

SEMICONDUCTOR DEVICES

p - n Junctions: Part 3



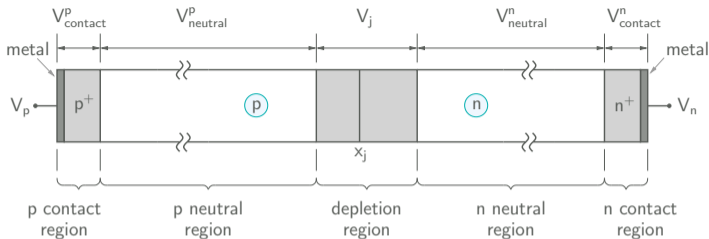
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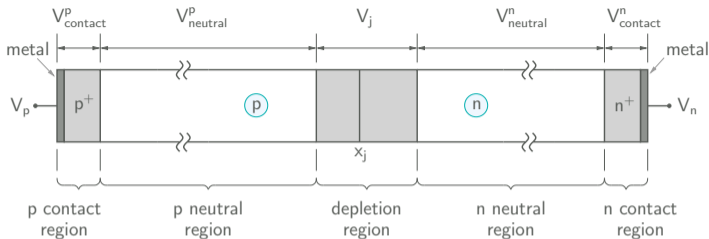
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Indian Institute of Technology Bombay

pn junction: derivation of I - V equation



Continuity equation for holes ($x > x_n$):
$$\frac{\partial p(x, t)}{\partial t} = -\frac{1}{q} \frac{\partial J_p}{\partial x} - (R - G) = 0 \text{ (in DC conditions).}$$

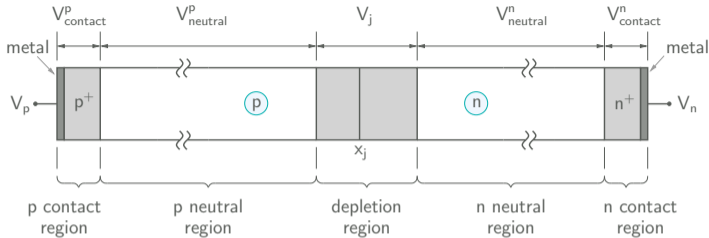
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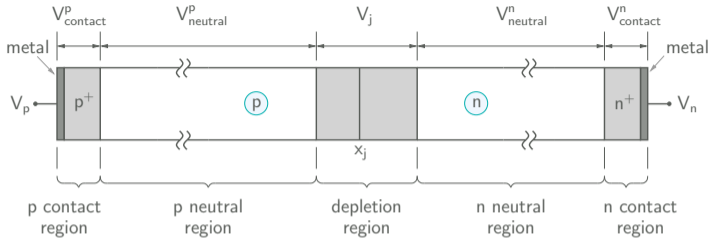


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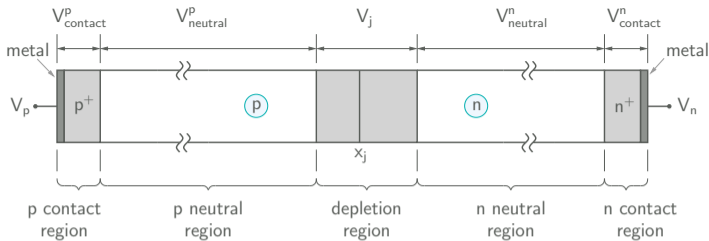


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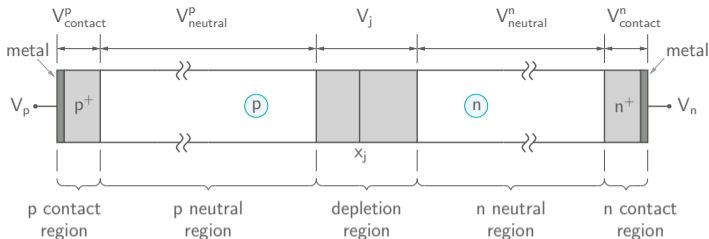
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$\rightarrow D_p \frac{d^2 p}{dx^2} - \frac{p - p_{n0}}{\tau_p} = 0$ or $\frac{d^2 \Delta p}{dx^2} - \frac{\Delta p}{L_p^2} = 0$, where $L_p = \sqrt{D_p \tau_p}$ is the hole diffusion length.

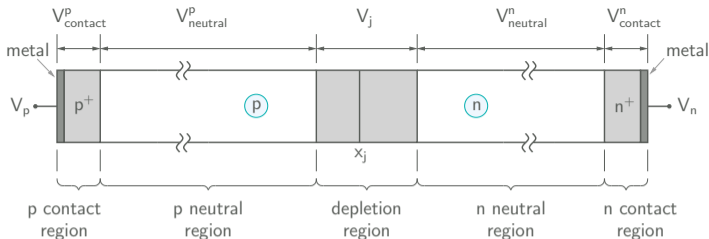


Example: For an abrupt, uniformly doped silicon pn junction at $T = 300$ K, with $N_a = 10^{16} \text{ cm}^{-3}$, $N_d = 10^{17} \text{ cm}^{-3}$, $\mu_p = 500 \text{ cm}^2/\text{V-s}$, calculate the diffusion length L_p for $\tau_p = 1 \text{ ns}$, 10 ns , 100 ns , $1 \mu\text{s}$, and $10 \mu\text{s}$, and compare it with the zero-bias value of W , the depletion width.



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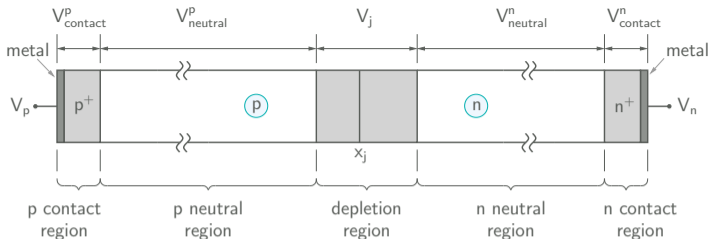
Solution:
$$V_{bi} = V_T \log \left(\frac{N_a N_d}{n_i^2} \right) = (0.0259 \text{ V}) \log \left(\frac{10^{16} \times 10^{17}}{(1.5 \times 10^{10})^2} \right) = 0.75 \text{ V}.$$



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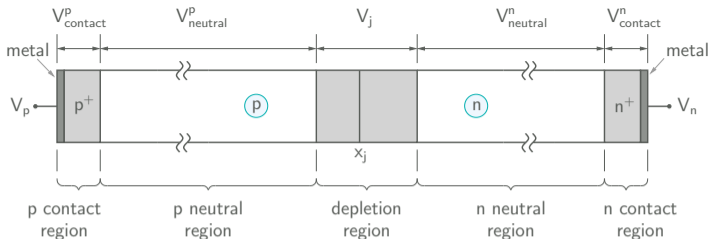


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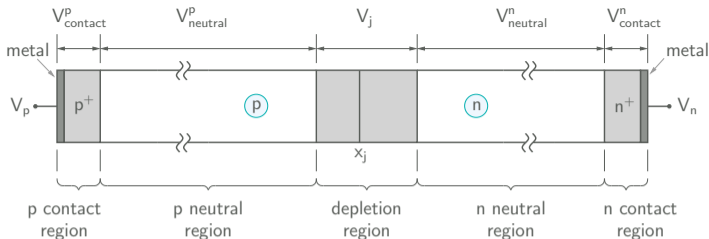
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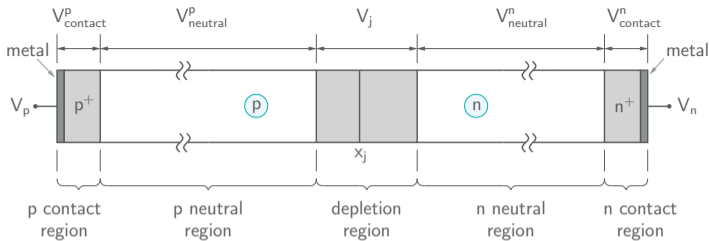
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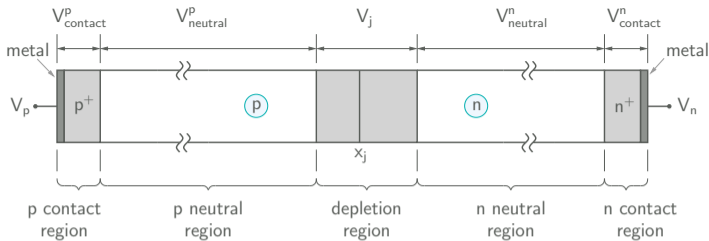
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For $\tau_p = 1 \text{ ns}$, $L_p = \sqrt{12.9 \frac{\text{cm}^2}{\text{s}} \times (1 \times 10^{-9} \text{ s})} = 1.14 \times 10^{-4} \text{ cm} = 1.14 \mu\text{m}.$



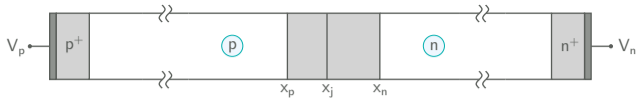
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τ_p	L_p (μm)
1 ns	1.14
10 ns	3.6
100 ns	11.4
1 μs	36.0
10 μs	113.8

Note that $L_p \gg W|_{0V}$ ($0.33 \mu\text{m}$), a typical situation.



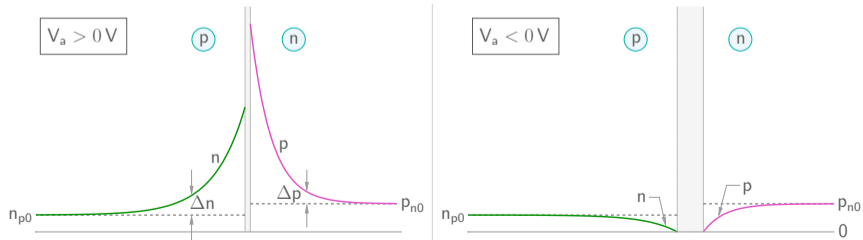
Hole continuity equation ($x > x_n$): $\frac{d^2 \Delta p}{dx^2} - \frac{\Delta p}{L_p^2} = 0$.

