

bjt\_widlar.sqproj

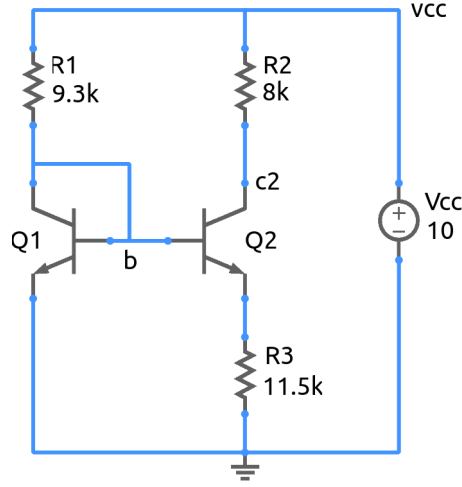


Figure 1: BJT Widlar current source.

Fig. 1 shows the circuit diagram of a Widlar current source. In a simple current mirror (see [bjt\\_mirror\\_1.pdf](#)), the mirrored current cannot be very different from the reference current<sup>1</sup>, certainly not by orders of magnitude. In a Widlar current source, a current much smaller than the reference current can be obtained.

Assuming the BJTs to be operating in the active region, we get

$$I_{R1} \approx I_{C1} = \frac{V_{CC} - 0.7}{R_1}. \quad (1)$$

Since  $I_C \approx I_s e^{V_{BE}/V_T}$  in the active mode, we have  $V_{BE} = V_T \ln(I_C/I_s)$  for each of the transistors. Using these equations, we can write KVL for the loop involving the two B-E junctions as

$$V_{BE1} = V_{BE2} + I_{C2} R_3 \rightarrow V_T \ln(I_{C1}/I_{s1}) = V_T \ln(I_{C2}/I_{s2}) + I_{C2} R_3. \quad (2)$$

If the two transistors are identical, we have

$$V_T \ln(I_{C1}/I_{C2}) = I_{C2} R_3. \quad (3)$$

By selecting an appropriate value of  $R_3$ , the desired  $I_{C2}$  can be obtained.

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<sup>1</sup>The two currents are related by the ratio of the areas of the two transistors.

## Exercise Set

1. For the component values given in the figure, compute  $I_{C1}$  and  $I_{C2}$ . Verify with simulation. (Note that computing  $I_{C2}$  will involve solving the above nonlinear equation (Eq. 3) in an iterative manner.)
2. Calculate  $R_3$  required for  $I_{C2} = 50 \mu\text{A}$ . Verify with simulation.

## References

1. A. S. Sedra, K. C. Smith, and A. N. Chandorkar, *Microelectronic Circuits: Theory and Applications*, Fifth edition, Oxford University Press, 2009.
2. P. R. Grey and R. G. Meyer, *Analysis and Design of Analog Integrated Circuits*, John Wiley and Sons, 1995.
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