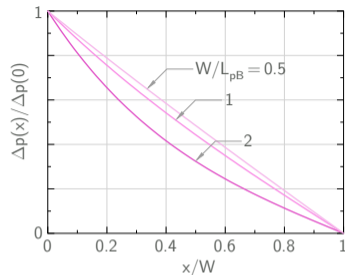
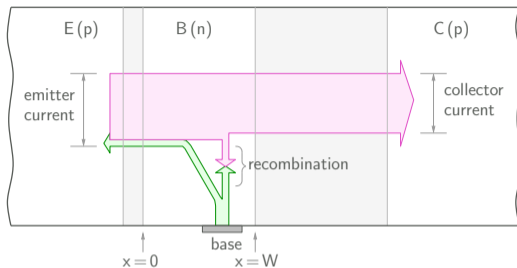






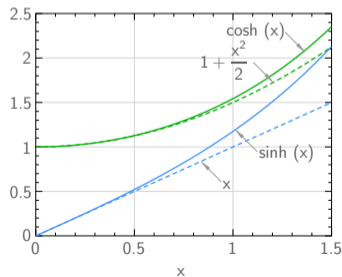
$$\gamma = \frac{I_{pE}}{I_{pE} + I_{nE}} = \frac{1}{1 + (I_{nE}/I_{pE})} = \frac{1}{1 + \left( \frac{D_{nE}}{D_{pB}} \frac{L_{pB}}{L_{nE}} \frac{N_{dB}}{N_{aE}} \right) \frac{\sinh(W/L_B)}{\cosh(W/L_B)}}$$

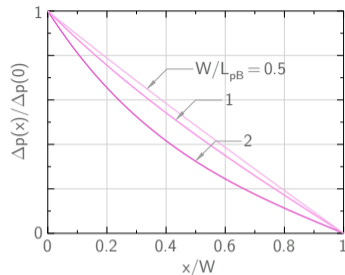
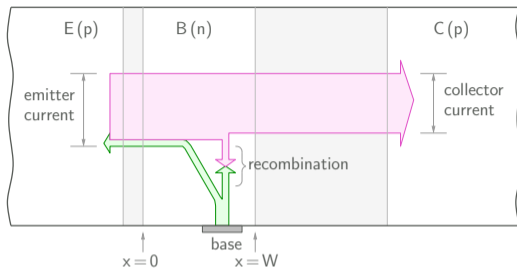




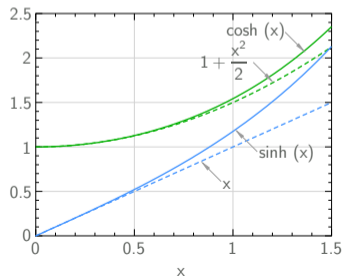
$$\gamma = \frac{I_{pE}}{I_{pE} + I_{nE}} = \frac{1}{1 + (I_{nE}/I_{pE})} = \frac{1}{1 + \left( \frac{D_{nE}}{D_{pB}} \frac{L_{pB}}{L_{nE}} \frac{N_{dB}}{N_{aE}} \right) \frac{\sinh(W/L_B)}{\cosh(W/L_B)}}$$

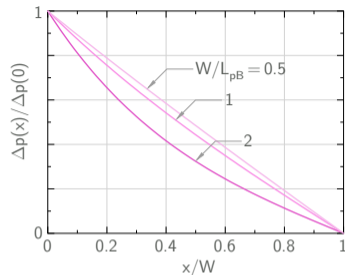
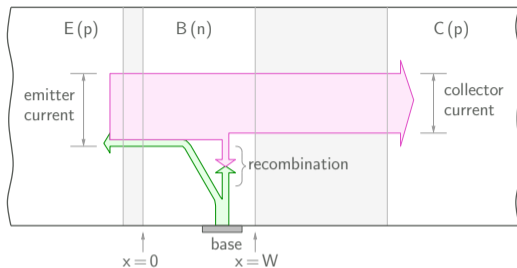
$$\approx \frac{1}{1 + \left( \frac{D_{nE}}{D_{pB}} \frac{L_{pB}}{L_{nE}} \frac{N_{dB}}{N_{aE}} \right) \frac{W/L_{pB}}{1 + \frac{1}{2}(W/L_{pB})^2}}$$



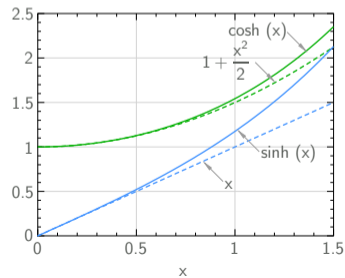


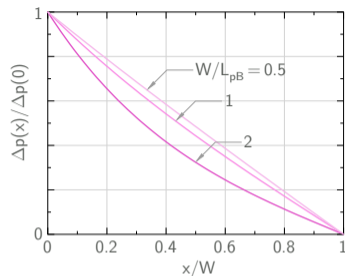
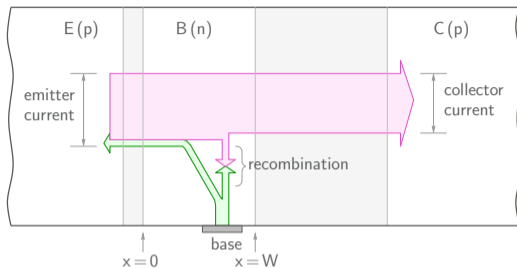
$$\begin{aligned} \gamma &= \frac{I_{pE}}{I_{pE} + I_{nE}} = \frac{1}{1 + (I_{nE}/I_{pE})} = \frac{1}{1 + \left( \frac{D_{nE}}{D_{pB}} \frac{L_{pB}}{L_{nE}} \frac{N_{dB}}{N_{aE}} \right) \frac{\sinh(W/L_B)}{\cosh(W/L_B)}} \\ &\approx \frac{1}{1 + \left( \frac{D_{nE}}{D_{pB}} \frac{L_{pB}}{L_{nE}} \frac{N_{dB}}{N_{aE}} \right) \frac{W/L_{pB}}{1 + \frac{1}{2}(W/L_{pB})^2}} \\ &\approx \frac{1}{1 + \left( \frac{D_{nE}}{D_{pB}} \frac{W}{L_{nE}} \frac{N_{dB}}{N_{aE}} \right)} \end{aligned}$$