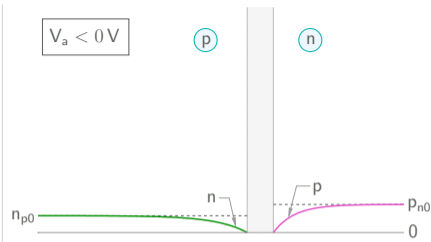
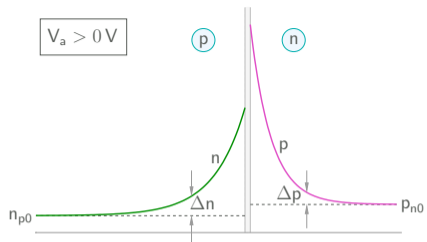
Boundary conditions: $\Delta p(x_n) = p_{n0} \exp\left(\frac{V_a}{V_T}\right) - p_{n0} = p_{n0} \left[\exp\left(\frac{V_a}{V_T}\right) - 1 \right]$

$\Delta p(x \rightarrow \infty) = p(x \rightarrow \infty) - p_{n0} = 0$

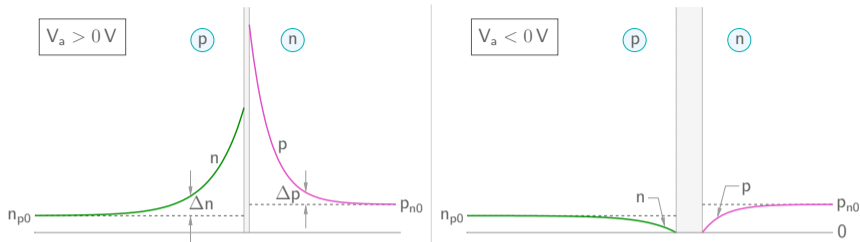
$\rightarrow \Delta p(x) = p_{n0} \left[\exp\left(\frac{V_a}{V_T}\right) - 1 \right] \exp\left(-\frac{x - x_n}{L_p}\right), \quad x > x_n,$

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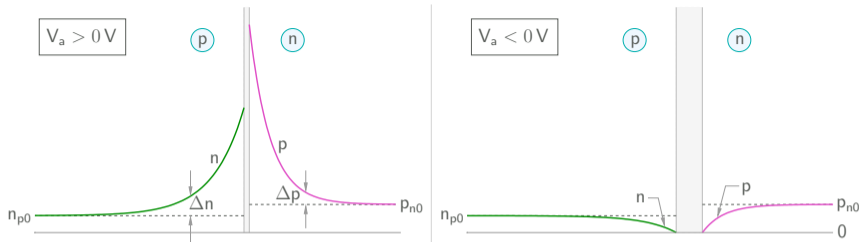


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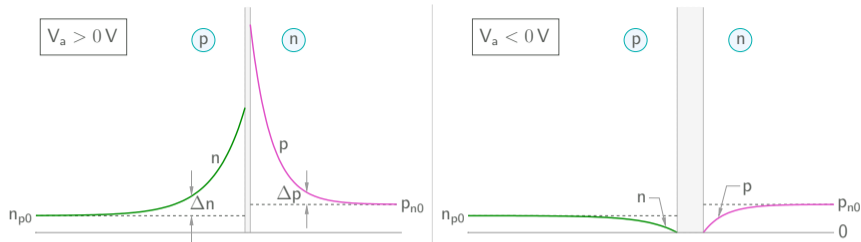


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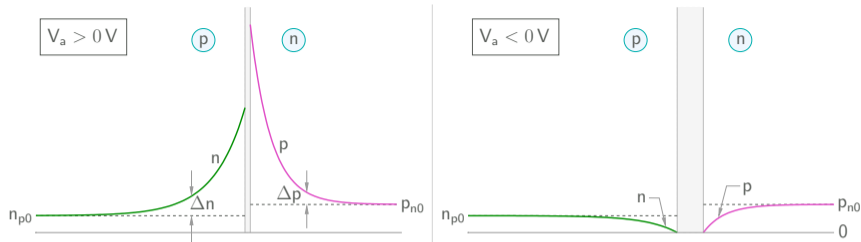
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For forward bias, $\Delta p(x_n)$ and $\Delta n(x_p)$ are positive.



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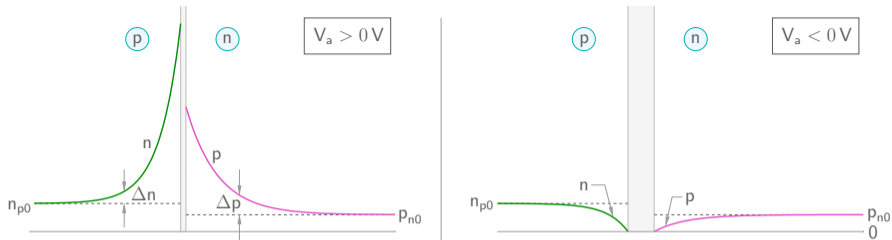
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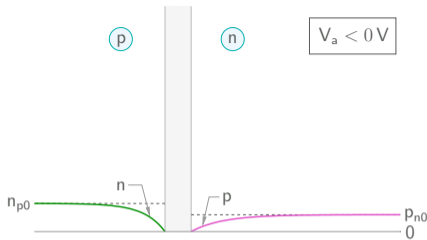
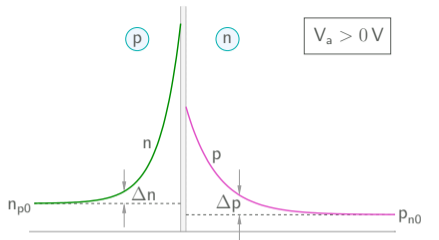
For forward bias, $\Delta p(x_n)$ and $\Delta n(x_p)$ are positive.

For reverse bias, $\Delta p(x_n)$ and $\Delta n(x_p)$ are negative.



Consider an abrupt, uniformly doped silicon pn junction at $T = 300\text{ K}$, with $N_a = 5 \times 10^{16}\text{ cm}^{-3}$ and $N_d = 10^{18}\text{ cm}^{-3}$. Compute $\Delta n(x_p)$ and $\Delta p(x_n)$ for $V_a = 0.1, 0.2, 0.3, 0.6, 0.7, -0.1, -0.2, -0.5, -1,$ and -2 V . ($n_i = 1.5 \times 10^{10}\text{ cm}^{-3}$ for silicon at $T = 300\text{ K}$.)

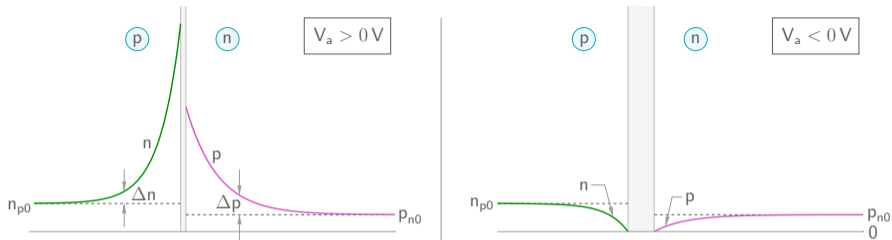
Solution: $p_{p0} \approx N_a = 5 \times 10^{16}\text{ cm}^{-3} \rightarrow n_{p0} = \frac{n_i^2}{p_{p0}} = \frac{(1.5 \times 10^{10})^2}{5 \times 10^{16}} = 4.5 \times 10^3\text{ cm}^{-3}$.



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$$n_{n0} \approx N_d = 1 \times 10^{18} \text{ cm}^{-3} \rightarrow p_{n0} = \frac{n_i^2}{n_{n0}} = \frac{(1.5 \times 10^{10})^2}{1 \times 10^{18}} = 2.25 \times 10^2 \text{ cm}^{-3}.$$

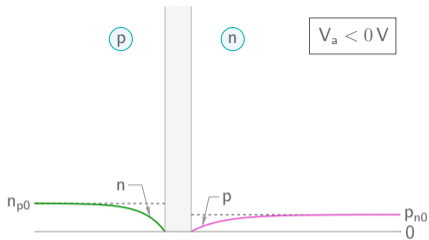
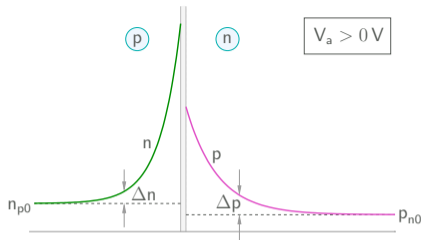


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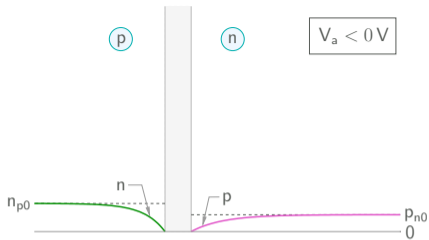
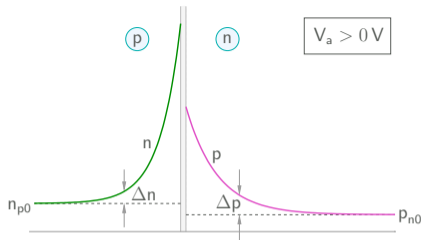
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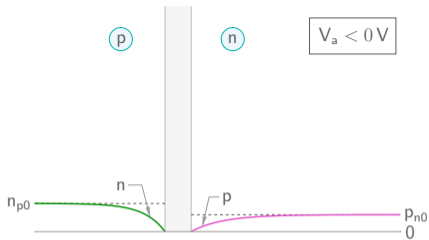
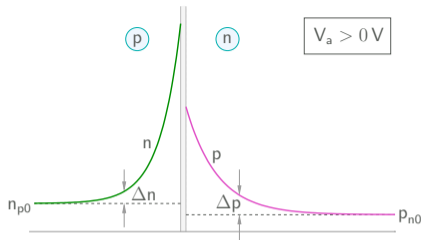


V_a (V)	$\Delta n(x_p)$ (cm^{-3})	$\Delta p(x_n)$ (cm^{-3})	V_a (V)	$\Delta n(x_p)$ (cm^{-3})	$\Delta p(x_n)$ (cm^{-3})
0	0	0	0	0	0
0.1	2.09×10^5	1.05×10^4	-0.1	-4.41×10^3	-2.20×10^2
0.2	1.02×10^7	5.08×10^5	-0.2	-4.50×10^3	-2.25×10^2
0.3	4.83×10^8	2.41×10^7	-0.5	-4.50×10^3	-2.25×10^2
0.6	5.18×10^{13}	2.59×10^{12}	-1	-4.50×10^3	-2.25×10^2
0.7	2.46×10^{15}	1.23×10^{14}	-2	-4.50×10^3	-2.25×10^2



V_a (V)	$\Delta n(x_p)$ (cm^{-3})	$\Delta p(x_n)$ (cm^{-3})	V_a (V)	$\Delta n(x_p)$ (cm^{-3})	$\Delta p(x_n)$ (cm^{-3})
0	0	0	0	0	0
0.1	2.09×10^5	1.05×10^4	-0.1	-4.41×10^3	-2.20×10^2
0.2	1.02×10^7	5.08×10^5	-0.2	-4.50×10^3	-2.25×10^2
0.3	4.83×10^8	2.41×10^7	-0.5	-4.50×10^3	-2.25×10^2
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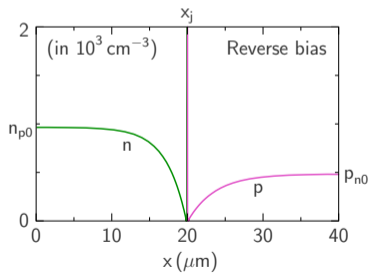
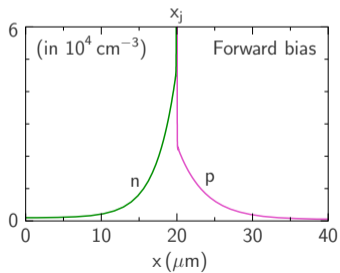
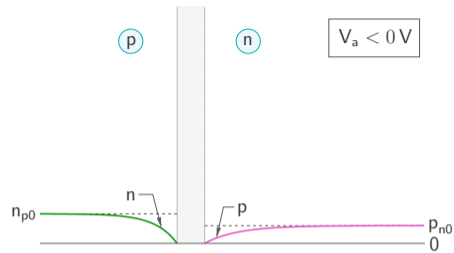
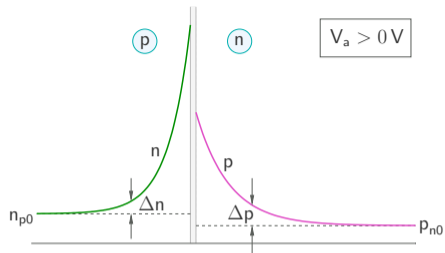
* Forward bias:
 $\Delta p(x_n)$ and $\Delta n(x_p)$ increase by several orders of magnitude as V_a is increased.



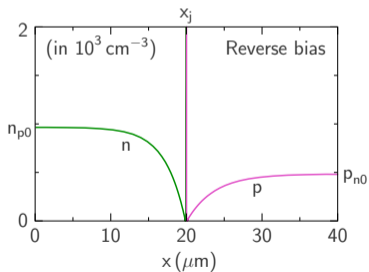
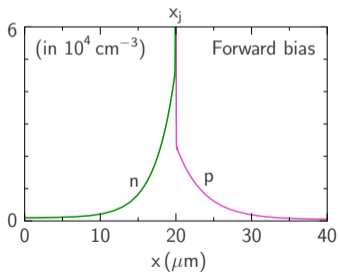
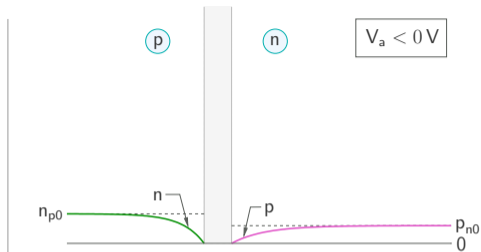
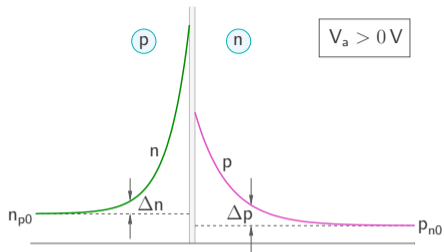
V_a (V)	$\Delta n(x_p)$ (cm^{-3})	$\Delta p(x_n)$ (cm^{-3})	V_a (V)	$\Delta n(x_p)$ (cm^{-3})	$\Delta p(x_n)$ (cm^{-3})
0	0	0	0	0	0
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- * Forward bias:
 $\Delta p(x_n)$ and $\Delta n(x_p)$ increase by several orders of magnitude as V_a is increased.
- * Reverse bias:
 $\Delta p(x_n) \approx -p_{n0}$, $\Delta n(x_p) \approx -n_{p0}$.

pn junction under forward bias: simulation results

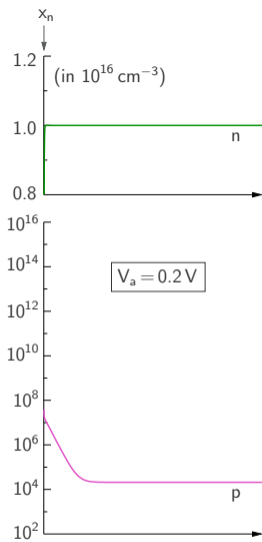


pn junction under forward bias: simulation results

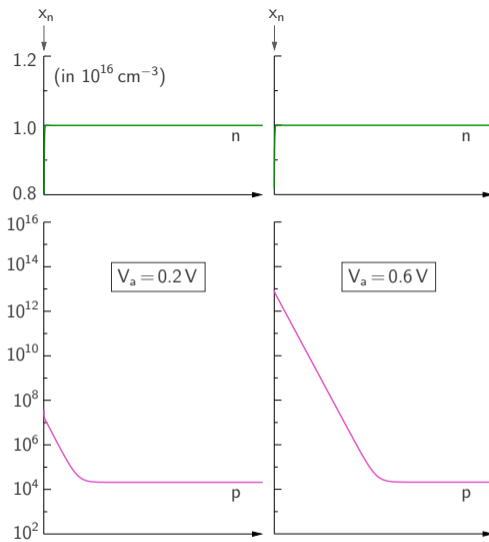


- * As we have seen earlier, the minority carrier diffusion lengths (i.e., L_n on the p -side, L_p on the n -side) are typically much larger than the depletion width.

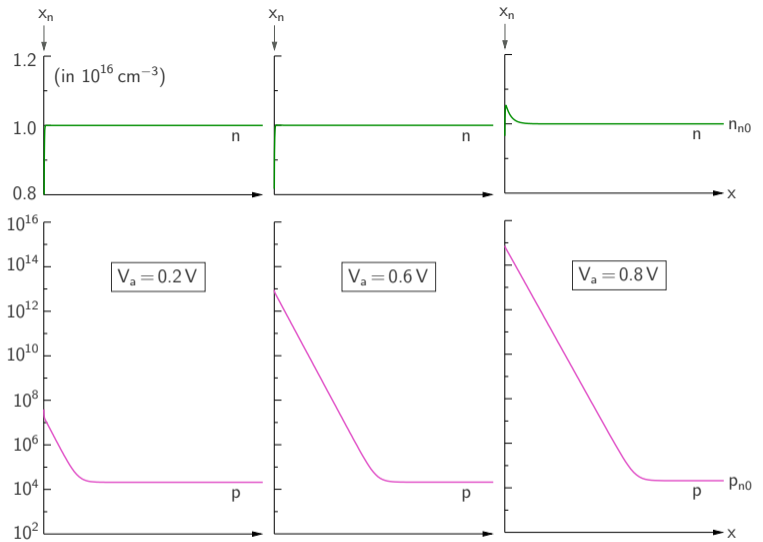
High-injection regime



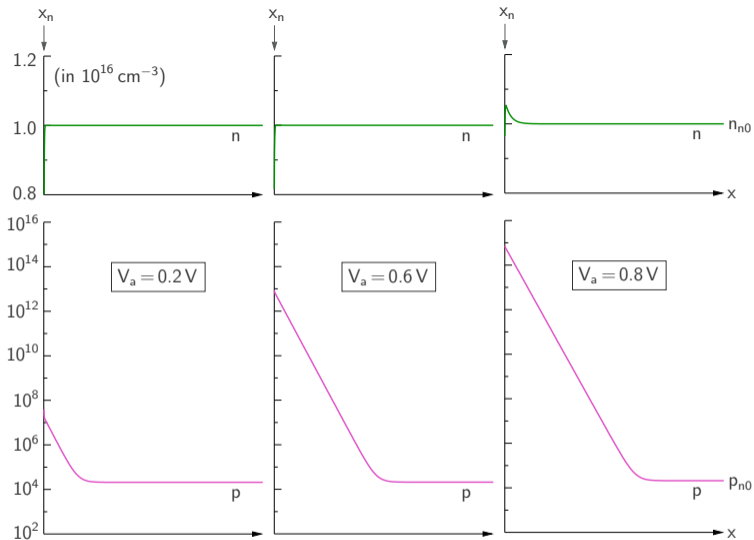
High-injection regime



High-injection regime

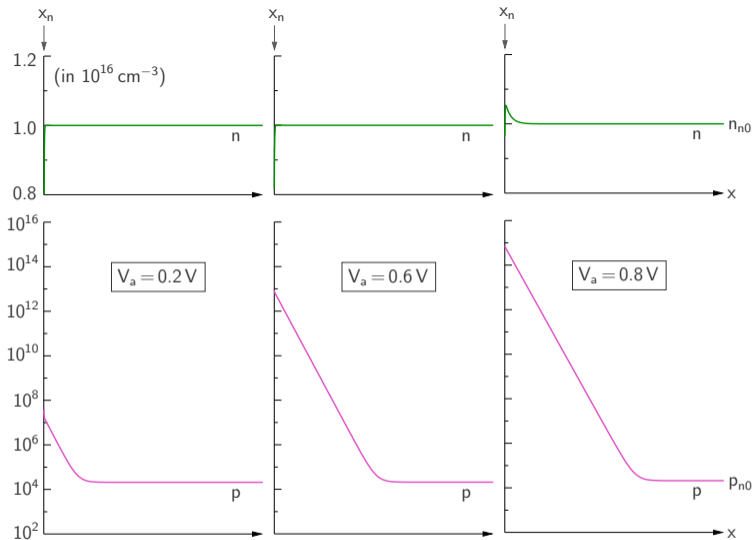


High-injection regime



- * As the forward bias is increased, the minority carrier concentration increases rapidly, and at some point becomes comparable to the majority carrier concentration. This regime is called the “high-injection” regime.

High-injection regime

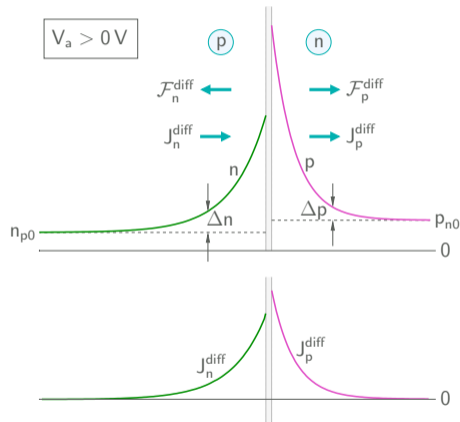


- * As the forward bias is increased, the minority carrier concentration increases rapidly, and at some point becomes comparable to the majority carrier concentration. This regime is called the “high-injection” regime.
- * In the high-injection regime, the majority carrier concentration also increases appreciably (e.g., $\Delta n \approx \Delta p$ on the n side), and the overall charge neutrality is maintained in the neutral regions.

pn junction: current flow under forward bias

$$\Delta p(x) = p_{n0} \left[\exp\left(\frac{V_a}{V_T}\right) - 1 \right] \exp\left(-\frac{x - x_n}{L_p}\right), \quad x > x_n.$$

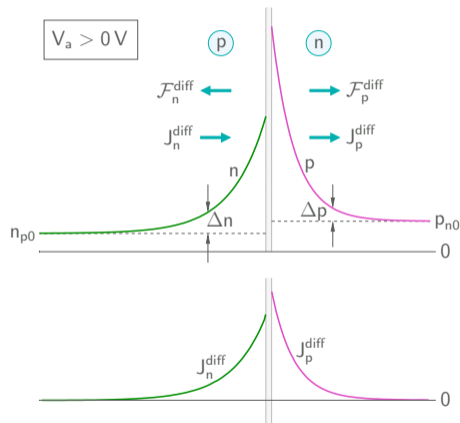
$$\Delta n(x) = n_{p0} \left[\exp\left(\frac{V_a}{V_T}\right) - 1 \right] \exp\left(-\frac{x_p - x}{L_n}\right), \quad x < x_p.$$



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Note that, although $\mathcal{F}_n^{\text{diff}}$ (for $x < x_p$) and $\mathcal{F}_p^{\text{diff}}$ (for $x > x_n$) are in opposite directions, J_n^{diff} (for $x < x_p$) and J_p^{diff} (for $x > x_n$) are in the same direction.



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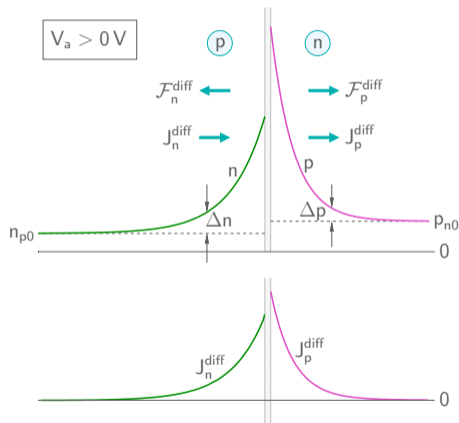
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In particular, we are interested in $J_n^{\text{diff}}(x_p)$ and $J_p^{\text{diff}}(x_n)$.