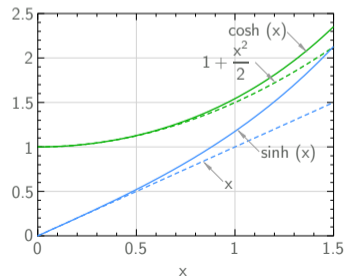
$$\gamma \approx \frac{1}{1 + \left( \frac{D_{nE}}{D_{pB}} \frac{W}{L_{nE}} \frac{N_{dB}}{N_{aE}} \right)}$$

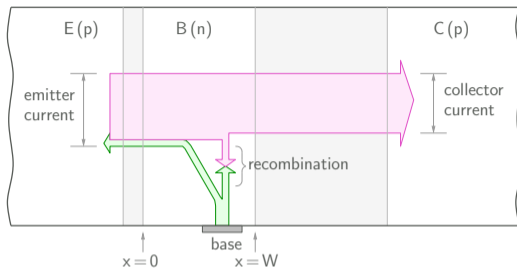




$$\gamma \approx \frac{1}{1 + \left( \frac{D_{nE}}{D_{pB}} \frac{W}{L_{nE}} \frac{N_{dB}}{N_{aE}} \right)}$$

\*  $\gamma \rightarrow 1$  if  $N_{aE} \gg N_{dB}$ .

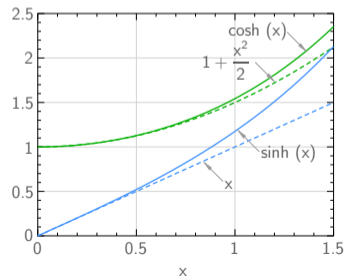
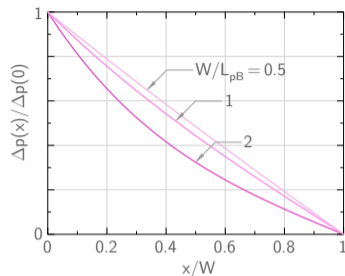




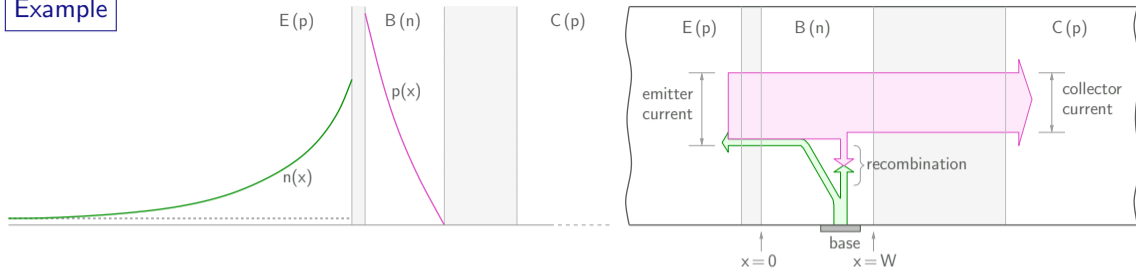
$$\gamma \approx \frac{1}{1 + \left( \frac{D_{nE}}{D_{pB}} \frac{W}{L_{nE}} \frac{N_{dB}}{N_{aE}} \right)}$$

\*  $\gamma \rightarrow 1$  if  $N_{aE} \gg N_{dB}$ .

\* It is now clear why a higher doping density in the emitter region (compared to the base doping density) is desirable.



## Example



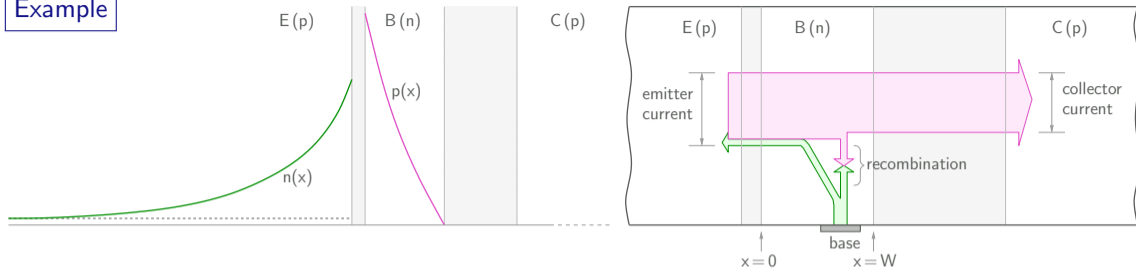
Consider a *pnp* BJT with  $N_{aE} = 10^{18} \text{ cm}^{-3}$ ,  $N_{dB} = 5 \times 10^{16} \text{ cm}^{-3}$ ,  $N_{aC} = 10^{15} \text{ cm}^{-3}$ , and with a base width  $W = 2 \mu\text{m}$  ( $T = 300 \text{ K}$ ).