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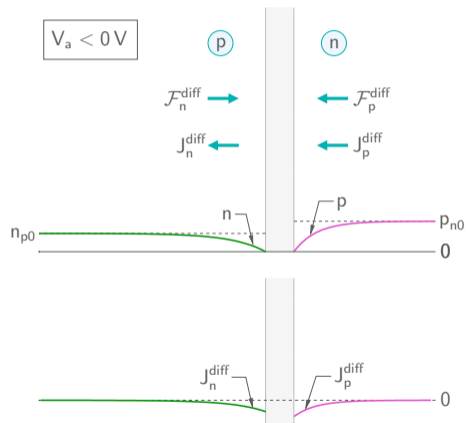
$$J_p^{\text{diff}}(x_n) = \frac{qD_p p_{n0}}{L_p} \left(e^{V_a/V_T} - 1 \right).$$



pn junction: current flow under reverse bias

$$\Delta p(x) = p_{n0} \left[\exp\left(\frac{V_a}{V_T}\right) - 1 \right] \exp\left(-\frac{x - x_n}{L_p}\right), \quad x > x_n.$$

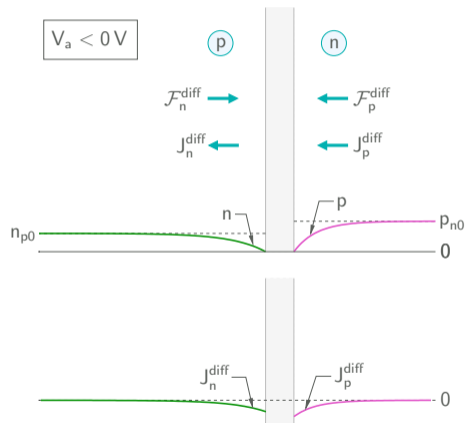
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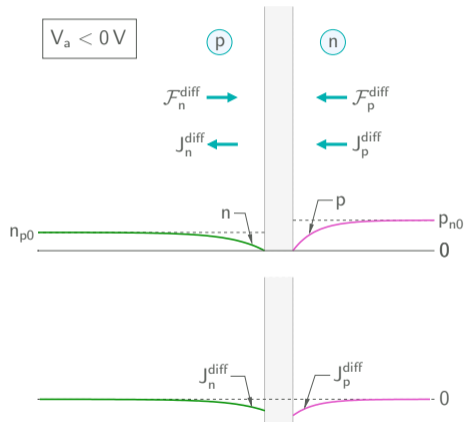
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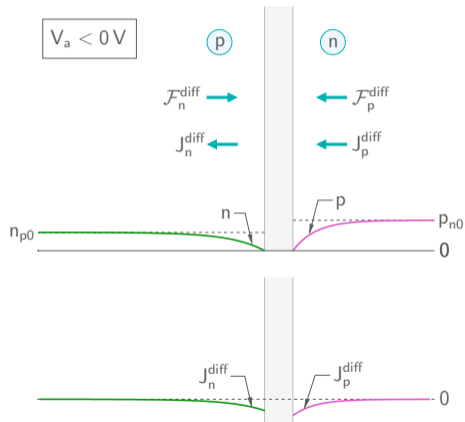
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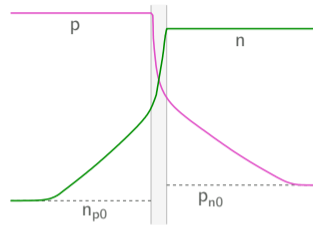
The currents are much smaller under reverse bias.



pn junction: What is happening inside the depletion region?

Consider x in the depletion region, i.e., $x_p < x < x_n$.

$$J_p^{\text{diff}} \approx -J_p^{\text{drift}} \rightarrow \int d\psi = -V_T \int \frac{1}{p} dp \rightarrow \frac{p(x)}{p(x_p)} = \exp \frac{\psi(x_p) - \psi(x)}{V_T}.$$

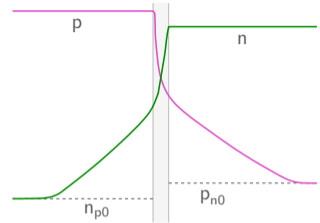


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$$J_n^{\text{diff}} \approx -J_n^{\text{drift}} \rightarrow \int d\psi = +V_T \int \frac{1}{n} dn \rightarrow \frac{n(x)}{n(x_p)} = \exp \frac{\psi(x) - \psi(x_p)}{V_T}.$$



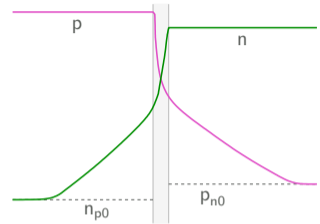
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$$\rightarrow p(x)n(x) = p(x_p)n(x_p) = p_{p0}n_{p0}e^{V_a/V_T} = n_i^2 e^{V_a/V_T}$$



pn junction: What is happening inside the depletion region?

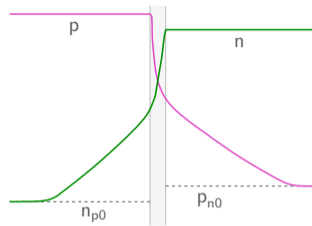
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If $V_a > 0V$, $pn > n_i^2$ in the depletion region; else, $pn < n_i^2$.



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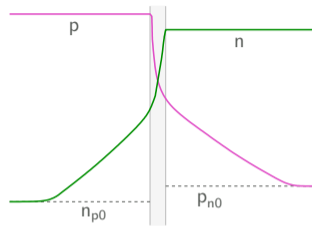
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If $V_a > 0\text{V}$, $pn > n_i^2$ in the depletion region; else, $pn < n_i^2$.

$$R - G = \frac{pn - n_i^2}{\tau_n(n + n_1) + \tau_p(p + p_1)}$$

\rightarrow we have a net recombination inside the depletion region if $V_a > 0\text{V}$, and a net generation if $V_a < 0\text{V}$.



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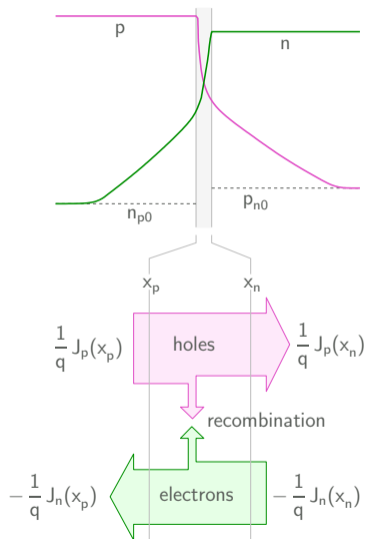
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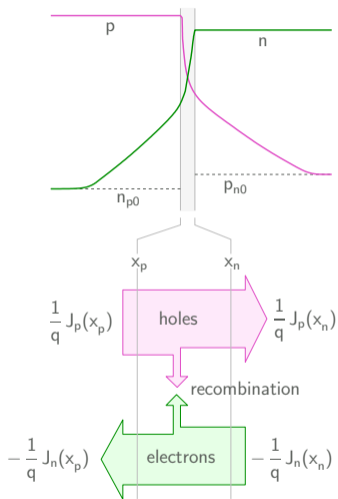
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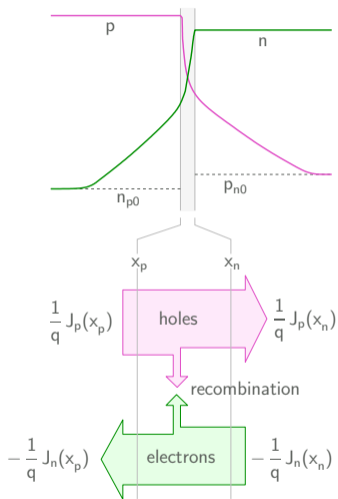
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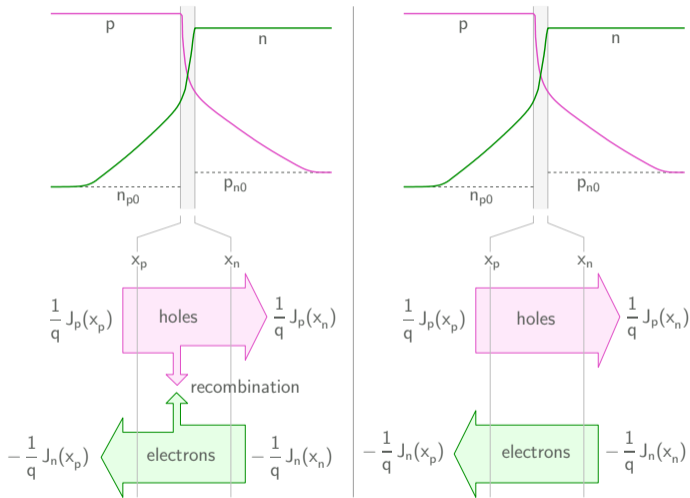


pn junction: What is happening inside the depletion region?



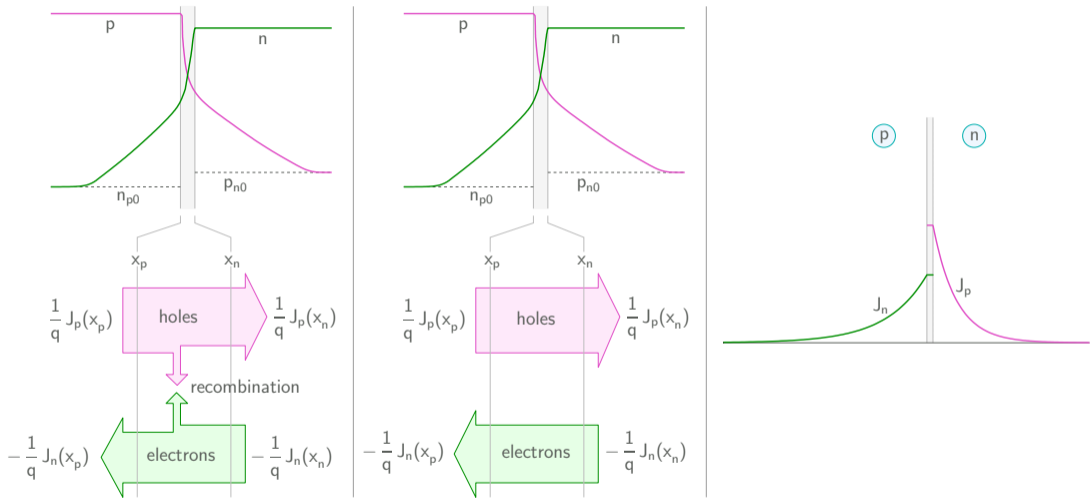
* To obtain a first-order I - V model, we ignore G-R in the depletion region.

pn junction: What is happening inside the depletion region?



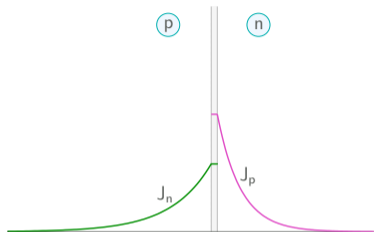
* To obtain a first-order I - V model, we ignore G-R in the depletion region.

pn junction: What is happening inside the depletion region?

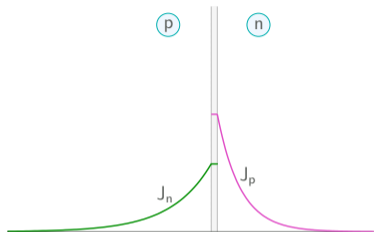


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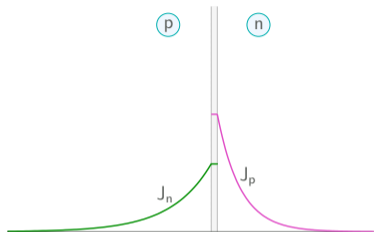
pn junction: total current density



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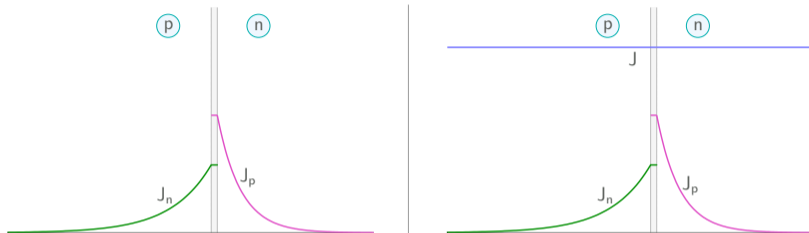


- * The total current density is the same throughout the device.