(a) Calculate  $\alpha_T$ ,  $\gamma$ ,  $\alpha$ , and  $\beta$ , using the following parameters.

$$\mu_{nE} = 250 \text{ cm}^2/\text{V-s}, \mu_{pB} = 500 \text{ cm}^2/\text{V-s}, \mu_{nC} = 1500 \text{ cm}^2/\text{V-s},$$

$$\tau_{nE} = 0.2 \mu\text{s}, \tau_{pB} = 1 \mu\text{s}, \tau_{nC} = 1 \mu\text{s}.$$

## Example



Consider a *pnp* BJT with  $N_{aE} = 10^{18} \text{ cm}^{-3}$ ,  $N_{dB} = 5 \times 10^{16} \text{ cm}^{-3}$ ,  $N_{aC} = 10^{15} \text{ cm}^{-3}$ , and with a base width  $W = 2 \mu\text{m}$  ( $T = 300 \text{ K}$ ).

(a) Calculate  $\alpha_T$ ,  $\gamma$ ,  $\alpha$ , and  $\beta$ , using the following parameters.

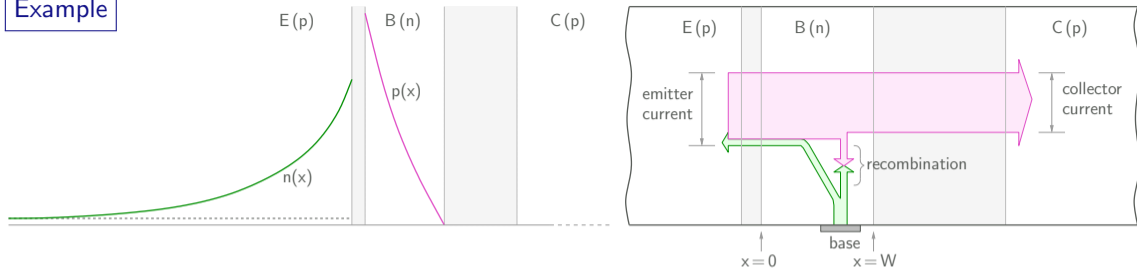
$$\mu_{nE} = 250 \text{ cm}^2/\text{V-s}, \mu_{pB} = 500 \text{ cm}^2/\text{V-s}, \mu_{nC} = 1500 \text{ cm}^2/\text{V-s},$$

$$\tau_{nE} = 0.2 \mu\text{s}, \tau_{pB} = 1 \mu\text{s}, \tau_{nC} = 1 \mu\text{s}.$$

(b) Repeat (a) for the BJT operating in the reverse active mode.



## Example



Solution:

The minority carrier diffusion lengths are

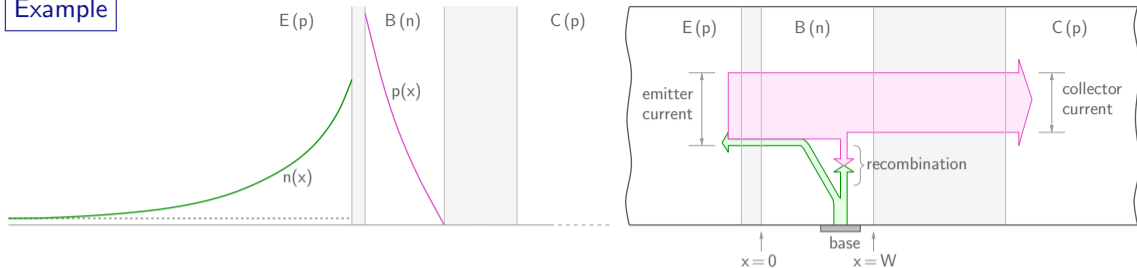
$$L_{nE} = \sqrt{D_{nE}\tau_{nE}} = \sqrt{V_T\mu_{nE}\tau_{nE}} = \sqrt{0.0258 \times 250 \times 0.2 \times 10^{-6}} = 1.14 \times 10^{-3} \text{ cm} = 11.4 \mu\text{m}.$$

$$L_{pB} = \sqrt{D_{pB}\tau_{pB}} = \sqrt{V_T\mu_{pB}\tau_{pB}} = \sqrt{0.0258 \times 500 \times 1 \times 10^{-6}} = 3.59 \times 10^{-3} \text{ cm} = 35.9 \mu\text{m}.$$

$$L_{nC} = \sqrt{D_{nC}\tau_{nC}} = \sqrt{V_T\mu_{nC}\tau_{nC}} = \sqrt{0.0258 \times 1500 \times 1 \times 10^{-6}} = 6.22 \times 10^{-3} \text{ cm} = 62.2 \mu\text{m}.$$

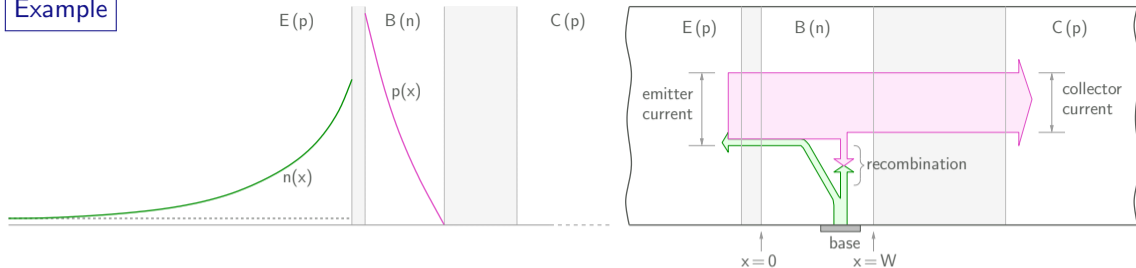
Note that  $L_{pB} \gg W$  ( $2 \mu\text{m}$ ).