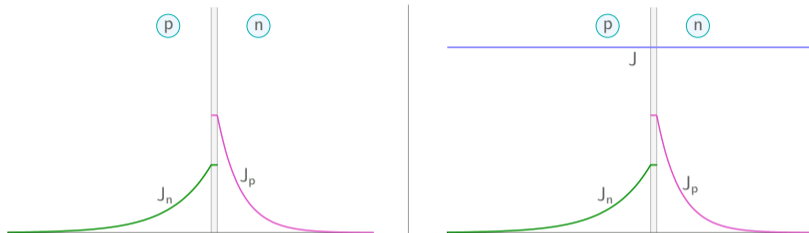


- * The total current density is the same throughout the device.
- * If there is no G-R in the depletion region, we have $J = J_n(x_p) + J_p(x_n)$.

pn junction: total current density

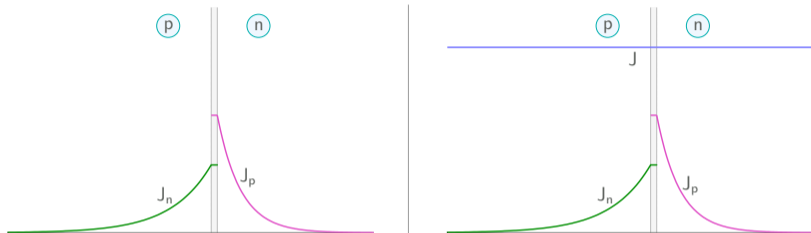


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- * The total current density is the same throughout the device.
- * If there is no G-R in the depletion region, we have $J = J_n(x_p) + J_p(x_n)$.
- * Using our earlier results for $J_p(x_n)$ and $J_n(x_p)$, we get

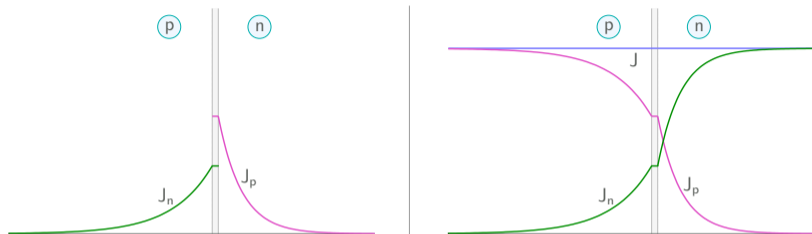
$$J = J_p(x_n) + J_n(x_p) = \left[\frac{qD_p p_{n0}}{L_p} + \frac{qD_n n_{p0}}{L_n} \right] (e^{V_a/V_T} - 1).$$



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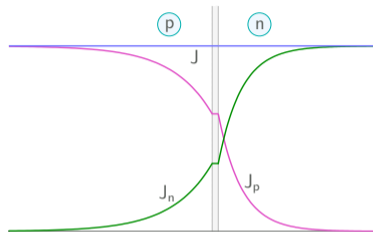
- * We can now obtain $J_n(x > x_n)$ and $J_p(x < x_p)$ using $J_n(x) + J_p(x) = J$.



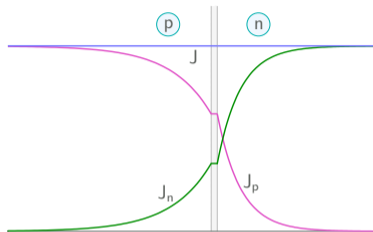
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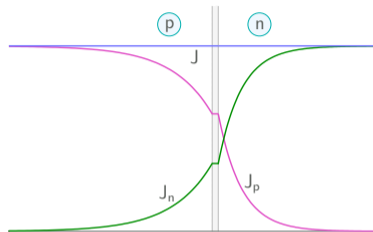


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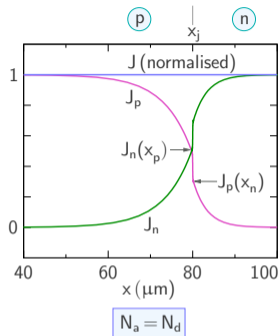
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Consider the situation sufficiently far from the depletion region (i.e., about $5L_n$ on the p -side and $5L_p$ on the n -side).

- * The current density is due to majority carriers (drift component).
- * Since the majority carrier concentration is large, a very small electric field suffices to produce the required current density ($J_n^{\text{drift}} = qn\mu_n\mathcal{E}$, $J_p^{\text{drift}} = qp\mu_p\mathcal{E}$).

pn junction under forward bias: numerical results

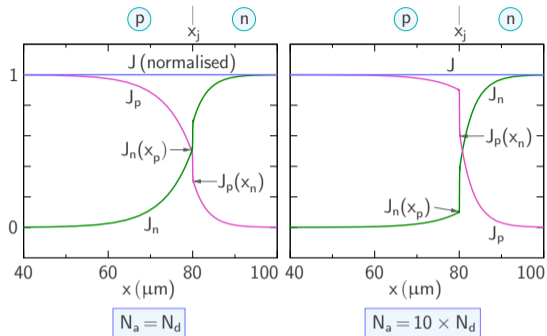


Doping densities:

(1) $N_a = N_d = 10^{16} \text{ cm}^{-3}$

(Parameters: $V_a = 0.5 \text{ V}$, $\mu_n = 1400 \text{ cm}^2/\text{V}\cdot\text{s}$, $\mu_p = 500 \text{ cm}^2/\text{V}\cdot\text{s}$, $\tau_n = 10 \text{ ns}$, $\tau_p = 10 \text{ ns}$, $T = 300 \text{ K}$)

pn junction under forward bias: numerical results



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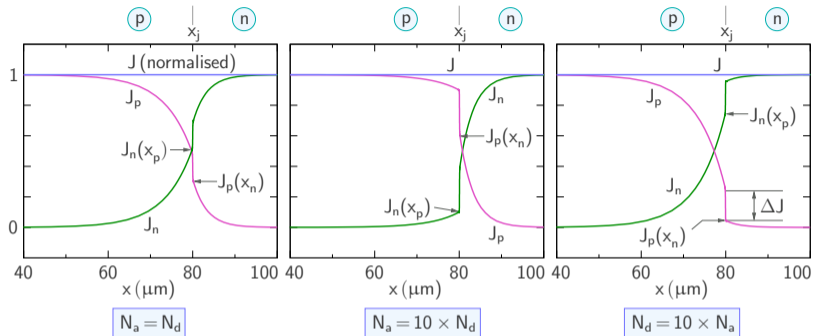
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pn junction under forward bias: numerical results

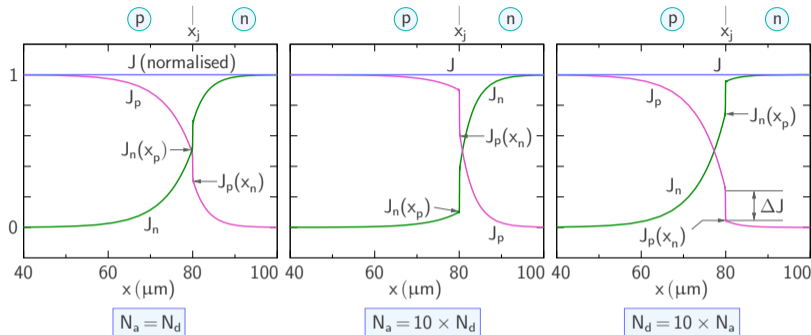


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pn junction under forward bias: numerical results



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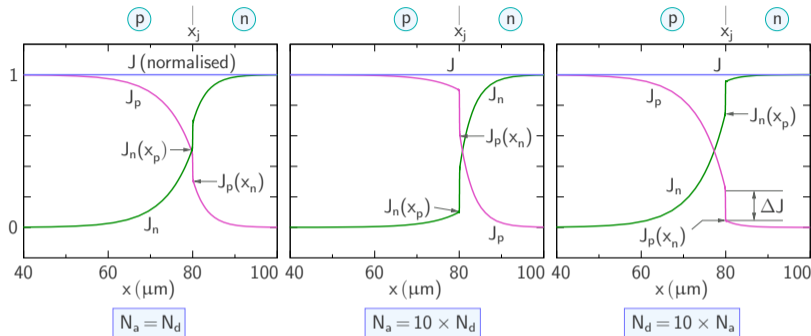
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* The ratio $\frac{J_p(x_n)}{J_n(x_p)}$ is $\frac{D_p}{D_n} \frac{L_n}{L_p} \frac{p_{n0}}{n_{p0}} = \frac{D_p}{D_n} \frac{\sqrt{D_n \tau_n}}{\sqrt{D_p \tau_p}} \frac{N_a}{N_d} = \sqrt{\frac{\mu_p \tau_n}{\mu_n \tau_p}} \frac{N_a}{N_d}$.

pn junction under forward bias: numerical results



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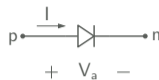
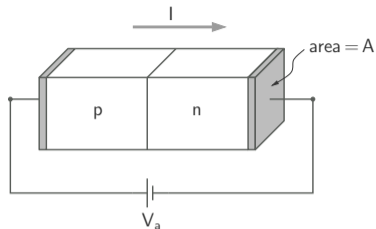
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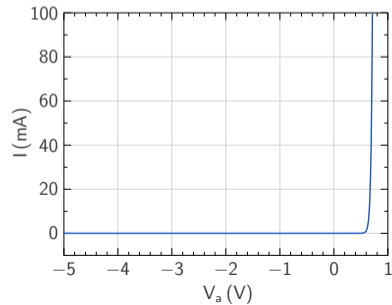
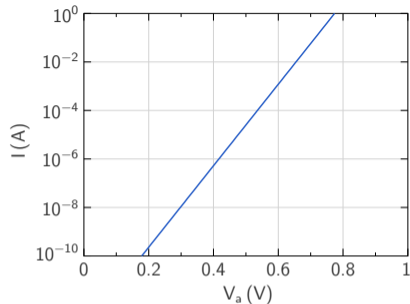
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- * Because of recombination, there is a change in J_p and J_n across the depletion region (which has been ignored in our analysis). This change is seen as vertical lines in the figure since the depletion width is much smaller than the diffusion lengths.

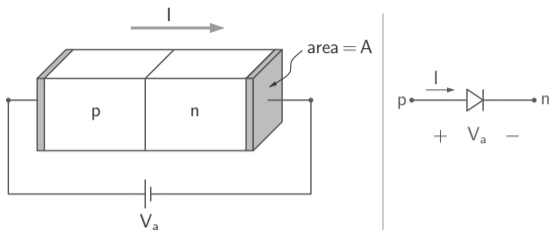
Diode I-V equation



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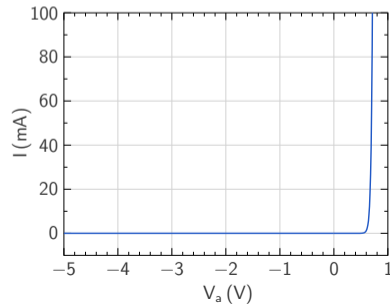
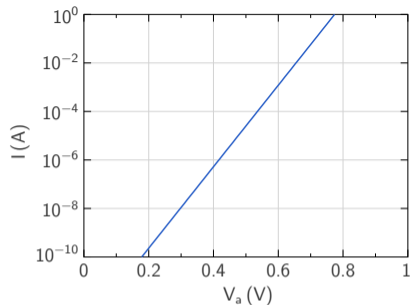


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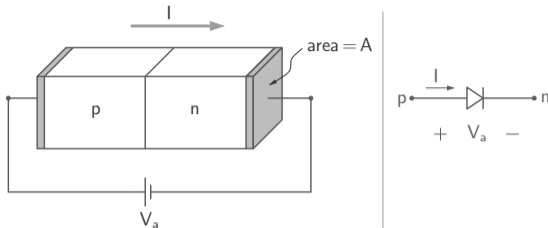