## Example



$$(a) \frac{D_{nE}}{D_{pB}} \frac{W}{L_{nE}} \frac{N_{dB}}{N_{aE}} = \frac{\mu_{nE}}{\mu_{pB}} \frac{W}{L_{nE}} \frac{N_{dB}}{N_{aE}} = \frac{250}{500} \frac{2 \times 10^{-4}}{1.14 \times 10^{-3}} \frac{5 \times 10^{16}}{10^{18}} = 4.386 \times 10^{-3}.$$

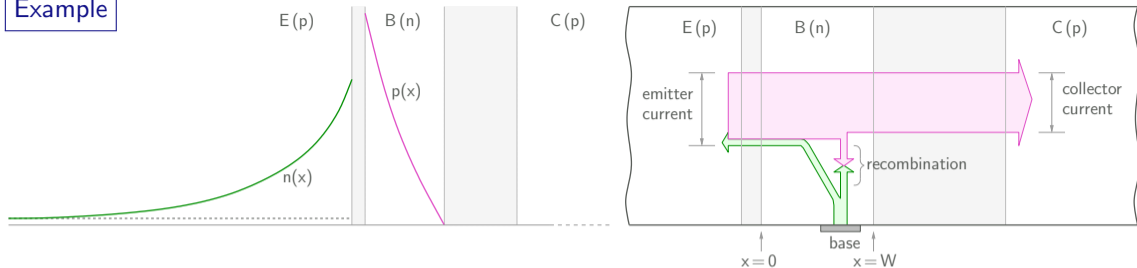
## Example



$$(a) \frac{D_{nE}}{D_{pB}} \frac{W}{L_{nE}} \frac{N_{dB}}{N_{aE}} = \frac{\mu_{nE}}{\mu_{pB}} \frac{W}{L_{nE}} \frac{N_{dB}}{N_{aE}} = \frac{250}{500} \frac{2 \times 10^{-4}}{1.14 \times 10^{-3}} \frac{5 \times 10^{16}}{10^{18}} = 4.386 \times 10^{-3}.$$

$$\gamma = \frac{1}{1 + 4.386 \times 10^{-3}} = 0.9956.$$

## Example

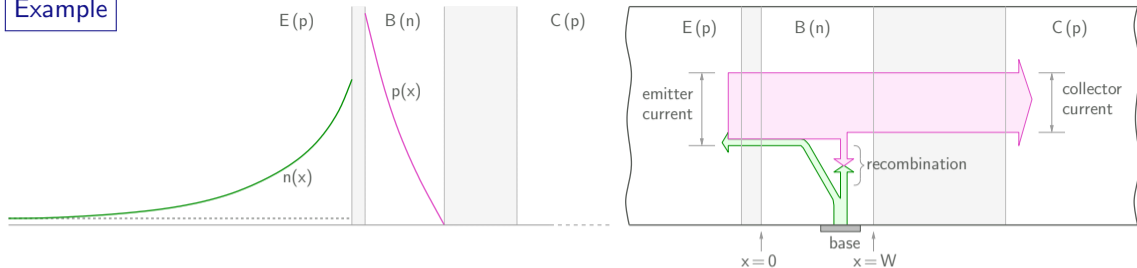


$$(a) \frac{D_{nE}}{D_{pB}} \frac{W}{L_{nE}} \frac{N_{dB}}{N_{aE}} = \frac{\mu_{nE}}{\mu_{pB}} \frac{W}{L_{nE}} \frac{N_{dB}}{N_{aE}} = \frac{250}{500} \frac{2 \times 10^{-4}}{1.14 \times 10^{-3}} \frac{5 \times 10^{16}}{10^{18}} = 4.386 \times 10^{-3}.$$

$$\gamma = \frac{1}{1 + 4.386 \times 10^{-3}} = 0.9956.$$

$$\alpha_T = \frac{1}{1 + \frac{1}{2}(W/L_{pB})^2} = \frac{1}{1 + \frac{1}{2}(2.0/35.9)^2} = 0.9985.$$

## Example



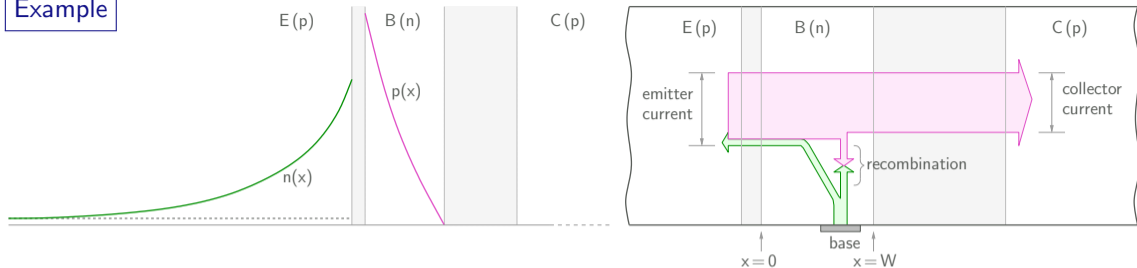
$$(a) \frac{D_{nE}}{D_{pB}} \frac{W}{L_{nE}} \frac{N_{dB}}{N_{aE}} = \frac{\mu_{nE}}{\mu_{pB}} \frac{W}{L_{nE}} \frac{N_{dB}}{N_{aE}} = \frac{250}{500} \frac{2 \times 10^{-4}}{1.14 \times 10^{-3}} \frac{5 \times 10^{16}}{10^{18}} = 4.386 \times 10^{-3}.$$

$$\gamma = \frac{1}{1 + 4.386 \times 10^{-3}} = 0.9956.$$

$$\alpha_T = \frac{1}{1 + \frac{1}{2}(W/L_{pB})^2} = \frac{1}{1 + \frac{1}{2}(2.0/35.9)^2} = 0.9985.$$

$$\alpha = \gamma \alpha_T = 0.9940 \rightarrow \beta = \frac{\alpha}{1 - \alpha} = 166.$$

## Example



(b) With  $E \leftrightarrow C$ ,

$$\frac{D_{nE}}{D_{pB}} \frac{W}{L_{nE}} \frac{N_{dB}}{N_{aE}} \rightarrow \frac{D_{nC}}{D_{pB}} \frac{W}{L_{nC}} \frac{N_{dB}}{N_{aC}} = \frac{\mu_{nC}}{\mu_{pB}} \frac{W}{L_{nC}} \frac{N_{dB}}{N_{aC}} = \frac{1500}{500} \frac{2 \times 10^{-4}}{6.22 \times 10^{-3}} \frac{5 \times 10^{16}}{10^{15}} = 4.823.$$

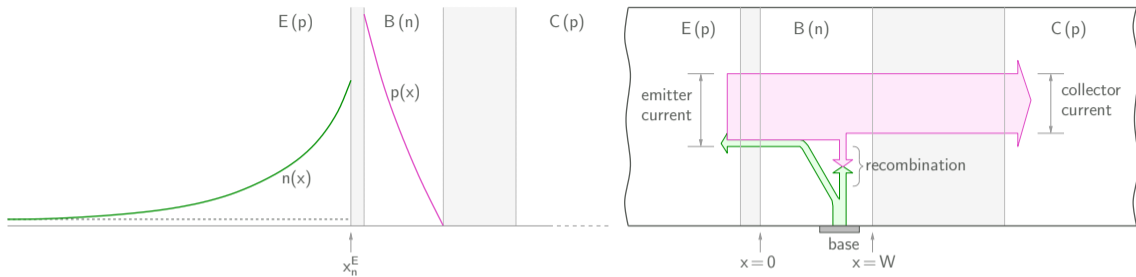
$$\gamma = \frac{1}{1 + 4.823} = 0.1717, \quad \alpha_T = \frac{1}{1 + \frac{1}{2}(2/35.9)^2} = 0.9985.$$

$$\rightarrow \alpha = \gamma \alpha_T = 0.1714 \rightarrow \beta = 0.2, \text{ a disaster.}$$

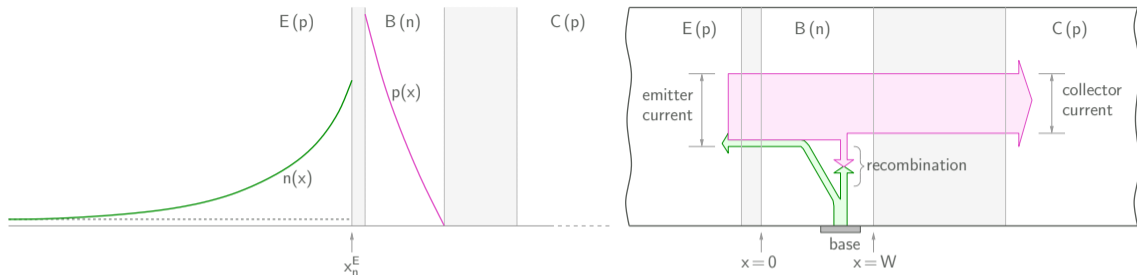
Conclusion:  $N_{aE} \gg N_{dB}$  is crucial.

(Note that we have treated  $W$  as a constant, but it would vary with bias conditions.)

$\gamma$  with  $W \ll L_{pB}$



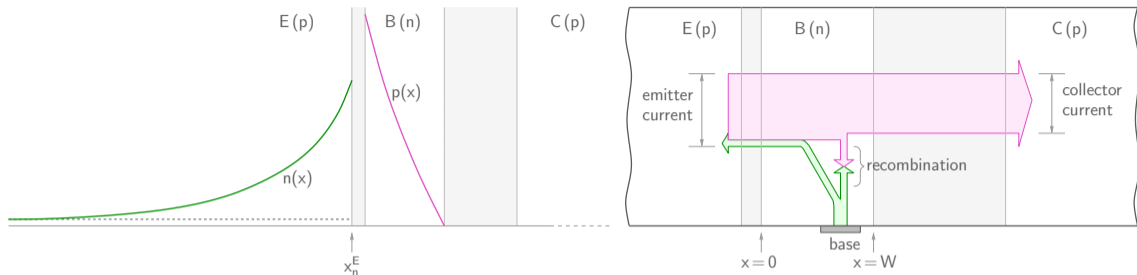
$\gamma$  with  $W \ll L_{pB}$



$$I_{nE} = -q A D_{nE} \frac{d}{dx} \left[ n_{0E} \exp\left(\frac{V_{EB}}{V_T}\right) \exp\left(-\frac{x_n^E - x}{L_{nE}}\right) \right] \text{ at } x = x_n^E$$