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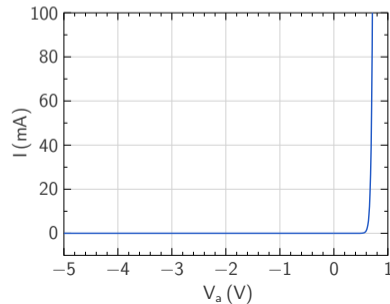
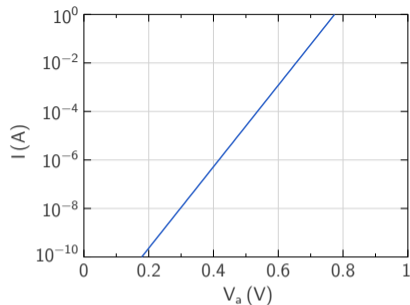
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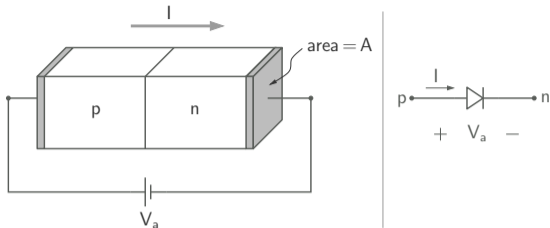
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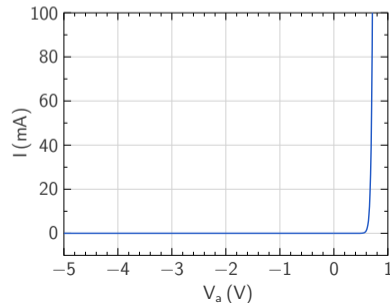
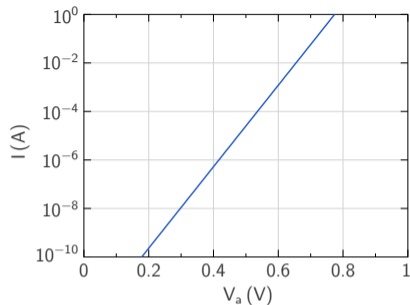
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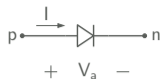
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- * This equation is known as the “Shockley diode equation.”
- * Under reverse bias, with V_R equal to a few V_T or larger, $e^{V_a/V_T} = e^{-V_R/V_T} \approx 0$, and $I \approx -I_s$, i.e., the diode current “saturates” (at $-I_s$). I_s is therefore called the “reverse saturation current.”

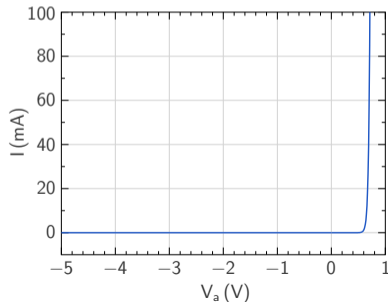
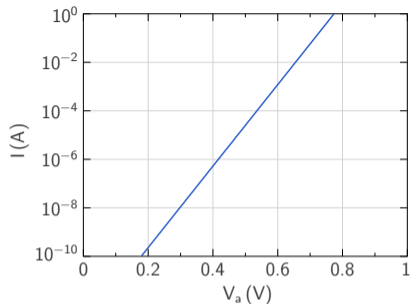


Diode I - V equation

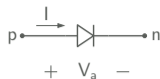


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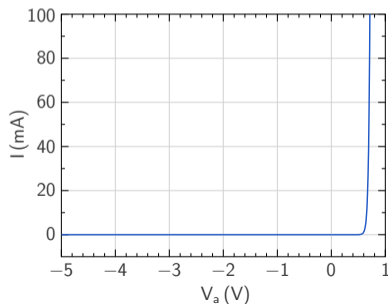
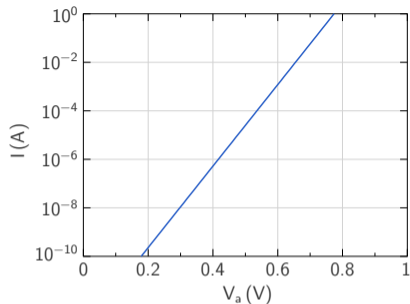
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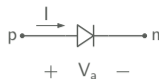
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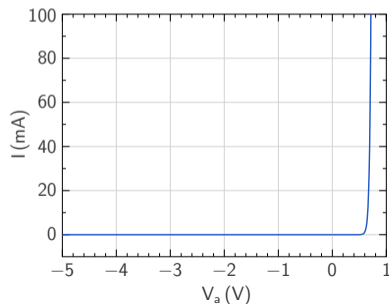
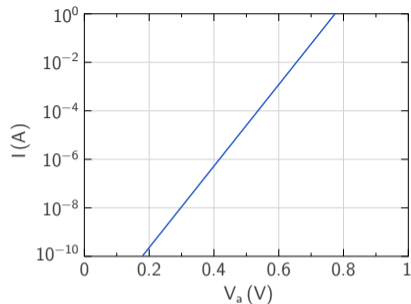


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- * In a real diode, other factors often dominate in reverse bias, including generation in the depletion region and surface leakage. Also, as we will see, a real diode cannot withstand indefinitely large reverse voltages and will “break down” at some point.
- * Recombination in the depletion region under forward bias can be incorporated in the Shockley equation with an “ideality factor” η ($1 < \eta < 2$):

$$I = I_{s1} \exp\left(\frac{V_a}{\eta_1 V_T}\right) + I_{s2} \exp\left(\frac{V_a}{\eta_2 V_T}\right)$$
$$\approx I_s^{\text{eff}} \exp\left(\frac{V_a}{\eta V_T}\right)$$



Example

For an abrupt, uniformly doped silicon pn junction diode, $N_a = 10^{17} \text{ cm}^{-3}$, $N_d = 2 \times 10^{16} \text{ cm}^{-3}$, $\mu_n = 1500 \text{ cm}^2/\text{V-s}$, $\mu_p = 500 \text{ cm}^2/\text{V-s}$, $\tau_n = 2 \text{ }\mu\text{s}$, $\tau_p = 5 \text{ }\mu\text{s}$, $A = 10^{-3} \text{ cm}^2$. Compute the following for a forward bias of 0.65 V at $T = 300 \text{ K}$:

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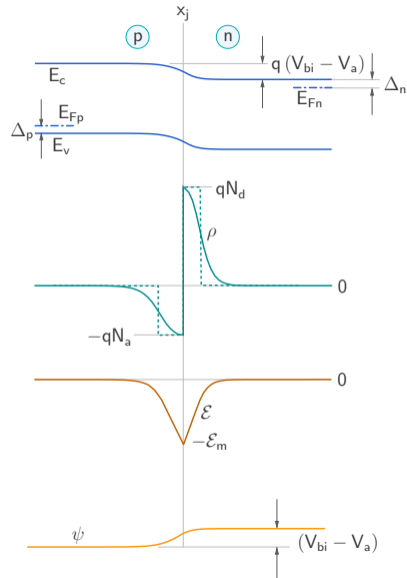
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Example (continued)

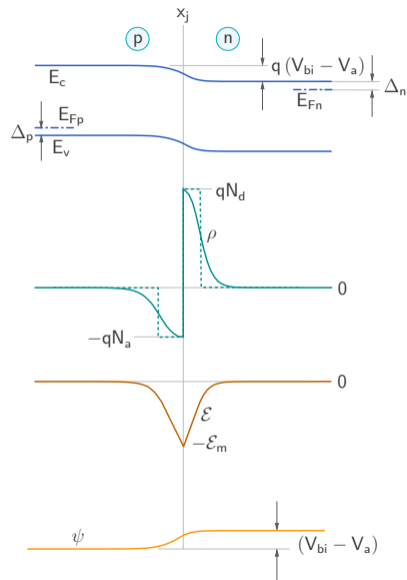
$$V_{bi} = V_T \log \left(\frac{N_a N_d}{n_i^2} \right) = (0.0259 \text{ V}) \log \frac{(10^{17})(2 \times 10^{16})}{(1.5 \times 10^{10})^2} = 0.77 \text{ V}.$$



Example (continued)

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$$W = \sqrt{\frac{2\epsilon}{q} \frac{N_a + N_d}{N_a N_d} (V_{bi} - V_a)}$$

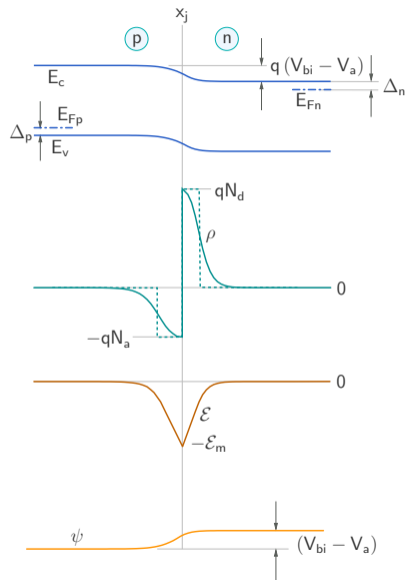


Example (continued)

$$V_{bi} = V_T \log \left(\frac{N_a N_d}{n_i^2} \right) = (0.0259 \text{ V}) \log \frac{(10^{17})(2 \times 10^{16})}{(1.5 \times 10^{10})^2} = 0.77 \text{ V}.$$

$$W = \sqrt{\frac{2\epsilon}{q} \frac{N_a + N_d}{N_a N_d} (V_{bi} - V_a)}$$

$$= \sqrt{\frac{2 \times 11.8 \times 8.85 \times 10^{-14}}{1.6 \times 10^{-19}} \frac{1.2 \times 10^{17}}{2 \times 10^{33}} \times 0.12} \text{ cm} = 0.097 \mu\text{m}.$$



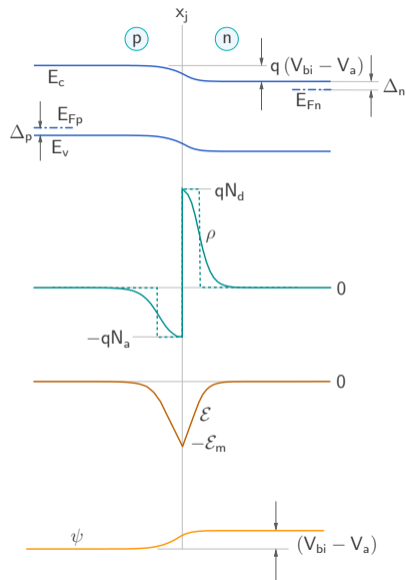
Example (continued)

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$$(V_{bi} - V_a) = \frac{1}{2} \epsilon_m W$$



Example (continued)

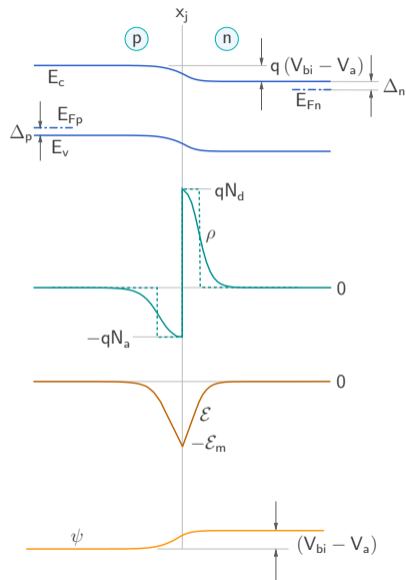
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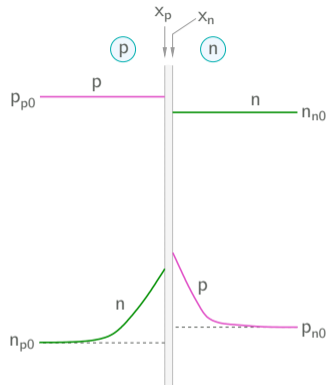
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$$(V_{bi} - V_a) = \frac{1}{2} \epsilon_m W$$

$$\rightarrow \epsilon_m = \frac{2(V_{bi} - V_a)}{W} = \frac{2 \times 0.12 \text{ V}}{0.097 \times 10^{-4} \text{ cm}} = 25 \text{ kV/cm}.$$