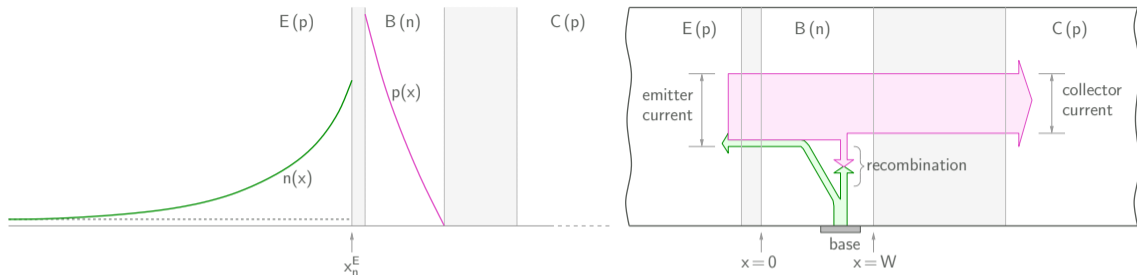
$\gamma$  with  $W \ll L_{pB}$



$$I_{nE} = -q A D_{nE} \frac{d}{dx} \left[ n_{0E} \exp\left(\frac{V_{EB}}{V_T}\right) \exp\left(-\frac{x_n^E - x}{L_{nE}}\right) \right] \text{ at } x = x_n^E$$

$$= q A D_{nE} n_{0E} \exp\left(\frac{V_{EB}}{V_T}\right) \times \frac{1}{L_{nE}}$$

$\gamma$  with  $W \ll L_{pB}$

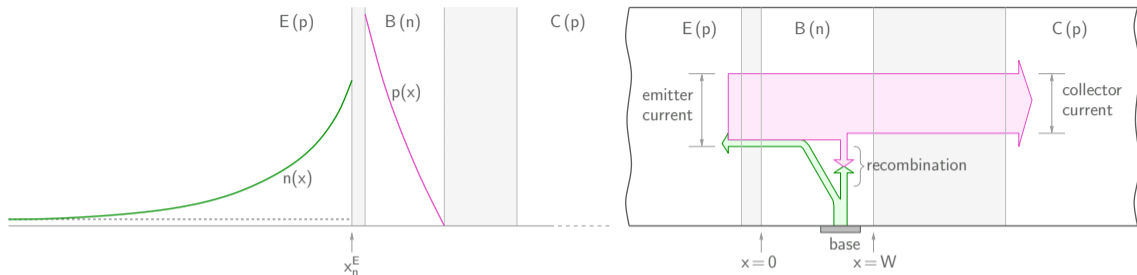


$$I_{nE} = -q A D_{nE} \frac{d}{dx} \left[ n_{0E} \exp\left(\frac{V_{EB}}{V_T}\right) \exp\left(-\frac{x_n^E - x}{L_{nE}}\right) \right] \text{ at } x = x_n^E$$

$$= q A D_{nE} n_{0E} \exp\left(\frac{V_{EB}}{V_T}\right) \times \frac{1}{L_{nE}}$$

$$I_{pE} \approx q A D_{pB} p_{0B} \exp\left(\frac{V_{EB}}{V_T}\right) \times \frac{1}{W}$$

$\gamma$  with  $W \ll L_{pB}$



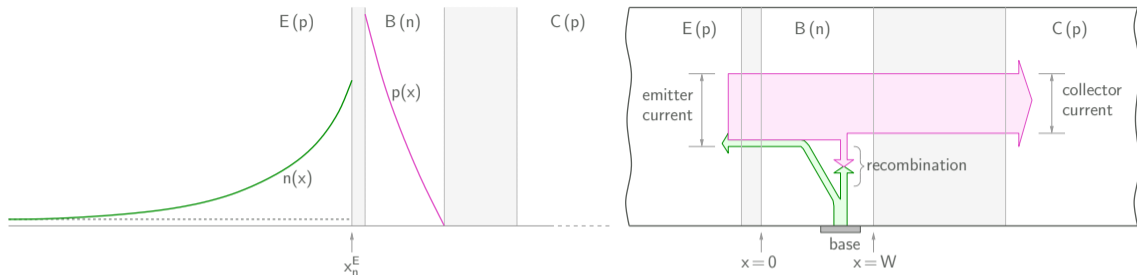
$$I_{nE} = -q A D_{nE} \frac{d}{dx} \left[ n_{0E} \exp\left(\frac{V_{EB}}{V_T}\right) \exp\left(-\frac{x_n^E - x}{L_{nE}}\right) \right] \text{ at } x = x_n^E$$

$$= q A D_{nE} n_{0E} \exp\left(\frac{V_{EB}}{V_T}\right) \times \frac{1}{L_{nE}}$$

$$I_{pE} \approx q A D_{pB} p_{0B} \exp\left(\frac{V_{EB}}{V_T}\right) \times \frac{1}{W}$$

$$\frac{I_{nE}}{I_{pE}} \approx \frac{D_{nE}}{D_{pB}} \frac{n_{0E}}{p_{0B}} \frac{W}{L_{nE}}$$