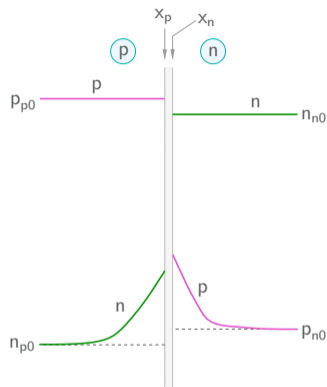
Example (continued)



Example (continued)

The equilibrium minority carrier densities are

$$p_{n0} = \frac{n_i^2}{n_{n0}} \approx \frac{n_i^2}{N_d} = \frac{(1.5 \times 10^{10})^2}{2 \times 10^{16}} = 1.125 \times 10^4 \text{ cm}^{-3},$$

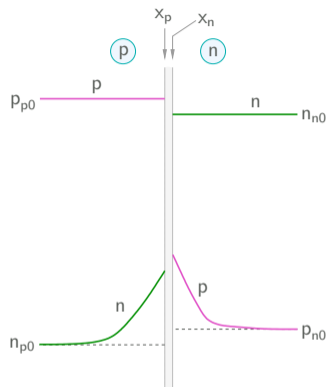


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Example (continued)

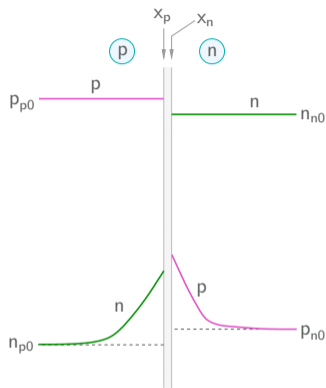
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The minority carrier densities at x_p and x_n are

$$n(x_p) = n_{p0} e^{V_a/V_T} = 2.25 \times 10^3 \times e^{0.65/0.0259} = 1.8 \times 10^{14} \text{ cm}^{-3},$$



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The equilibrium minority carrier densities are

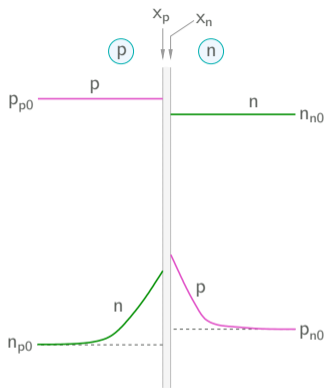
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Example (continued)

The diffusion coefficients are

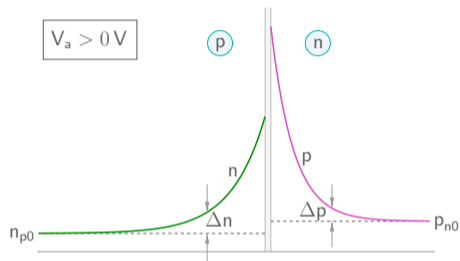
$$D_p = V_T \mu_p = 0.0259 \times 500 = 12.9 \text{ cm}^2/\text{s},$$

$$D_n = V_T \mu_n = 0.0259 \times 1500 = 38.7 \text{ cm}^2/\text{s}.$$

The minority carrier diffusion lengths in the neutral regions are

$$L_p = \sqrt{D_p \tau_p} = \sqrt{12.9 \times 5 \times 10^{-6}} \text{ cm} = 80.3 \text{ } \mu\text{m},$$

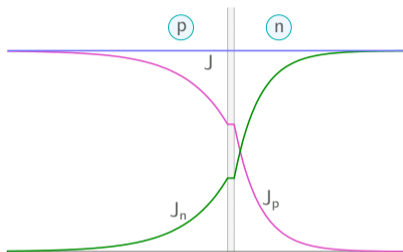
$$L_n = \sqrt{D_n \tau_n} = \sqrt{38.7 \times 2 \times 10^{-6}} \text{ cm} = 88 \text{ } \mu\text{m}.$$



Example (continued)

The minority carrier current densities at x_n and x_p are

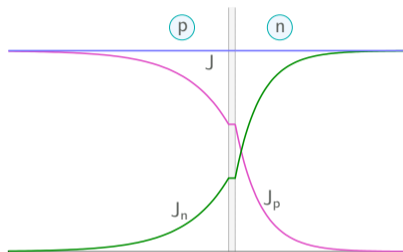
$$J_p(x_n) = \frac{qD_p p_{n0}}{L_p} (e^{V_a/V_T} - 1)$$



Example (continued)

The minority carrier current densities at x_n and x_p are

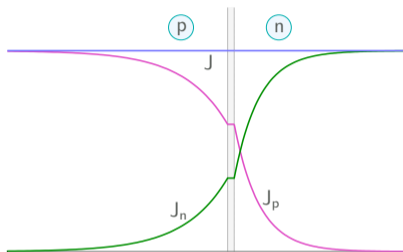
$$\begin{aligned} J_p(x_n) &= \frac{qD_p p_{n0}}{L_p} (e^{V_a/V_T} - 1) \\ &= \frac{1.6 \times 10^{-19} \times 12.9 \times 1.125 \times 10^4}{80.3 \times 10^{-4}} \times 8.12 \times 10^{10} \end{aligned}$$



Example (continued)

The minority carrier current densities at x_n and x_p are

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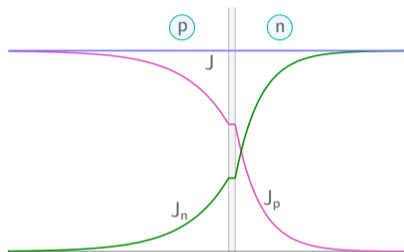


Example (continued)

The minority carrier current densities at x_n and x_p are

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Example (continued)

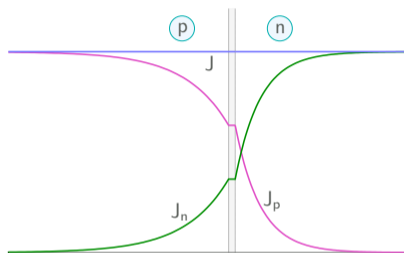
The minority carrier current densities at x_n and x_p are

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The diode current I is

$$I = A(J_p(x_n) + J_n(x_p))$$



Example (continued)

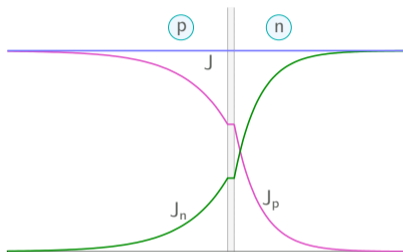
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The diode current I is

$$\begin{aligned} I &= A(J_p(x_n) + J_n(x_p)) \\ &= 10^{-3} \text{ cm}^2 \times (0.235 + 0.13) \text{ A/cm}^2 \end{aligned}$$



Example (continued)

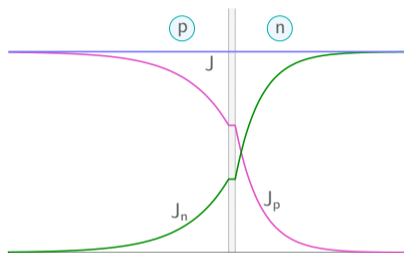
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$$\begin{aligned}I &= A(J_p(x_n) + J_n(x_p)) \\&= 10^{-3} \text{ cm}^2 \times (0.235 + 0.13) \text{ A/cm}^2 \\&= 0.365 \text{ mA}.\end{aligned}$$

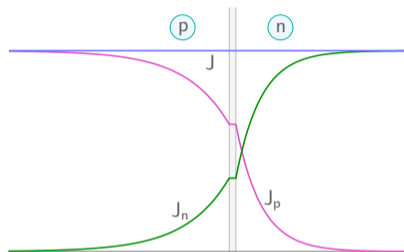


Example (continued)

In the neutral n region more than $5L_p$ away from the depletion region,

$J \approx J_n = q\mu_n \mathcal{E}_{\text{neutral}}^n n_{n0}$, leading to

$$\mathcal{E}_{\text{neutral}}^n = \frac{J}{q\mu_n n_{n0}} = \frac{0.365 \left[\frac{\text{A}}{\text{cm}^2} \right]}{1.6 \times 10^{-19} [\text{C}] \times 1500 \left[\frac{\text{cm}^2}{\text{V-s}} \right] \times 2 \times 10^{16} \left[\frac{1}{\text{cm}^3} \right]}$$

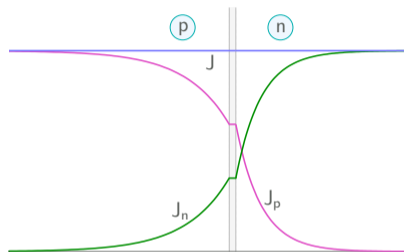


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Example (continued)

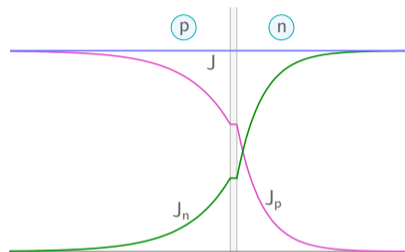
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Similarly,

$$\mathcal{E}_{\text{neutral}}^p = \frac{J}{q\mu_p p_{p0}} = \frac{0.365 \left[\frac{\text{A}}{\text{cm}^2} \right]}{1.6 \times 10^{-19} [\text{C}] \times 500 \left[\frac{\text{cm}^2}{\text{V}\cdot\text{s}} \right] \times 1 \times 10^{17} \left[\frac{1}{\text{cm}^3} \right]}$$



Example (continued)

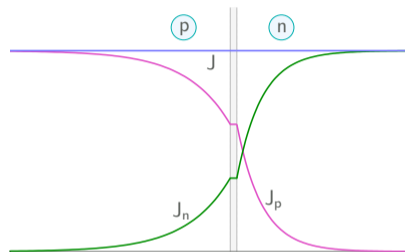
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Example (continued)

In the neutral n region more than $5L_p$ away from the depletion region,

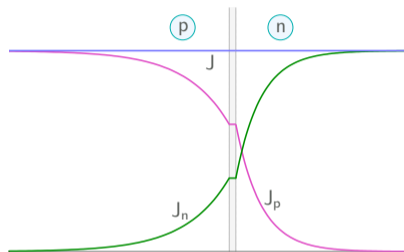
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Note that these values are much smaller than \mathcal{E}_m in the depletion region (25 kV/cm).