$\gamma$  with  $W \ll L_{pB}$



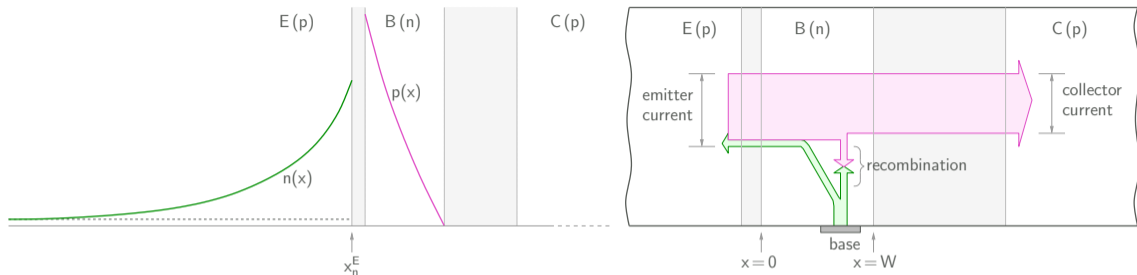
$$I_{nE} = -q A D_{nE} \frac{d}{dx} \left[ n_{0E} \exp\left(\frac{V_{EB}}{V_T}\right) \exp\left(-\frac{x_n^E - x}{L_{nE}}\right) \right] \text{ at } x = x_n^E$$

$$= q A D_{nE} n_{0E} \exp\left(\frac{V_{EB}}{V_T}\right) \times \frac{1}{L_{nE}}$$

$$I_{pE} \approx q A D_{pB} p_{0B} \exp\left(\frac{V_{EB}}{V_T}\right) \times \frac{1}{W}$$

$$\frac{I_{nE}}{I_{pE}} \approx \frac{D_{nE}}{D_{pB}} \frac{n_{0E}}{p_{0B}} \frac{W}{L_{nE}} = \frac{D_{nE}}{D_{pB}} \frac{n_i^2}{N_{aE}} \frac{N_{dB}}{n_i^2} \frac{W}{L_{nE}}$$

$\gamma$  with  $W \ll L_{pB}$



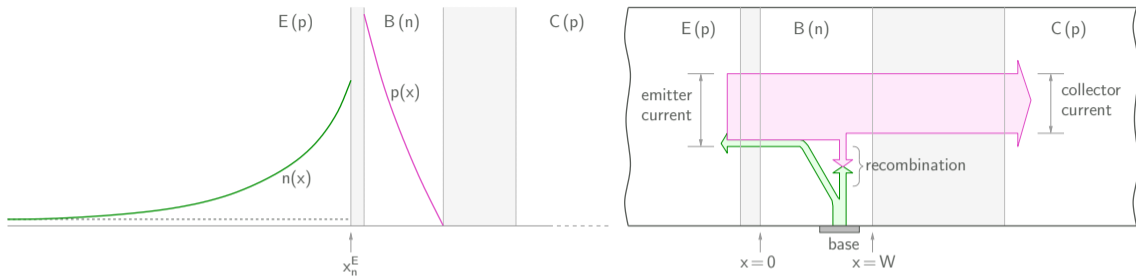
$$I_{nE} = -q A D_{nE} \frac{d}{dx} \left[ n_{0E} \exp\left(\frac{V_{EB}}{V_T}\right) \exp\left(-\frac{x_n^E - x}{L_{nE}}\right) \right] \text{ at } x = x_n^E$$

$$= q A D_{nE} n_{0E} \exp\left(\frac{V_{EB}}{V_T}\right) \times \frac{1}{L_{nE}}$$

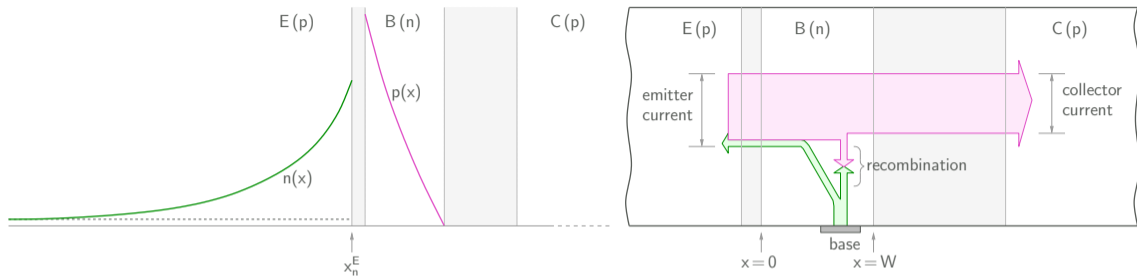
$$I_{pE} \approx q A D_{pB} p_{0B} \exp\left(\frac{V_{EB}}{V_T}\right) \times \frac{1}{W}$$

$$\frac{I_{nE}}{I_{pE}} \approx \frac{D_{nE}}{D_{pB}} \frac{n_{0E}}{p_{0B}} \frac{W}{L_{nE}} = \frac{D_{nE}}{D_{pB}} \frac{n_i^2}{N_{aE}} \frac{N_{dB}}{n_i^2} \frac{W}{L_{nE}} = \frac{D_{nE}}{D_{pB}} \frac{N_{dB}}{N_{aE}} \frac{W}{L_{nE}}$$

$\gamma$  with  $W \ll L_{pB}$

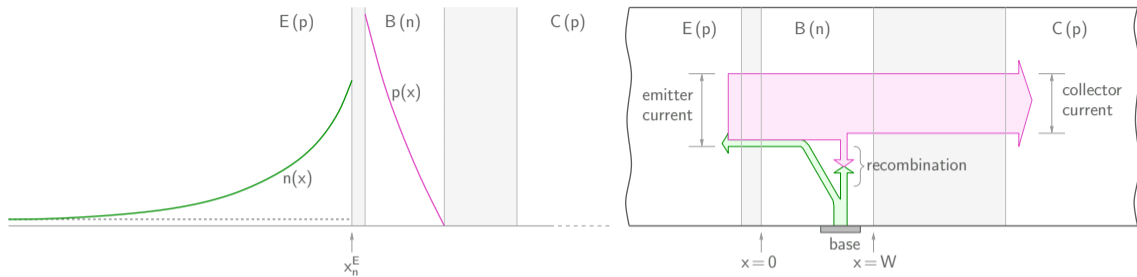


$\gamma$  with  $W \ll L_{pB}$



$$\frac{I_{nE}}{I_{pE}} \approx \frac{D_{nE}}{D_{pB}} \frac{N_{dB}}{N_{aE}} \frac{W}{L_{nE}}$$

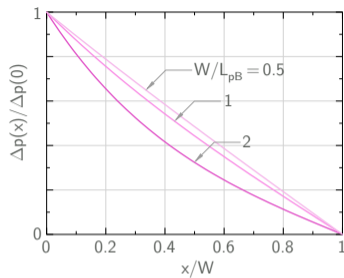
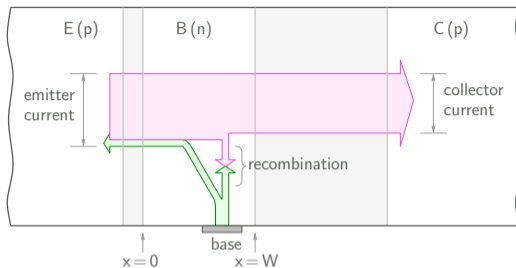
$\gamma$  with  $W \ll L_{pB}$



$$\frac{I_{nE}}{I_{pE}} \approx \frac{D_{nE}}{D_{pB}} \frac{N_{dB}}{N_{aE}} \frac{W}{L_{nE}}$$

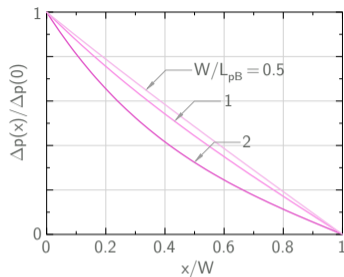
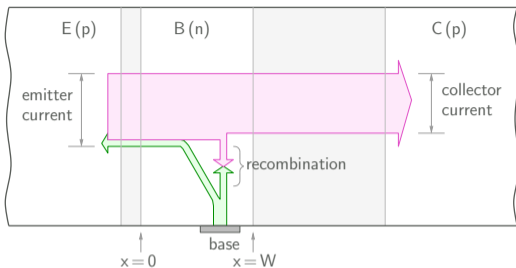
$$\rightarrow \gamma = \frac{I_{pE}}{I_{pE} + I_{nE}} = \frac{1}{1 + \frac{I_{nE}}{I_{pE}}} \approx \frac{1}{1 + \left( \frac{D_{nE}}{D_{pB}} \frac{W}{L_{nE}} \frac{N_{dB}}{N_{aE}} \right)}$$





When  $W \ll L_{pB}$ ,  $\Delta p(x)$  is linear.

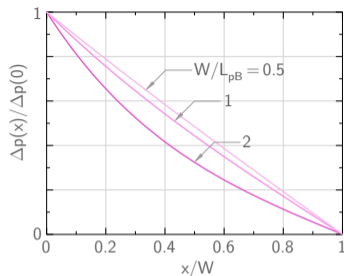
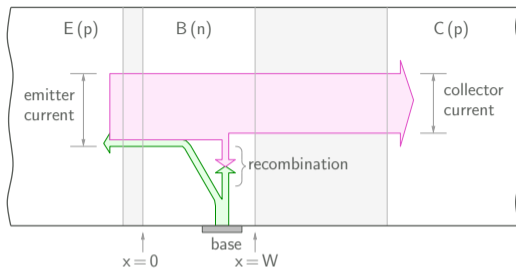
$$\alpha_T = \frac{I_{pC}}{I_{pE}} = \frac{I_{pC}}{I_{pC} + I_{pB}} = \frac{1}{1 + \frac{I_{pB}}{I_{pC}}}$$



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$$I_{pC} = -q A D_{pB} \frac{dp}{dx}(W) \approx q A D_{pB} \frac{\Delta p(0)}{W}$$

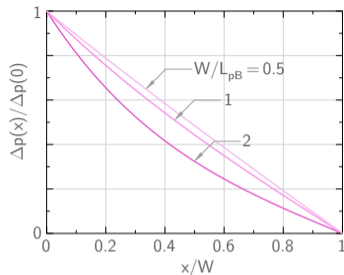
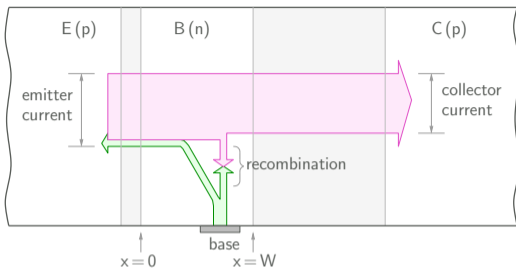


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$$I_{pB} = \frac{Q_p}{\tau_{pB}} = \frac{q A \frac{1}{2} \Delta p(0) W}{\tau_{pB}} \rightarrow \alpha_T \approx \frac{1}{1 + \frac{1}{2} \left( \frac{W}{L_{pB}} \right)^2}$$