The reverse saturation current I_s is given by

$$I_s = A \times \left(\frac{qD_p p_{n0}}{L_p} + \frac{qD_n n_{p0}}{L_n} \right)$$

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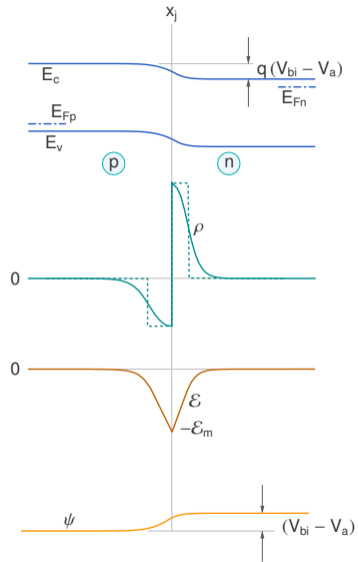
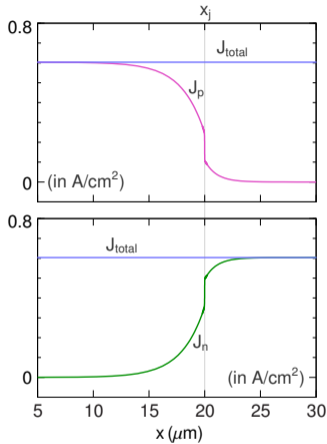
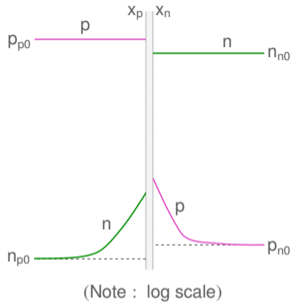
$$\begin{aligned} I_s &= A \times \left(\frac{qD_p p_{n0}}{L_p} + \frac{qD_n n_{p0}}{L_n} \right) \\ &= 10^{-3} \left(\frac{1.6 \times 10^{-19} \times 12.9 \times 1.125 \times 10^4}{80.3 \times 10^{-4}} + \frac{1.6 \times 10^{-19} \times 38.7 \times 2.25 \times 10^3}{88 \times 10^{-4}} \right) \\ &= 4.5 \times 10^{-15} \text{ A (i.e., 4.5 fA)}. \end{aligned}$$

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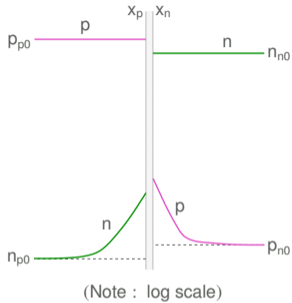
$$\begin{aligned}
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 &= 4.5 \times 10^{-15} \text{ A (i.e., 4.5 fA)}.
 \end{aligned}$$

Note how small I_s is. The only reason we can get significant currents (\sim mA) in forward bias is the *huge* exponential factor (e^{V_a/V_T}) in the Shockley equation.

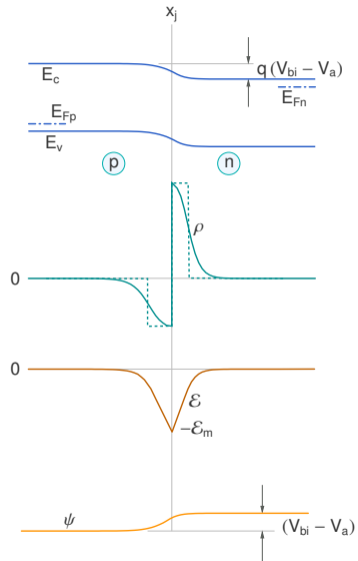
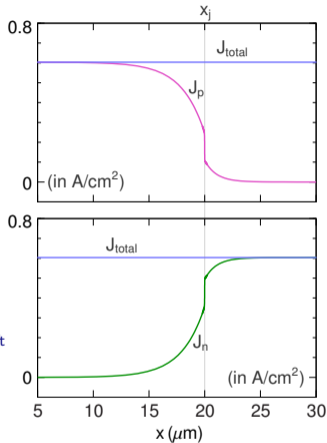
Forward bias: summary



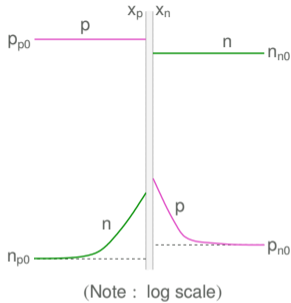
Forward bias: summary



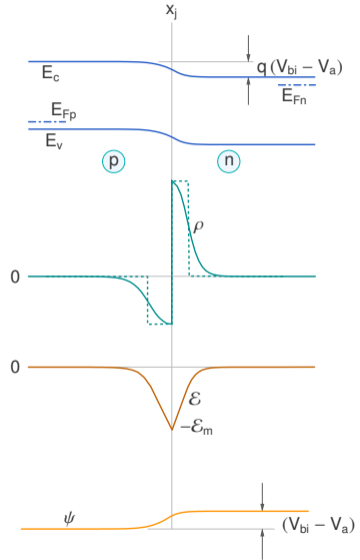
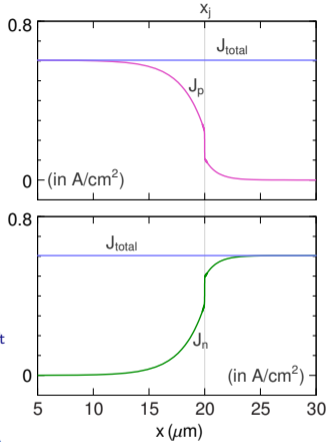
* On the p side, $5 L_n$ away from x_p , $J \approx J_p^{\text{drift}}$ resulting from a small (positive) electric field.



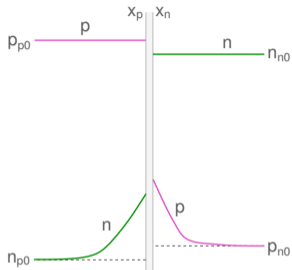
Forward bias: summary



- * On the p side, $5 L_n$ away from x_p , $J \approx J_p^{drift}$ resulting from a small (positive) electric field.
- * Holes continuously injected from the p side recombine with electrons on the n side.

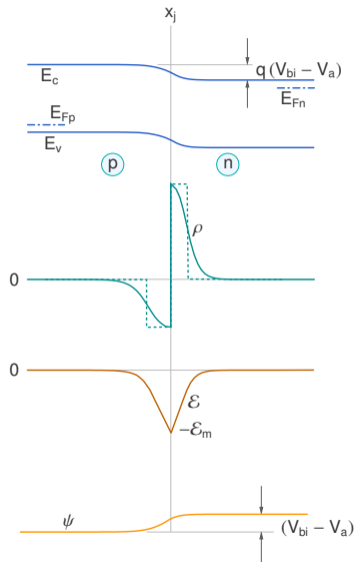
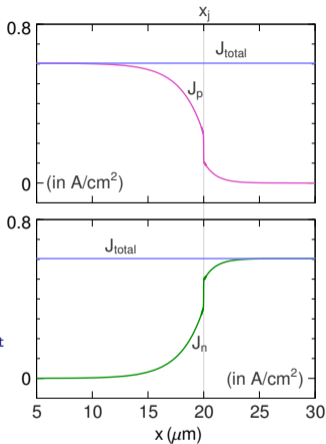


Forward bias: summary

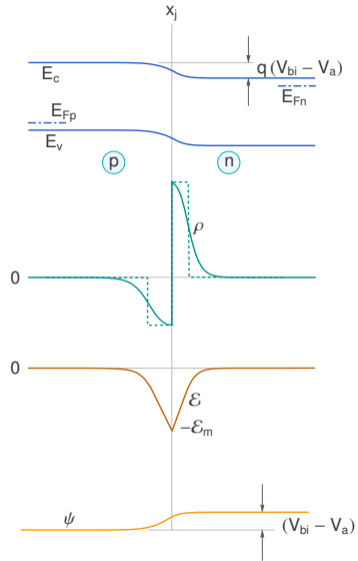
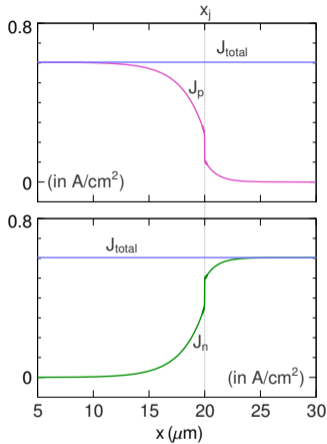
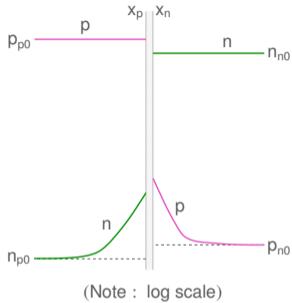


(Note : log scale)

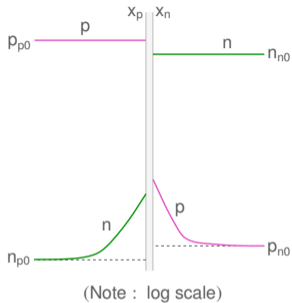
- * On the p side, $5 L_n$ away from x_p , $J \approx J_p^{\text{drift}}$ resulting from a small (positive) electric field.
- * Holes continuously injected from the p side recombine with electrons on the n side.
- * $5 L_p$ away from x_n , there are no excess holes, and $p \approx p_{n0}$.



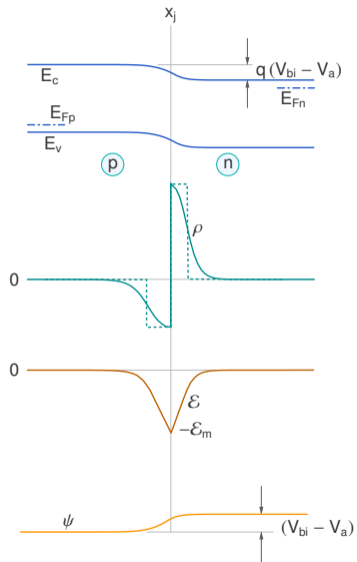
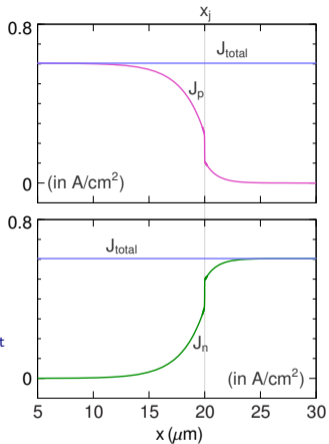
Forward bias: summary



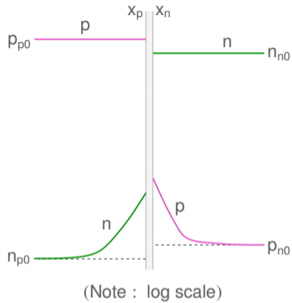
Forward bias: summary



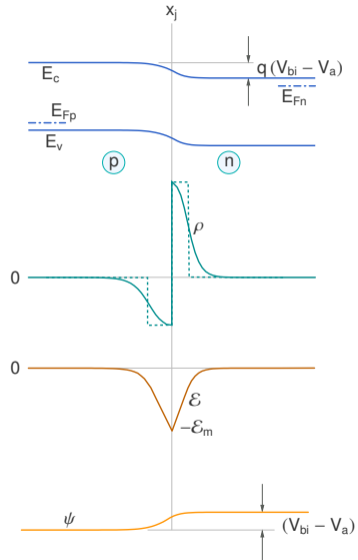
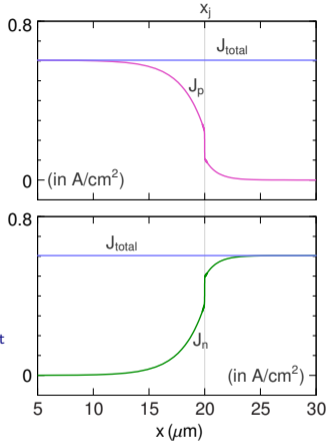
* On the n side, $5 L_p$ away from x_n , $J \approx J_n^{\text{drift}}$ resulting from a small (positive) electric field.



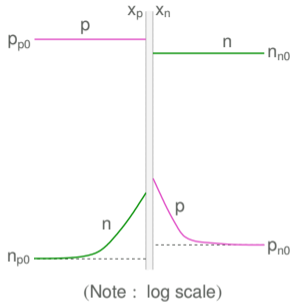
Forward bias: summary



- * On the n side, $5 L_p$ away from x_n , $J \approx J_n^{\text{drift}}$ resulting from a small (positive) electric field.
- * Electrons continuously injected from the n side recombine with holes on the p side.



Forward bias: summary



- * On the n side, $5 L_p$ away from x_n , $J \approx J_n^{\text{drift}}$ resulting from a small (positive) electric field.
- * Electrons continuously injected from the n side recombine with holes on the p side.
- * $5 L_n$ away from x_p , there are no excess electrons, and $n \approx n_{p0}$.

